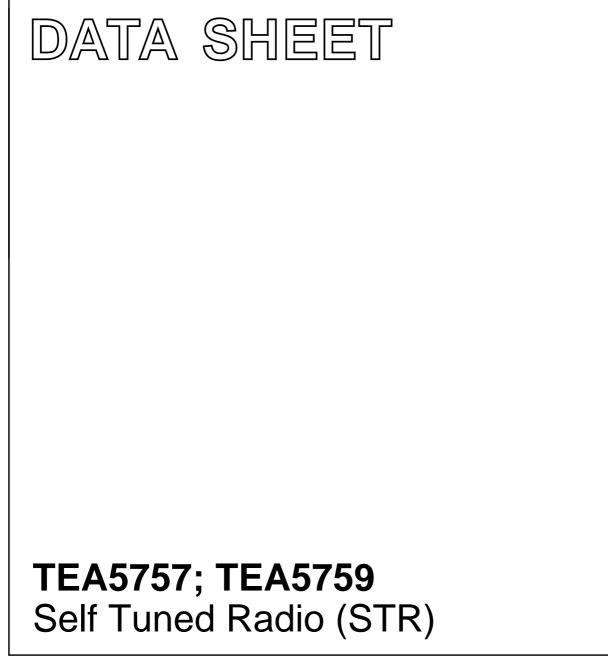
INTEGRATED CIRCUITS



Product specification Supersedes data of 1996 Jan 09 File under Integrated Circuits, IC01

1999 Aug 26



HILIP

### TEA5757; TEA5759

#### FEATURES

- The tuning system has an optimized IC partitioning both from application (omitting interferences) and flexibility (removable front panel option) point of view: the tuning synthesizer is on-chip with the radio
- The tuning quality is superior and requires no IF-counter for stop-detection; it is insensitive to ceramic filter tolerances
- In combination with the microcontroller, fast, low-power operation of preset mode, manual-search, auto-search and auto-store are possible
- The local (internal) controller function facilitates reduced and simplified microcontroller software
- The high integration level (radio and tuning synthesizer on one chip) means fewer external components with regard to the communication between the radio and the microcontroller (90% less components compared to the digital tuning application of a radio IC with external PLL tuning function) and a simple and small printed-circuit board
- There will be no application considerations for the tuning system, with regards to quality and high integration level, since there will be no external 110 MHz buffers, loop filter or false lock elimination
- The inherent FUZZY LOGIC behaviour of the Self Tuned Radio (STR), which mimics hand tuning, yields a potentially fast yet reliable tuning operation
- The level of the incoming signal at which the radio must lock is software programmable
- Two programmable ports

- High selectivity with distributed IF gain
- Soft mute
- Signal dependent stereo-blend
- High impedance MOSFET input on AM
- Wide supply voltage range of 2.5 to 12 V
- Low current consumption 18 mA at AM and FM (including tuning synthesizer)
- · High input sensitivity
- Low output distortion
- Due to the new tuning concept, the tuning is independent of the channel spacing.

#### **GENERAL DESCRIPTION**

The TEA5757; TEA5759 is a 44-pin integrated AM/FM stereo radio circuit including a novel tuning concept. The radio part is based on the TEA5712.

The TEA5757 is used in FM-standards in which the local oscillator frequency is above the radio frequency (e.g. European and American standards).

The TEA5759 is the version in which the oscillator frequency is below the radio frequency (e.g. Japanese standard).

The new tuning concept combines the advantages of hand tuning with electronic facilities and features. User 'intelligence' is incorporated into the tuning algorithm and an improvement of the analog signal processing is used for the AFC function.

# PACKAGE TYPE NUMBER NAME DESCRIPTION VERSION TEA5757H QFP44 plastic quad flat package; 44 leads (lead length 1.3 mm); body 10 × 10 × 1.75 mm SOT307-2

#### **ORDERING INFORMATION**

#### TEA5757; TEA5759

#### QUICK REFERENCE DATA

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V <sub>CC1</sub>	supply voltage		2.5	-	12	V
V <sub>CC2</sub>	supply voltage for tuning		-	-	12	V
V <sub>tune</sub>	tuning voltage		0.7	-	V <sub>CC2</sub> - 0.75	V
I <sub>CC1</sub>	supply current	AM mode	12	15	18	mA
		FM mode	13	16	19	mA
I <sub>DD</sub>	supply current	AM mode	_	3.3	-	mA
		FM mode	-	2.7	-	mA
I <sub>CC2</sub>	supply current for tuning in preset mode (band-end to band-end)		-	-	800	μA
T <sub>amb</sub>	ambient temperature		-15	-	+60	°C
AM perfor	mance; note 1					•
V <sub>10</sub>	AF output voltage	V <sub>i1</sub> = 5 mV	36	45	70	mV
V <sub>i1</sub>	RF sensitivity input voltage	(S+N)/N = 26 dB	40	55	70	μV
THD	total harmonic distortion	V <sub>i1</sub> = 1 mV	-	0.8	2.0	%
FM perfor	mance; note 2					•
V <sub>10</sub>	AF output voltage	V <sub>i5</sub> = 1 mV	40	48	57	mV
V <sub>i5</sub>	RF limiting sensitivity	V <sub>10</sub> at –3 dB; V <sub>10</sub> is 0 dB at V <sub>i5</sub> = 1 mV	0.4	1.2	3.8	μV
THD	total harmonic distortion	IF filter SFE10.7MS3A20K-A	-	0.3	0.8	%
MPX perfe	brmance; note 3	-				
α <sub>cs</sub>	channel separation		26	30	_	dB

Notes

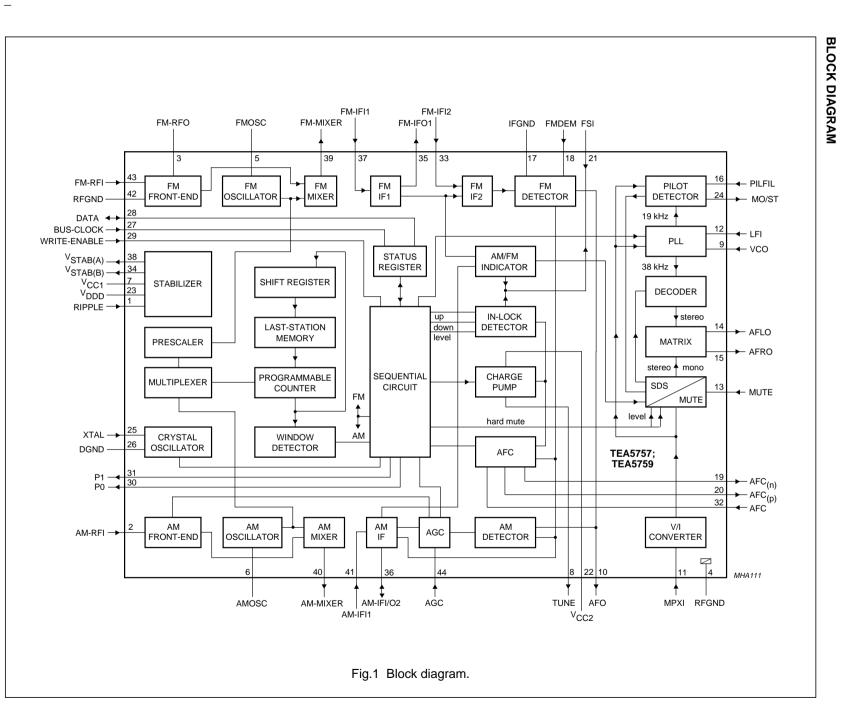
- V<sub>CC1</sub> = 3 V; V<sub>CC2</sub> = 12 V; V<sub>DDD</sub> = 3 V; f<sub>i</sub> = 1 MHz; m = 0.3; f<sub>m</sub> = 1 kHz; measured in Fig.9 with S1 in position A and S2 in position B; V<sub>n</sub> refers to pin voltages; V<sub>i(n)</sub> refers to test circuit (see Fig.9).
- 2.  $V_{CC1} = 3 \text{ V}; V_{CC2} = 12 \text{ V}; V_{DDD} = 3 \text{ V}; f_i = 100 \text{ MHz}; \Delta f_m = 22.5 \text{ kHz}; f_m = 1 \text{ kHz}; measured in Fig.9 with S2, S3 and S5 in position A; V_n refers to pin voltages; V_{i(n)} refers to test circuit (see Fig.9).$
- 3.  $V_{CC1} = 3 \text{ V}$ ;  $V_{CC2} = 12 \text{ V}$ ;  $V_{DDD} = 3 \text{ V}$ ;  $V_{i3(L + R)} = 155 \text{ mV}$ ;  $V_{pilot} = 15.5 \text{ mV}$ ;  $f_i = 1 \text{ kHz}$ ; measured in Fig.9 with S2 and S3 in position B.

Philips Semiconductors

Product specification

# Self Tuned Radio (STR)

TEA5757; TEA5759



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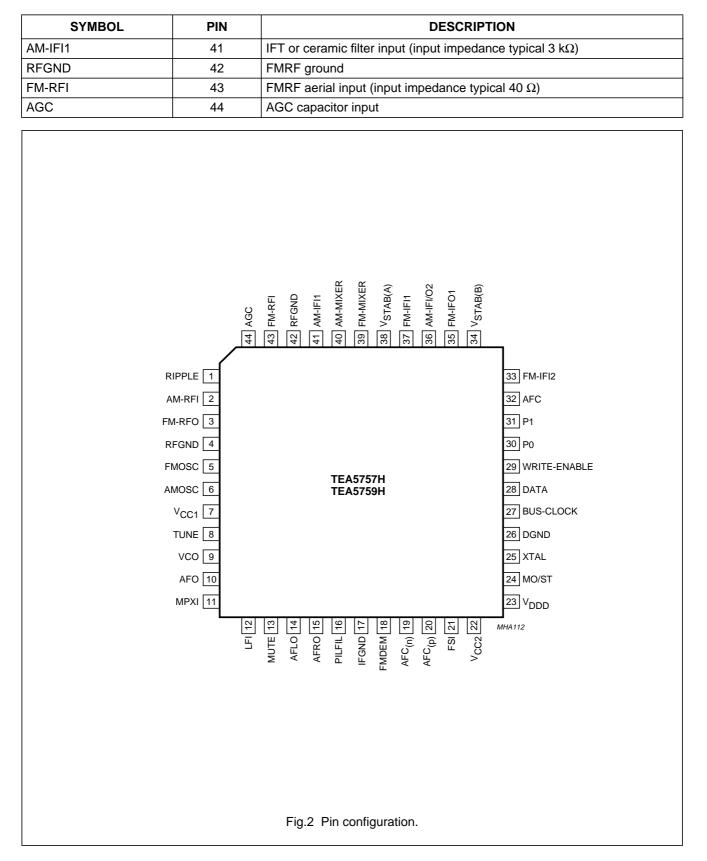
# TEA5757; TEA5759

#### PINNING

SYMBOL	PIN	DESCRIPTION
RIPPLE	1	ripple capacitor input
AM-RFI	2	AMRF input
FM-RFO	3	parallel tuned FMRF circuit to ground
RFGND	4	RF ground and substrate
FMOSC	5	parallel tuned FM-oscillator circuit to ground
AMOSC	6	parallel tuned AM-oscillator circuit to ground
V <sub>CC1</sub>	7	supply voltage
TUNE	8	tuning current output
VCO	9	voltage controlled oscillator input
AFO	10	AM/FM AF output (output impedance typical 5 k $\Omega$ )
MPXI	11	stereo decoder input (input impedance typical 150 k $\Omega$ )
LFI	12	loop-filter input
MUTE	13	mute input
AFLO	14	left channel output (output impedance typical 4.3 k $\Omega$ )
AFRO	15	right channel output (output impedance typical 4.3 k $\Omega$ )
PILFIL	16	pilot detector filter input
IFGND	17	ground of IF, detector and MPX stage
FMDEM	18	ceramic discriminator input
AFC <sub>(n)</sub>	19	AFC negative output
AFC <sub>(p)</sub>	20	AFC positive output
FSI	21	field-strength indicator
V <sub>CC2</sub>	22	supply voltage for tuning
V <sub>DDD</sub>	23	digital supply voltage
MO/ST	24	mono/stereo and tuning indication output
XTAL	25	crystal input
DGND	26	digital ground
BUS-CLOCK	27	bus-clock input
DATA	28	bus data input/output
WRITE-ENABLE	29	bus write-enable input
P0	30	programmable output port (P0)
P1	31	programmable output port (P1)
AFC	32	450 kHz LC-circuit
FM-IFI2	33	FMIF input 2 (input impedance typical 330 $\Omega$ )
V <sub>STAB(B)</sub>	34	internal stabilized supply voltage (B)
FM-IFO1	35	FMIF output 1 (output impedance typical 330 $\Omega$ )
AM-IFI/O2	36	input/output to IF-Tank (IFT); output: current source
FM-IFI1	37	FMIF input 1 (input impedance typical 330 $\Omega$ )
V <sub>STAB(A)</sub>	38	internal stabilized supply voltage (A)
FM-MIXER	39	ceramic filter output (output impedance typical 330 $\Omega$ )
AM-MIXER	40	open-collector output to IFT

#### Product specification

#### Self Tuned Radio (STR)



#### TEA5757; TEA5759

#### FUNCTIONAL DESCRIPTION

The TEA5757; TEA5759 is an integrated AM/FM stereo radio circuit including digital tuning and control functions.

#### The radio

The AM circuit incorporates a double balanced mixer, a one-pin low-voltage oscillator (up to 30 MHz) and is designed for distributed selectivity.

The AM input is designed to be connected to the top of a tuned circuit. AGC controls the IF amplification and for large signals it lowers the input impedance of the AM front-end.

The first AM selectivity can be an IF-Tank (IFT) as well as an IFT combined with a ceramic filter; the second one is an IFT.

The FM circuit incorporates a tuned RF stage, a double balanced mixer, a one-pin oscillator and is designed for distributed IF ceramic filters. The FM quadrature detector uses a ceramic resonator (or LC).

The PLL stereo decoder incorporates a signal dependent stereo-blend circuit and a soft-mute circuit.

#### Tuning

The tuning concept of the Self Tuned Radio (STR) is based on FUZZY LOGIC: it mimics hand tuning (hand tuning is a combination of coarse and fine tuning to the qualitatively best frequency position). As a consequence the tuning system is very fast.

The tuning algorithm, which is controlled by the sequential circuit (see Fig.1), is completely integrated; so there are only a few external components needed.

The bus and the microcontroller can be kept very simple. The bus only consists of three wires (BUS-CLOCK, DATA and WRITE-ENABLE). The microcontroller must basically give two instructions:

- Preset operation
- Search operation.

#### PRESET OPERATION

In preset mode, the microcontroller has to load information such as frequency band, frequency and mono/stereo. This information has to be sent via the bus to the STR. The internal algorithm controls the tuning sequence as follows:

1. The information is loaded into the shift register, the last-station memory and the counter.

- 2. The Automatic Frequency Control (AFC) is switched off.
- 3. The counter starts counting the frequency and the tuning voltage is varied until the desired frequency roughly equals the real frequency.
- 4. The AFC is then switched on and the counter is switched off.
- 5. The real frequency is more precisely tuned to the desired frequency.

After the AFC has tuned the real frequency to the desired frequency an in-lock signal can be generated. In order to get a reliable in-lock signal, there are two parameters measured: the field strength and the S-curve. The field strength indicates the strength of the station and by looking at the S-curve the system can distinguish false in-locks from real in-locks (false in-locks occur on the wrong slope of the S-curve).

In the event of fading or pulling the in-lock signal becomes logic 0 and the synthesizer will be switched on again and the algorithm will be repeated.

#### SEARCH OPERATION

During a search operation, the only action the microcontroller has to take is: sending the desired band plus the direction and the search sensitivity level to the STR. The search operation is performed by the charge pump until an in-lock signal is generated (combination of measuring the field strength and the S-curve). The AFC then fine tunes to the station. The frequency belonging to the found station will be counted by the counter and written into the last-station memory and the shift register of the counter. At this time the frequency is available in the shift register and can be read by the microcontroller. The microcontroller decides whether the frequency is within the desired frequency band. If so, this frequency can be stored under a preset and if not, a new search action should be started.

To ensure that the search function operates correctly under all conditions the following search sequence must be applied:

- Store the current frequency in the memory
- Issue the search command
- Wait for data valid and read the new frequency
- If the new frequency is the same as the stored frequency, issue a pre-set step (e.g. 50 kHz) and start the search sequence again.

#### TEA5757; TEA5759

#### Description of the bus

The TEA5757; TEA5759 radio has a bus which consists of three wires, as shown in Table 1.

#### Table 1 Bus signals

SIGNAL	GNAL DESCRIPTION	
BUS-CLOCK software driven clock input		27
DATA	data input/output	28
WRITE-ENABLE	write/read input	29

These three signals, together with the mono/stereo pin (MO/ST; pin 24), communicate with the microcontroller. The mono/stereo indicator has two functions, which are controlled by the BUS-CLOCK, as shown in Table 2.

 Table 3
 Explanation of the shift register bits

#### Table 2 Bus-clock functions

BUS-CLOCK	MO/ST (PIN 24)	RESULT
LOW	LOW	stereo
LOW	HIGH	mono
HIGH	LOW	tuned
HIGH	HIGH	not tuned

The TEA5757; TEA5759 has a 25-bit shift register; see Table 3 for an explanation of the shift register bits.

If in search mode no transmitter can be found, all frequency bits of the shift register are set to logic 0.

The bus protocol is depicted in Figs 3 and 4.

BIT DESCRIPTION		LOGIC STATE	RESULT		
S.24 (MSB)	search start/end	0	after a search when a station is found or after a preset		
		1	during the search action		
D.23	search up/down	0	indicates if the radio has to search down		
		1	indicates if the radio has to search up		
M.22	mono/stereo	0	stereo is allowed		
		1	mono is required (radio switched to forced mono)		
B0.21	band	see Table 4	selects FM/MW/LW/SW band		
B1.20					
P0.19	port	note 1	user programmable bits which e.g. can be used as band		
P1.18			switch driver		
S0.17	search-level of station	see Table 5	determines the locking field strength during an		
S1.16			automatic search, automatic store or manual search		
15	dummy	_	buffer		
F.14 to F.0 (LSB)	frequency	-	determine the tuning frequency of the radio; see Table 6 for the bit values		

#### Note

1. The output pins 30 and 31 can drive currents up to 5 mA; bits P0.19 and P1.18 control the output voltage of the control pins P0 (pin 30) and P1 (pin 31):

- a) Bit P0.19 LOW sets P0 (pin 30) to LOW.
- b) Bit P0.19 HIGH sets P0 (pin 30) to HIGH.
- c) Bit P1.18 LOW sets P1 (pin 31) to LOW.
- d) Bit P1.18 HIGH sets P1 (pin 31) to HIGH.

#### TEA5757; TEA5759

Table 4	Truth table for bits B0.21 and B1.20
---------	--------------------------------------

B0.21	B1.20	BAND SELECT
0	0	FM
0	1	MW
1	0	LW
1	1	SW

#### Table 5 Truth table for bits S1.16 and S0.17

S1.16		SIGNAL RECEPTION		
	S0.17	FM (μV)	ΑΜ (μV)	
0	0	>5	>28	
0	1	>10	>40	
1	0	>30	>63	
1	1	>150	>1000	

BIT	BIT VALUE	FM VALUE <sup>(1)</sup> (kHz)	AM VALUE <sup>(2)</sup> (kHz)			
F.14	2 <sup>14</sup>	_	16384			
F.13	2 <sup>13</sup>	102400	8192			
F.12	2 <sup>12</sup>	51200	4096			
F.11	2 <sup>11</sup>	25600	2048			
F.10	2 <sup>10</sup>	12800	1024			
F.9	2 <sup>9</sup>	6400	512			
F.8	2 <sup>8</sup>	3200	256			
F.7	27	1600	128			
F.6	26	800	64			
F.5	2 <sup>5</sup>	400	32			
F.4	24	200	16			
F.3	2 <sup>3</sup>	100	8			
F.2	2 <sup>2</sup>	50	4			
F.1	2 <sup>1</sup>	25	2			
F.0	20	12.5	1			

#### Table 6Values for bits F.14 to F.0

#### Notes

- 1. FM value of the affected oscillators:
  - a) FM VALUE = FMRF + FMIF (for TEA5757).
  - b) FM VALUE = FMRF FMIF (for TEA5759).
- 2. AM value of the affected oscillators: AM VALUE = AMRF + AMIF.

#### TEA5757; TEA5759

#### READING DATA

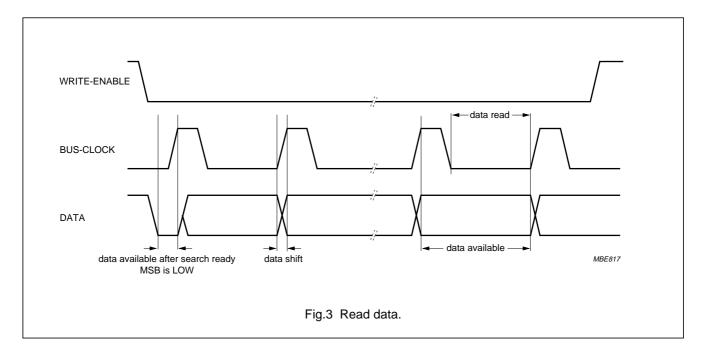
While WRITE-ENABLE is LOW data can be read by the microcontroller. At a rising edge of the BUS-CLOCK, data is shifted out of the register. This data is available from the point where the BUS-CLOCK is HIGH until the next rising edge of the BUS-CLOCK occurs (see Fig.3).

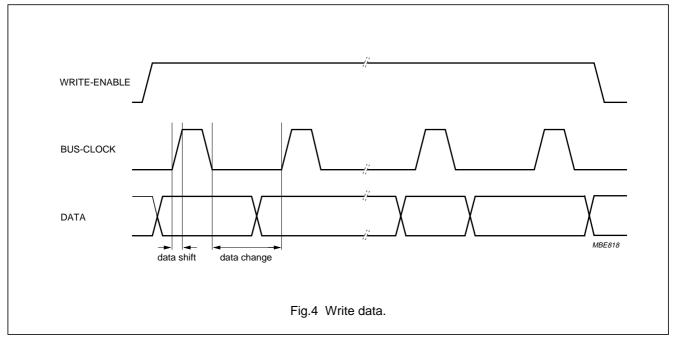
To read the entire shift register 24 clock pulses are necessary.

#### WRITING DATA

While WRITE-ENABLE is HIGH the microcontroller can transmit data to the TEA5757; TEA5759 (hard mute is active). At a rising edge of the BUS-CLOCK, the register shifts and accepts one bit into LSB. At clock LOW the microcontroller writes data (see Fig.4).

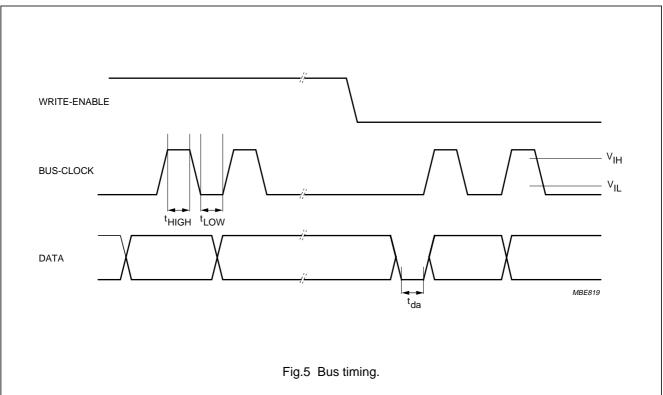
To write the entire shift register 25 clock pulses are necessary.





# TEA5757; TEA5759

**BUS TIMING** 



#### Table 7 Digital inputs

SYMBOL	PARAMETER	MIN.	MAX.	UNIT
Digital inputs				
V <sub>IH</sub>	HIGH-level input voltage	1.4	-	V
V <sub>IL</sub>	LOW-level input voltage	-	0.6	V
Timing			·	
f <sub>clk</sub>	clock input frequency	-	300	kHz
t <sub>HIGH</sub>	clock HIGH time	1.67	-	μs
t <sub>LOW</sub>	clock LOW time	1.67	-	μs
t <sub>da</sub>	shift register available after 'search ready'	-	14	μs

# TEA5757; TEA5759

#### LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
V <sub>CC1</sub>	supply voltage		0	13.2	V
P <sub>tot</sub>	total power dissipation	T <sub>amb</sub> = 70 °C	-	250	mW
T <sub>stg</sub>	storage temperature		-65	+150	°C
T <sub>amb</sub>	ambient temperature		–15	+60	°C
Tj	junction temperature		–15	+150	°C
V <sub>es</sub>	electrostatic handling voltage for all pins	note 1	_	±200	V

#### Note

1. Charge device model; equivalent to discharging a 200 pF capacitor via a 0  $\Omega$  series resistor.

#### THERMAL CHARACTERISTICS

SYMBOL	PARAMETER	CONDITIONS	VALUE	UNIT
R <sub>th(j-a)</sub>	thermal resistance from junction to ambient	in free air	65	K/W

# TEA5757; TEA5759

#### CHARACTERISTICS

 $V_{CC1}$  = 3 V;  $T_{amb}$  = 25 °C; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V <sub>CC1</sub>	supply voltage		2.5	_	12	V
V <sub>CC2</sub>	supply voltage for tuning		_	_	12	V
V <sub>DDD</sub>	supply voltage for digital part		2.5	_	12	V
V <sub>tune</sub>	tuning voltage		0.7	_	V <sub>CC2</sub> - 0.75	V
I <sub>CC2</sub>	supply current for tuning in preset mode (band-end to band-end)		-	-	800	μΑ
f <sub>BUS-CLOCK(max)</sub>	maximum BUS-CLOCK frequency		_	-	300	kHz
I <sub>CC1</sub>	current consumption during	AM mode	12	15	18	mA
	acquisition of V <sub>CC1</sub>		12.5	15.5	18.5	mA
I <sub>DD</sub>	current consumption during	AM mode	-	4.8	-	mA
acquisition of I <sub>DD</sub>		FM mode	-	5.5	-	mA
I <sub>CC1</sub>	current consumption after	AM mode	12	15	18	mA
	acquisition of V <sub>CC1</sub>	FM mode	13	16	19	mA
I <sub>DD</sub>	current consumption after	AM mode	-	3.3	-	mA
	acquisition of I <sub>DD</sub>	FM mode	-	2.7	-	mA
t <sub>search</sub>	synthesizer auto-search time for empty band	FM mode	_	-	10	S
t <sub>acq</sub>	synthesizer preset acquisition	FM	_	100	-	ms
	time between two band limits	MW	-	100	_	ms
		LW	-	200	_	ms
		SW	_	500	_	ms
f <sub>band</sub>	frequency band range of the	AM mode	0.144	_	30	MHz
	synthesizer	FM mode	50	_	150	MHz
$\Delta f_{FM}$	AFC inaccuracy of FM		_	_	1	kHz
$\Delta f_{AM}$	AFC inaccuracy of AM		_	_	100	Hz
I <sub>P0(sink)</sub>	sink current of software programmable output P0	V <sub>30</sub> = 3 V	4	6	-	mA
I <sub>P1(sink)</sub>	sink current of software programmable output P1	V <sub>31</sub> = 3 V	4	6	-	mA
I <sub>P0(source)</sub>	source current of software programmable output P0	V <sub>30</sub> = 0 V	5	9	-	mA
I <sub>P1(source)</sub>	source current of software programmable output P1	V <sub>31</sub> = 0 V	5	9	-	mA

#### TEA5757; TEA5759

#### AM CHARACTERISTICS

Input frequency  $f_i = 1$  MHz; m = 0.3;  $f_m = 1$  kHz; measured in test circuit at pin 10 (see Fig.9); S2 in position B; V<sub>i1</sub> measured at input of matching network at pin 2; matching network adjusted to maximum output voltage at low input level; V<sub>n</sub> refers to pin voltages; V<sub>i(n)</sub> refers to test circuit (see Fig.9); unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V <sub>10</sub>	AF output voltage	V <sub>i1</sub> = 5 mV	36	45	70	mV
V <sub>i1</sub>	RF sensitivity input voltage	(S+N)/N = 26 dB	40	55	70	μV
V <sub>i1</sub>	large signal voltage handling capacity	m = 0.8; THD ≤ 8%	150	300	-	mV
PSRR	power supply ripple rejection $\left(\frac{V_{10}}{\Delta V_7}\right)$	$\Delta V_7 = 100 \text{ mV} (\text{RMS});$ 100 Hz; V <sub>7</sub> = 3.0 V	-	-47	-	dB
li	input current (pin 2)	V <sub>44</sub> = 0.2 V	-	0	-	μΑ
C <sub>i</sub>	input capacitance (pin 2)	V <sub>44</sub> = 0.2 V	-	-	4	pF
G <sub>c</sub>	front-end conversion gain	V <sub>44</sub> = 0.2 V	5	10	14	dB
		V <sub>44</sub> = 0.9 V	-26	-14	0	dB
(S+N)/N	signal plus noise-to-noise ratio		-	50	-	dB
THD	total harmonic distortion	V <sub>i1</sub> = 1 mV	-	0.8	2.0	%
α <sub>450</sub>	IF suppression	V <sub>10</sub> = 30 mV	_	56	-	dB

#### **FM CHARACTERISTICS**

Input frequency  $f_i = 100 \text{ MHz}$ ;  $\Delta f = 22.5 \text{ kHz}$ ;  $f_m = 1 \text{ kHz}$ ; measured in test circuit (see Fig.9) at pin 10; S2 in position B; V<sub>n</sub> refers to pin voltages; V<sub>i(n)</sub> refers to test circuit (see Fig.9); unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V <sub>10</sub>	AF output voltage	V <sub>i5</sub> = 1 mV	40	48	57	mV
V <sub>i5</sub>	RF sensitivity input voltage	(S+N)/N = 26 dB	1	2	3.8	μV
V <sub>i5</sub>	RF limiting sensitivity	V <sub>10</sub> at –3 dB; V <sub>10</sub> is 0 dB at V <sub>i5</sub> = 1 mV	0.4	1.2	3.8	μV
V <sub>i5</sub>	large signal voltage handling capacity	THD < 5%	-	500	-	mV
PSRR	power supply ripple rejection $\left(\frac{V_{10}}{\Delta V_7}\right)$	$\Delta V_7 = 100 \text{ mV} (\text{RMS});$ 100 Hz; V <sub>7</sub> = 3.0 V	-44	-	-	dB
G <sub>c</sub>	front-end conversion gain $\left(\frac{V_{37}}{V_{15}}\right)$		12	18	22	dB
(S+N)/N	signal plus noise-to-noise ratio	V <sub>i5</sub> = 1 mV	-	62	-	dB
THD	total harmonic distortion	IF filter SFE10.7MS3A20K-A	-	0.3	0.8	%

# TEA5757; TEA5759

#### STEREO DECODER CHARACTERISTICS

 $V_{i3(L + R)} = 155 \text{ mV}$ ;  $V_{pilot} = 15.5 \text{ mV}$ ; f = 1 kHz; apply unmodulated RF signal of 100 mV to front-end to set radio to maximum channel separation; soft mute off (S4 in position A); unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V <sub>14/15</sub>	AF output voltage		-	160	-	mV
V <sub>pilot(s)</sub>	switch to stereo		-	8	12	mV
V <sub>pilot(m)</sub>	switch to mono		2	5	_	mV
V <sub>AF-L</sub> /V <sub>i3</sub>	MPX voltage gain		-1.5	-	+1.5	dB
(S+N)/N	signal plus noise-to-noise ratio	V <sub>pilot</sub> = 15.5 mV (stereo)	-	74	-	dB
THD	total harmonic distortion		-	0.5	1.0	%
$\alpha_{cs}$	channel separation		26	30	-	dB
α <sub>19</sub>	carrier and harmonic suppression	19 kHz (200 mV) = 0 dB	27	32	-	dB
α <sub>38</sub>		38 kHz	16	21	-	dB
α	stereo-blend	V <sub>i5</sub> = 200 μV	22	30	-	dB
		$V_{i5} = 20 \ \mu V$	-	1	2	dB
mute(s)	soft mute depth	$V_{i5} = 3 \mu V; V_{14} = V_{15}$	-1	0	-	dB
		$V_{i5} = 1 \ \mu V; \ V_{14} = V_{15}$	-	-6	-10	dB

#### TUNING CHARACTERISTICS

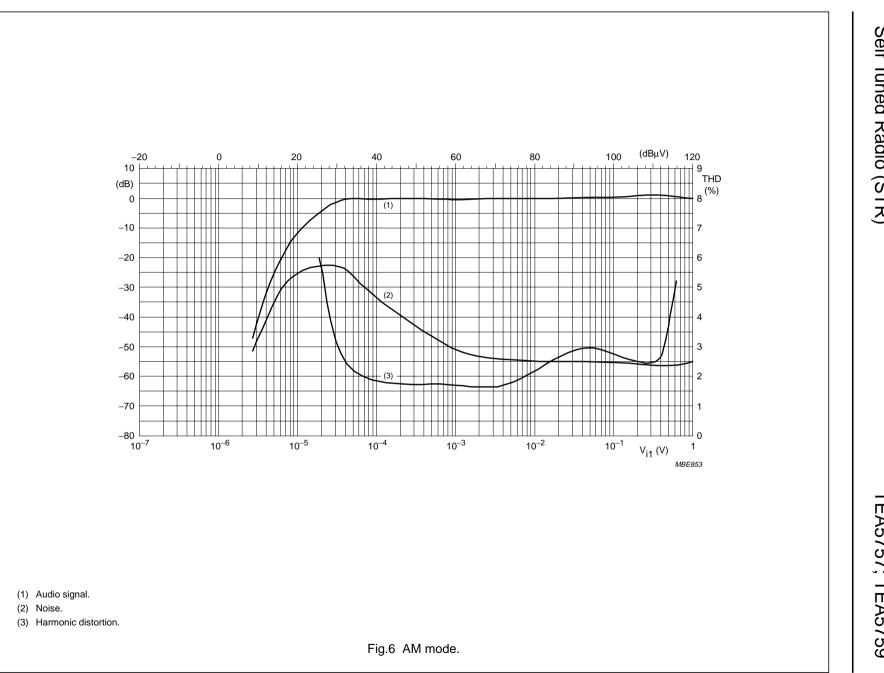
SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V <sub>FM</sub>	FM voltage levels	$\alpha_{-3 \text{ dB}}$ -point at V <sub>i5</sub> = 2 $\mu$ V				
	high (auto-store/search)	S0 = 1; S1 = 1	60	150	500	μV
	medium (auto-store/search)	S0 = 0; S1 = 1	10	30	55	μV
	low (auto-store/search)	S0 = 1; S1 = 0	4	10	20	μV
	nominal (preset mode/tuning indication)	S0 = 0; S1 = 0	3	5	9	μV
V <sub>AM</sub>	AM voltage levels	$\alpha_{-3 \text{ dB}}$ -point at V <sub>i5</sub> = 2 $\mu$ V				
	high (auto-store/search)	S0 = 1; S1 = 1	400	1000	2500	μV
	medium (auto-store/search)	S0 = 0; S1 = 1	50	63	80	μV
	low (auto-store/search)	S0 = 1; S1 = 0	32	40	50	μV
	nominal (preset mode/tuning indication)	S0 = 0; S1 = 0	25	28	40	μV
V <sub>AFC(off)</sub>	AFC voltage off mode	$\alpha_{-3 \text{ dB}}$ -point at V <sub>i5</sub> = 2 $\mu$ V				
		FM mode	_	3	-	μV
		AM mode	_	25	-	μV
mute(h)	hard mute depth	WRITE-ENABLE = HIGH	_	60	_	dB

Philips Semiconductors

Product specification

# Self Tuned Radio (STR)

TEA5757; TEA5759



1999 Aug 26

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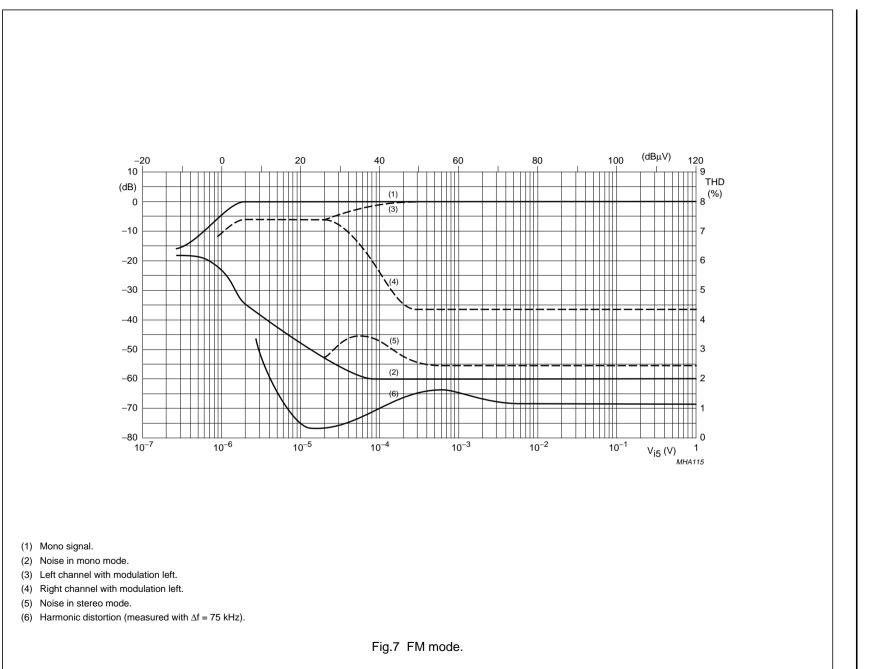
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Product specification

# Self Tuned Radio (STR)

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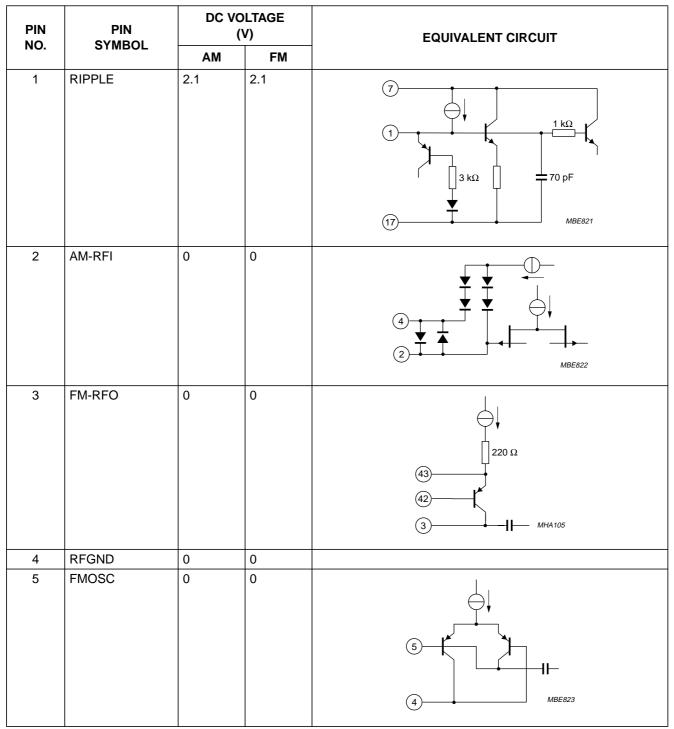
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# TEA5757; TEA5759

#### INTERNAL CIRCUITRY

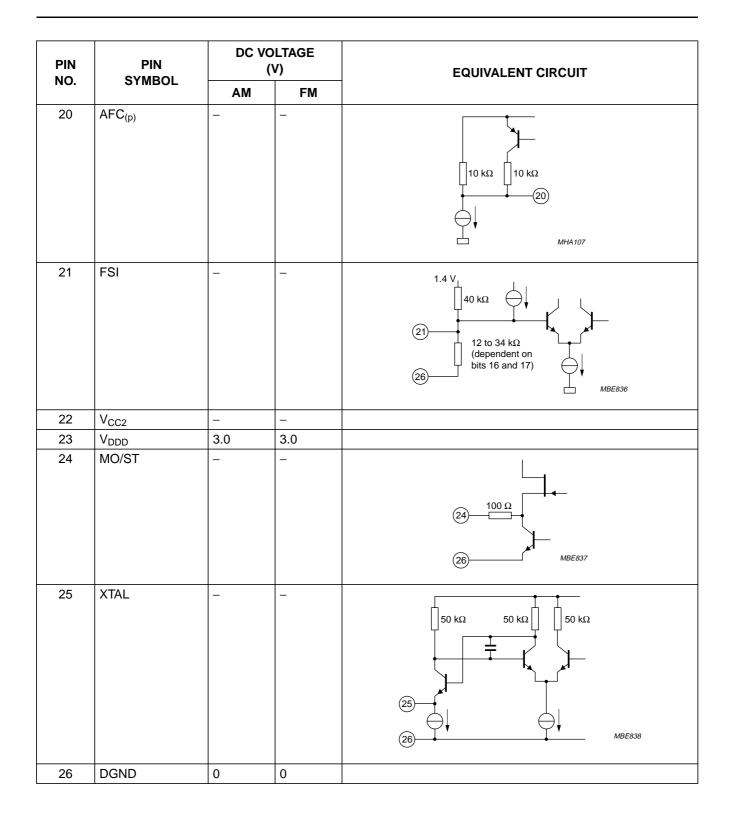
Table 8 Equivalent pin circuits and pin voltages

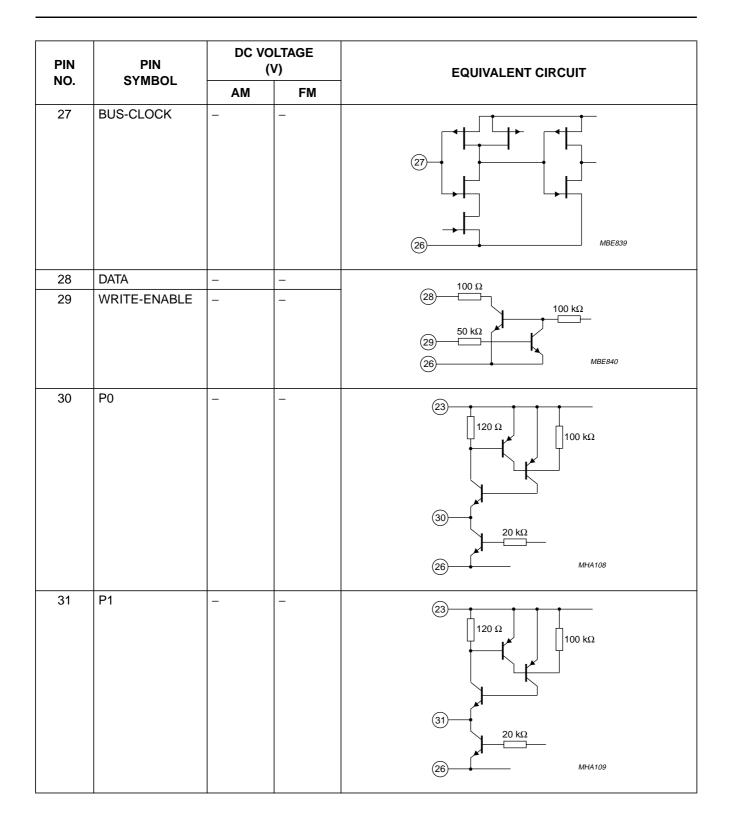


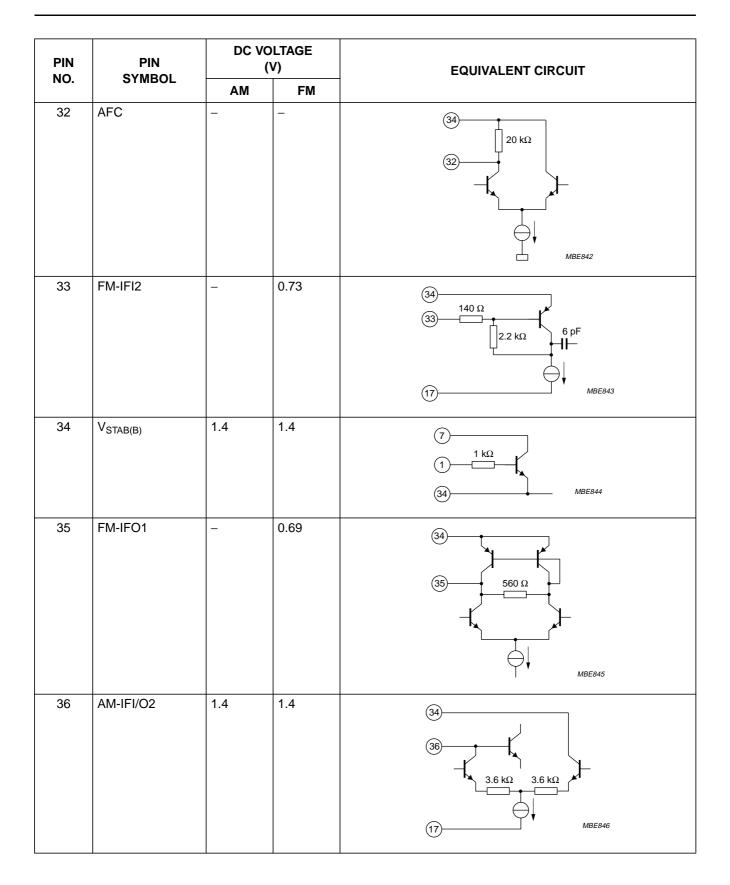
PIN	PIN PIN NO. SYMBOL				EQUIVALENT CIRCUIT
NO.	STMBOL	AM	FM		
6	AMOSC	0	0		
7	V <sub>CC1</sub>	3.0	3.0		
8	TUNE	_	-	22 (2) (3) (26) (26) (26) (27) (3) (4) (4) (4) (4) (4) (4) (4) (4	
9	VCO	1.3	0.95	9 1 KΩ 1 KΩ 1 KΩ 1 KΩ 10 KΩ MBE626	
10	AFO	0.6	0.7	10 10 5 KΩ MBE827	

PIN NO.	PIN SYMBOL		OLTAGE (V)	EQUIVALENT CIRCUIT
NO.	STMBOL	AM	FM	
11	MPXI	1.23	1.23	150 kΩ 150 kΩ 150 kΩ 150 kΩ 150 kΩ 150 kΩ 150 kΩ 150 kΩ 150 kΩ
12	LFI	0.1	0.8	12 13 KΩ 17 MBE829
13	MUTE	0.7	0.7	7 κΩ         50 κΩ           13         6000000000000000000000000000000000000
14	AFLO	0.65	0.65	(14) 5 kΩ 17 MBE831

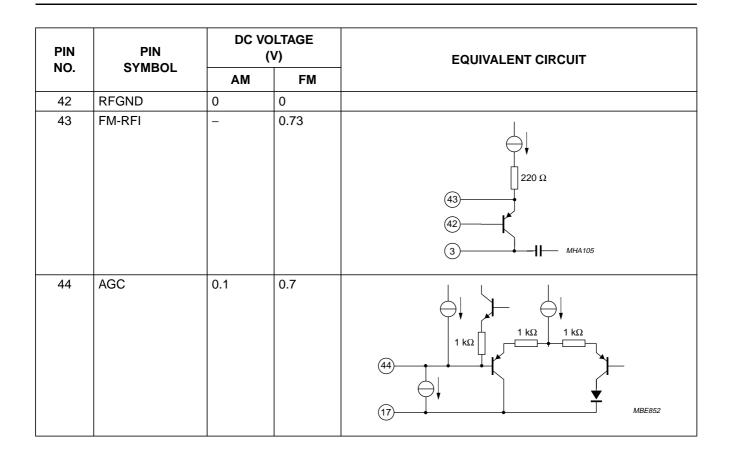
PIN	PIN		OLTAGE (V)	EQUIVALENT CIRCUIT
NO.	SYMBOL	AM	FM	
15	AFRO	0.65	0.65	(15 5 KΩ (17) MBE832
16	PILFIL	0.95	0.95	16 $10 \text{ k}\Omega$ $10 \text{ k}\Omega$ $10 \text{ k}\Omega$ $10 \text{ k}\Omega$ $10 \text{ k}\Omega$ $10 \text{ k}\Omega$ $10 \text{ k}\Omega$
17	IFGND	0	0	
18	FMDEM	_	1.0	180 Ω 910 Ω (17) MBE834
19	AFC <sub>(n)</sub>	_	-	10 kΩ 10 kΩ (19) (19) (19) (19) (19) (19) (19) (19) (19) (19) (10)





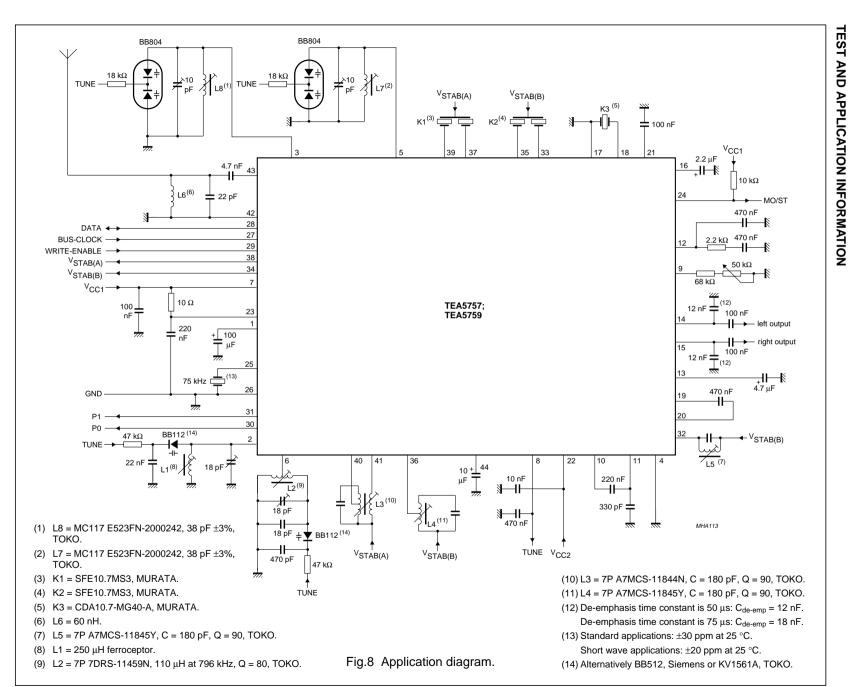


PIN NO.	PIN SYMBOL		OLTAGE (V)	EQUIVALENT CIRCUIT
NO.	STMBOL	AM	FM	
37	FM-IFI1	_	0.73	(17) МВЕВ47
38	V <sub>STAB(A)</sub>	1.4	1.4	7 1 kΩ 1 38 MBE848
39	FM-MIXER	_	1.0	39 680 Ω MHA110
40	AM-MIXER	1.4	1.4	(40) (38) (38) (MBE850)
41	AM-IFI1	1.4	1.4	(41) (41) (41) (17) (17) (17) (38) (7.5 kΩ (7.5 kΩ) (7.5 kΩ (7.5





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# Philips Semiconductors

Self Tuned Radio (STR)

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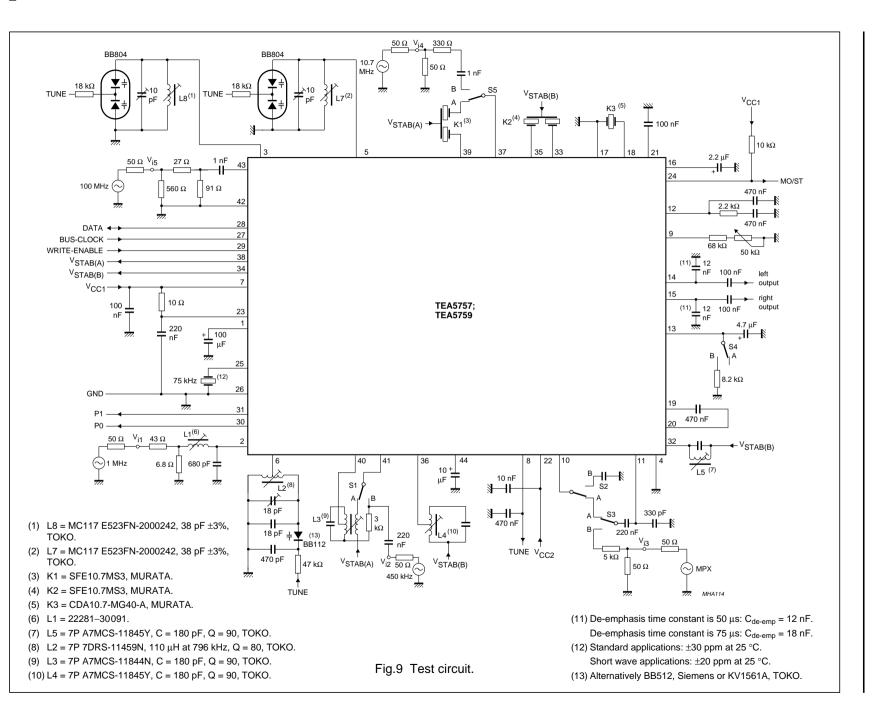
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Self

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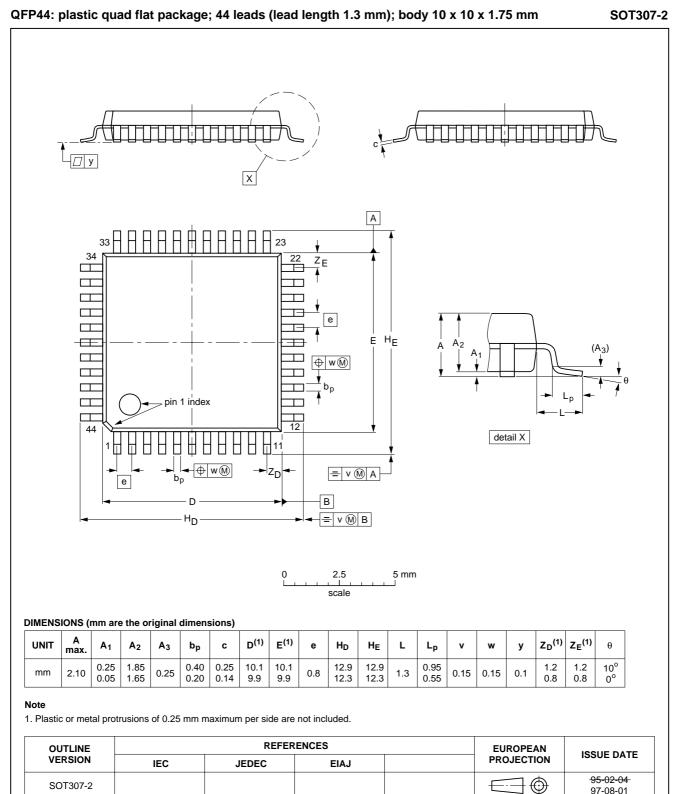
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#### TEA5757; TEA5759

#### PACKAGE OUTLINE



#### TEA5757; TEA5759

#### SOLDERING

#### Introduction to soldering surface mount packages

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our *"Data Handbook IC26; Integrated Circuit Packages"* (document order number 9398 652 90011).

There is no soldering method that is ideal for all surface mount IC packages. Wave soldering is not always suitable for surface mount ICs, or for printed-circuit boards with high population densities. In these situations reflow soldering is often used.

#### **Reflow soldering**

Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement.

Several methods exist for reflowing; for example, infrared/convection heating in a conveyor type oven. Throughput times (preheating, soldering and cooling) vary between 100 and 200 seconds depending on heating method.

Typical reflow peak temperatures range from 215 to 250 °C. The top-surface temperature of the packages should preferable be kept below 230 °C.

#### Wave soldering

Conventional single wave soldering is not recommended for surface mount devices (SMDs) or printed-circuit boards with a high component density, as solder bridging and non-wetting can present major problems.

To overcome these problems the double-wave soldering method was specifically developed.

If wave soldering is used the following conditions must be observed for optimal results:

- Use a double-wave soldering method comprising a turbulent wave with high upward pressure followed by a smooth laminar wave.
- For packages with leads on two sides and a pitch (e):
  - larger than or equal to 1.27 mm, the footprint longitudinal axis is preferred to be parallel to the transport direction of the printed-circuit board;
  - smaller than 1.27 mm, the footprint longitudinal axis must be parallel to the transport direction of the printed-circuit board.

The footprint must incorporate solder thieves at the downstream end.

• For packages with leads on four sides, the footprint must be placed at a 45° angle to the transport direction of the printed-circuit board. The footprint must incorporate solder thieves downstream and at the side corners.

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Typical dwell time is 4 seconds at 250 °C. A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.

#### Manual soldering

Fix the component by first soldering two diagonally-opposite end leads. Use a low voltage (24 V or less) soldering iron applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to 300 °C.

When using a dedicated tool, all other leads can be soldered in one operation within 2 to 5 seconds between 270 and 320  $^\circ\text{C}.$ 

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#### Suitability of surface mount IC packages for wave and reflow soldering methods

PACKAGE	SOLDERING METHOD			
FACKAGE	WAVE	REFLOW <sup>(1)</sup>		
BGA, SQFP	not suitable	suitable		
HLQFP, HSQFP, HSOP, HTSSOP, SMS	not suitable <sup>(2)</sup>	suitable		
PLCC <sup>(3)</sup> , SO, SOJ	suitable	suitable		
LQFP, QFP, TQFP	not recommended <sup>(3)(4)</sup>	suitable		
SSOP, TSSOP, VSO	not recommended <sup>(5)</sup>	suitable		

#### Notes

- 1. All surface mount (SMD) packages are moisture sensitive. Depending upon the moisture content, the maximum temperature (with respect to time) and body size of the package, there is a risk that internal or external package cracks may occur due to vaporization of the moisture in them (the so called popcorn effect). For details, refer to the Drypack information in the "Data Handbook IC26; Integrated Circuit Packages; Section: Packing Methods".
- 2. These packages are not suitable for wave soldering as a solder joint between the printed-circuit board and heatsink (at bottom version) can not be achieved, and as solder may stick to the heatsink (on top version).
- 3. If wave soldering is considered, then the package must be placed at a 45° angle to the solder wave direction. The package footprint must incorporate solder thieves downstream and at the side corners.
- 4. Wave soldering is only suitable for LQFP, TQFP and QFP packages with a pitch (e) equal to or larger than 0.8 mm; it is definitely not suitable for packages with a pitch (e) equal to or smaller than 0.65 mm.
- 5. Wave soldering is only suitable for SSOP and TSSOP packages with a pitch (e) equal to or larger than 0.65 mm; it is definitely not suitable for packages with a pitch (e) equal to or smaller than 0.5 mm.

#### DEFINITIONS

Data sheet status	
Objective specification	This data sheet contains target or goal specifications for product development.
Preliminary specification	This data sheet contains preliminary data; supplementary data may be published later.
Product specification	This data sheet contains final product specifications.
Limiting values	

Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability.

#### Application information

Where application information is given, it is advisory and does not form part of the specification.

#### LIFE SUPPORT APPLICATIONS

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