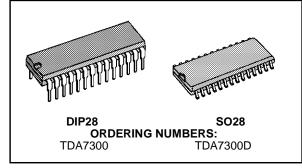


DIGITAL CONTROLLED STEREO AUDIO PROCESSOR

- SINGLE SUPPLY OPERATION
- FOUR STEREO INPUT SOURCE SELECTION
- MONO INPUT
- TREBLE, BASS, VOLUME, AND BALANCE CONTROL
- FOUR INDEPENDENT SPEAKER CONTROL (FRONT/REAR)
- SINGLE SUPPLY OPERATION
- ALL FUNCTIONS PROGRAMMABLE VIA SE-RIAL BUS
- VERY LOW NOISE AND VERY LOW DISTORTION
- POP FREE SWITCHING



sor for high quality audio applications in car radio and Hi-Fi systems.

Control is accomplished by serial bus microprocessor interface.

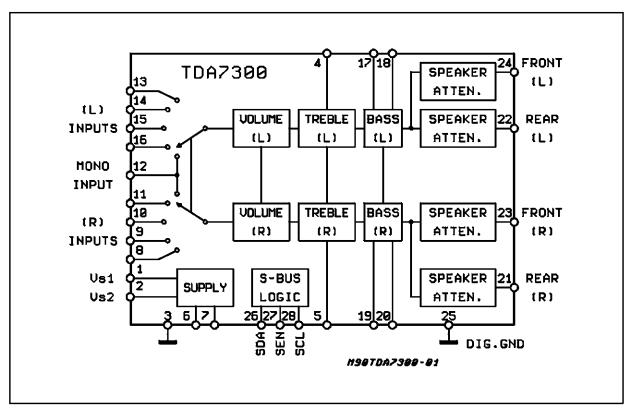
The AC signal setting is obtained by resistor networks and analog switches combined with operational amplifiers.

The results are: low noise, low distortion and high dynamic range.

DESCRIPTION

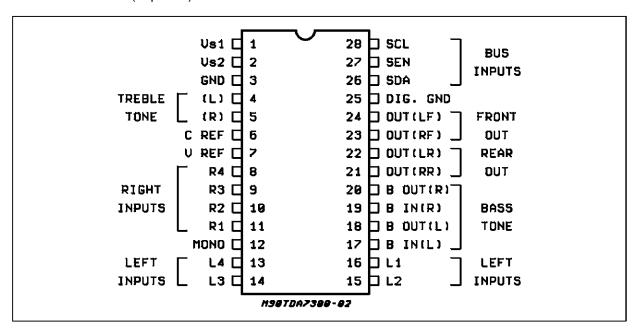
The TDA7300 is a volume, tone (bass and treble), balance (left/right) and fader (front/rear) proces-

BLOCK DIAGRAM



May 1991 1/16

PIN CONNECTION (Top view)



ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
Vs	Supply Voltage (V _{S1})	18	V
T _{amb}	Operating Ambient Temperature Range	-40 to +85	°C
T _{stg}	Storage Temperature	-40 to 150	°C

THERMAL DATA

Symbol	Description	SO28	DIP28	Unit
R _{th j-pins}	Thermal Resistance Junction-pins Max	85	65	°C/W

ELECTRICAL CHARACTERISTICS (T_{amb} = 25°C, V_{S1} = 12V or V_{S2} = 8.5V , R_L = 10k Ω and R_g = 600 Ω , f = 1KHz unless otherwise specified)

Symbol	Parameter	Test Condition	Min.	Тур.	Max.	Unit
SUPPLY (1))					
V _{S1}	Supply Voltage VS1		10	12	16	V
V _{S2}	Supply Voltage VS2		6	8.5	10	V
I _{S2}	Supply Current		15	30	40	mA
V_{ref}	Reference Voltage (pin 7)		3.5	4.3	5	V
SVR	Ripple Rejection at V _{S1}	f = 300Hz to 10KHz	80	97		dB
SVR	Ripple Rejection at V _{S2}	f = 300Hz to 10KHz	50	58		dB
INPUT SEL	ECTORS					
Ri	Input Resistance		30	45		ΚΩ
V _{IN max}	Max. Input Signal	GV = 0dB d = 0.3%	1.5	2.2		Vrms
INs	Input Separation	f = 1KHz (2)	90	100		dB
		f = 10KHz (2)	70	80		dB
V _{i (DC)}	Input DC Voltage		3.5	4.3	5	V

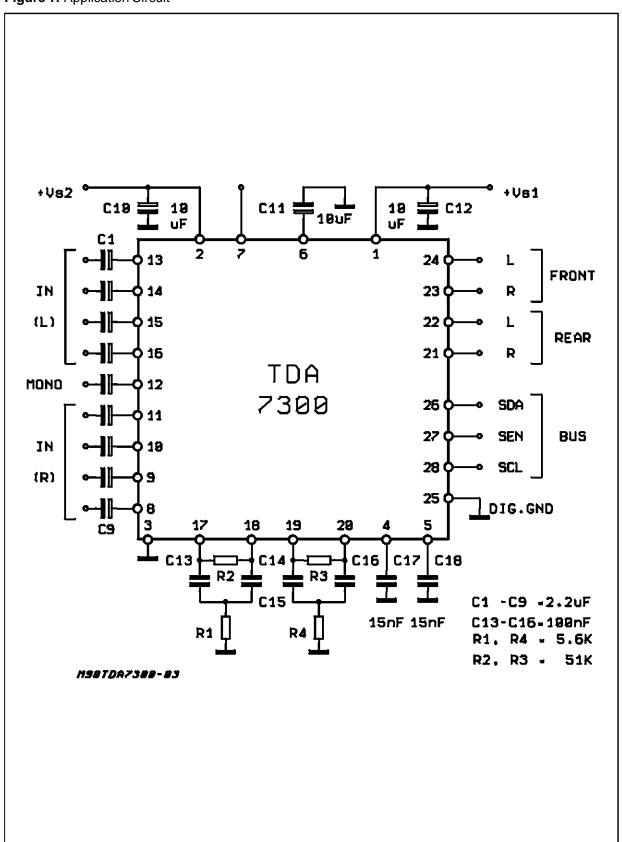
ELECTRICAL CHARACTERISTICS (continued)

Symbol	Parameter	Test Condition	Min.	Тур.	Max.	Unit
VOLUME C	ONTROLS					
	Control Range			78		dB
G _{max}	Max Gain		8	10	12	dB
	Max Attenuation		64	68		dB
	Step Resolution	G _V = -50 to 10dB		2	3	dB
	Attenuator Set Error				2	dB
	Tracking Error				2	dB
SPEAKER	ATTENUATORS					
	Control Range		35	38	41	dB
	Step Resolution			2	3	dB
	Attenuator Set Error				2	dB
	Tracking Error				2	dB
BASS AND	TREBLE CONTROL (3)					
	Control Range			±15		dB
	Step Resolution			2.5	3.5	dB
AUDIO OU	TPUT					
Vo	Max. Output Voltage	d = 0.3%	1.5	2.2		Vrms
R _L	Output Load Resistance		2			ΚΩ
CL	Output Load Capacitance				1	nF
Ro	Output Resistance			70	150	Ω
V _O (DC)	DC Voltage Level		3	3.8	4.5	V
GENERAL						
e _{NO}	Output Noise	BW = 22Hz to 22KHz, $G_v = 0dB$		6	15	μV
		Curve A $G_V = 0dB$		4		μν
S/N	Signal to Noise Ratio	All gain = 0dB V_0 = 1Vrms BW = 22Hz to 22KHz		105		dB
d	Distortion	$f = 1KHz; V_0 = 1V; G_v = 0$		0.01	0.1	%
	Frequency Response (-1dB)	$G_{v} = 0$ High Low	20		20	KHz Hz
S _C	Channnel Separation left/right	f = 1KHz f = 10KHz	90 70	100 80		dB dB
BUS INPUT	rs					
V _{IL}	Input LOW Voltage				0.8	V
V _{IH}	Input HIGH Voltage		2.4			V
Vo	Output Voltage SDA Acknowledge	I = 1.6mA			0.4	V
	Digital Input Current		-5		+5	μΑ



 ⁽¹⁾ The circuit can be supplied either at V_{S1} or without the use of the internal voltage regulator at V_{S2}. The circuit also operates at a supply voltage V_{S1} lower than 10V. In this case the ripple rejection of V_{S2} is valid, because the voltage regulator saturates to a saturation voltage of about 0.8V.
 (2) The selected input is grounded thru the 2.2μF capacitor.
 (3) Bass and Treble response see attached diagram. The center frequency and quality of the resonance behaviour can be choosen by the external circuitry. A standard first order bass response can be realized by a standard feedback network.

Figure 1: Application Circuit



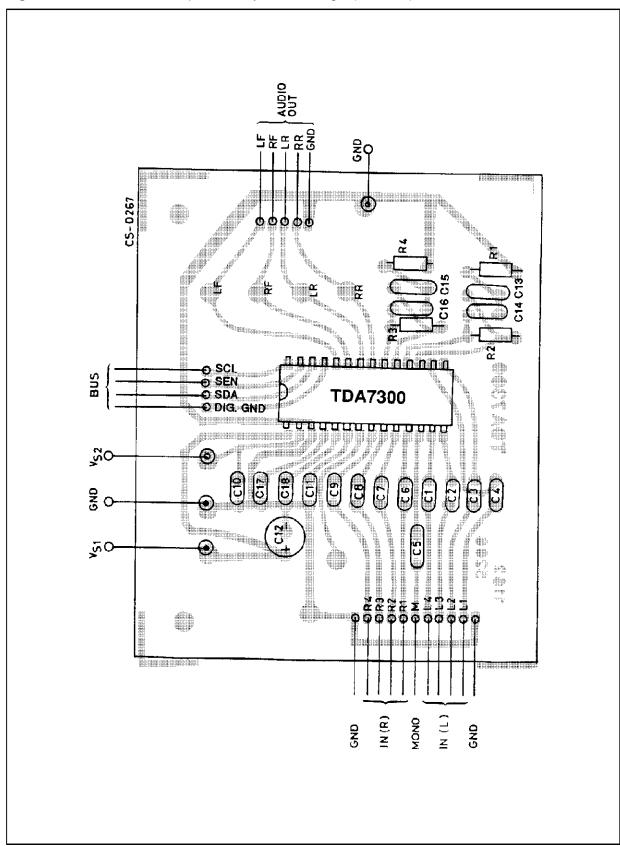


Figure 2: P.C. Board and Components Layout of the Fig.1 (1:1 scale)

Figure 3: Total Output Noise vs. Volume Setting

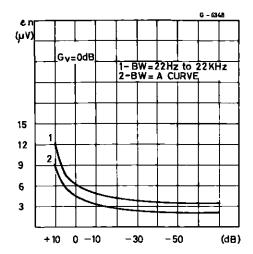


Figure 5: Distortion + Noise vs. Frequency

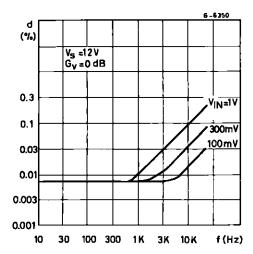


Figure 7: Distortion vs. Load Resistance

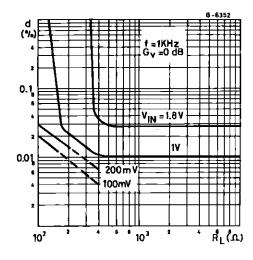


Figure 4: Signal to Noise Ratio vs. Volume Setting

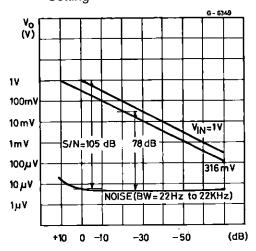


Figure 6: Distortion vs. Output Voltage

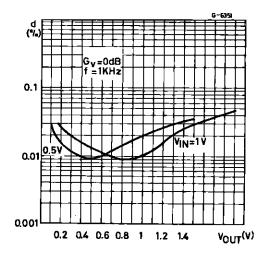


Figure 8: Channel Separation (L1 - R1) vs. Frequency

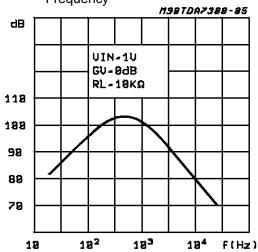


Figure 9: Input Separation (L1 - L2) vs. (V_{S1}) Frequency

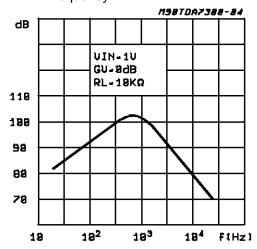


Figure 11: Supply Voltage Rejection (V_{S2}) vs. Frequency

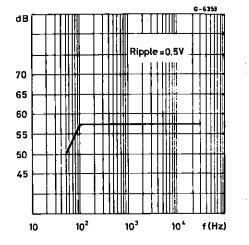


Figure 13: Supply Voltage Rejection vs. Vs2

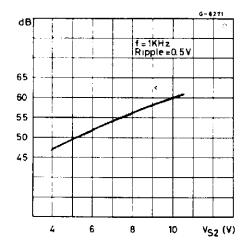


Figure 10: Supply Voltage Rejection (V_{S1}) vs. Frequency

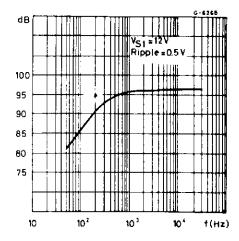


Figure 12: Supply Voltage Rejection vs. V_{S1}

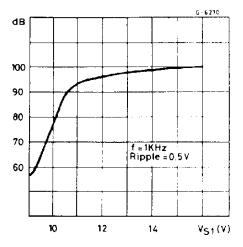
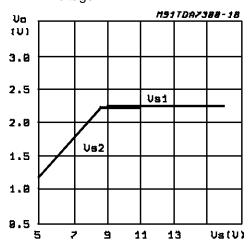


Figure 14: Clipping Level (Vrms) vs. Supply Voltage



APPLICATION INFORMATION

Volume Control Concept

Traditional electronic volume control circuits use a multiplier technique with all the disadvantages of high noise and distortion.

The used concept, as shown in Fig. 15 with digital switched resistor dividers, provides extremely low noise and distortion. The multiplexing of the resistive dividers is realized with a multiple-input operational amplifier.

Bass and Treble Control

The principle operation of the bass control is shown in Fig. 16. The external filter together with the internal buffer allows a flexible filter design according to the different requirements in car radios. The function of the treble is similar to the bass. A typical curve is shown in Fig.19.

Outputs

A special class-A output amplifier with a modulated sink current provides low distortion and ground compatibility with low current consumption.

Figure 15: Volume Control

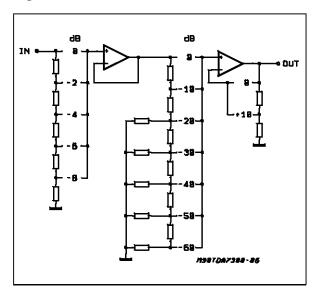


Figure 16: Bass Control

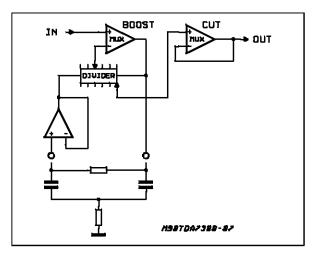


Figure 17: Quiescent Current vs. Supply Voltage

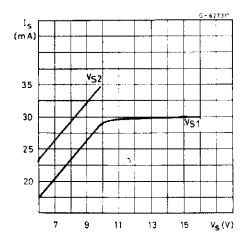
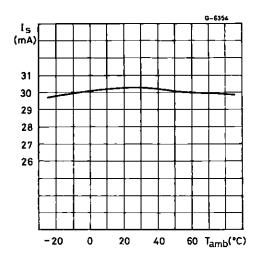


Figure 18: Quiescent Current vs. Temperature



APPLICATION INFORMATION (continued)

Figure 19: Typical Tone Response

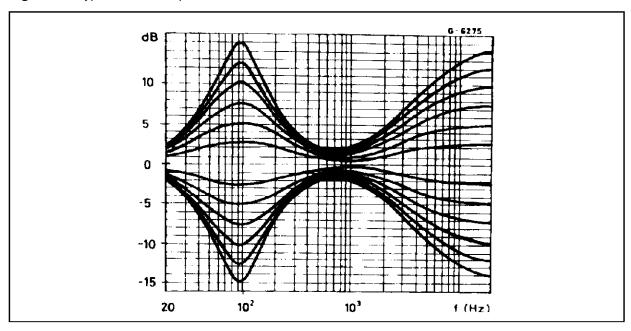
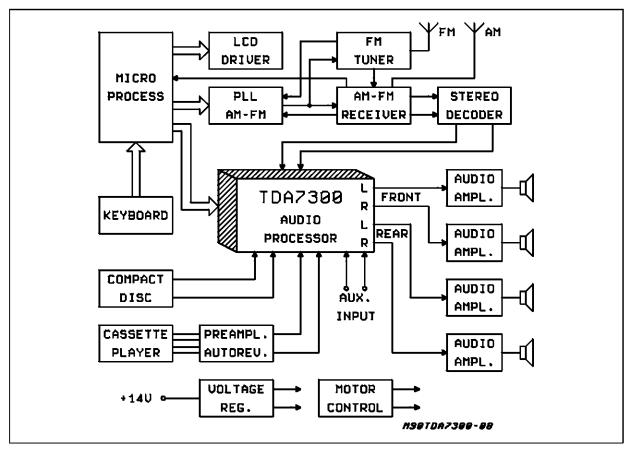


Figure 20: Complete Car-Radio System using Digital Controlled Audio Processor



APPLICATION INFORMATION (continued)

SERIAL BUS INTERFACE

S-BUS Interface and I²CBUS Compatibility

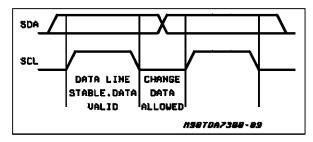
Data transmission from microprocessor to the TDA7300 and viceversa takes place thru the 3-wire S-BUS interface, consisting of the three lines SDA, SCL, SEN. If SDA and SEN inputs are short-circuited together, then the TDA7300 appears as a standard I²CBUS slave.

According to I²CBUS specification the S-BUS lines are connected to a positive supply voltage via pull-up resistors.

Data Validity

As shown in fig. 21, the data on the SDA line must be stable during the high period of the clock. The HIGH and LOW state of the data line can only change when the clock signal on the SCL line is LOW.

Figure 21: Data Validity on the I²CBUS



Start and Stop Conditions

I²CBUS:

as shown in fig.22 a start condition is a HIGH to

Figure 22: Timing Diagram of S-BUS and I²CBUS

LOW transition of the SDA line while SCL is HIGH. The stop condition is a LOW to HIGH transition of the SDA line while SCL is HIGH.

S-bus:

the start/stop conditions (points 1 and 6) are detected exclusively by a transition of the SEN line (1 \rightarrow 0 / 0 \rightarrow 1) while the SCL line is at the HIGH level

The SDA line is only allowed to change during the time the SCL line is low (points 2, 3, 4, 5). After the start information (point 1) the SEN line returns to the HIGH level and remains unchanged for all the time the transmission is performed.

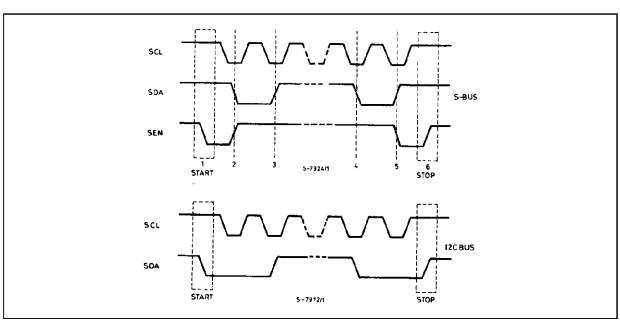
Byte Format

Every byte transferred on the SDA line must contain 8 bits. Each byte must be followed by an acknowledge bit. The MSB is transferred first.

Acknowledge

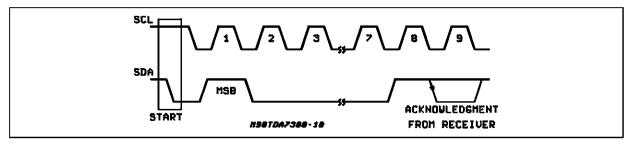
The master (μP) puts a resistive HIGH level on the SDA line during the acknowledge clock pulse (see fig. 23). The peripheral (audioprocessor) that acknowledges has to pull-down (LOW) the SDA line during the acknowledge clock pulse, so that the SDA line is stable LOW during this clock pulse.

The audioprocessor which has been addressed has to generate an acknowledge after the reception of each byte, otherwise the SDA line remains at the HIGH level during the ninth clock pulse time. In this case the master transmitter can generate the STOP information in order to abort the transfer.



APPLICATION INFORMATION (continued)

Figure 23: Acknowledge on the I²CBUS



Transmission without Acknowledge

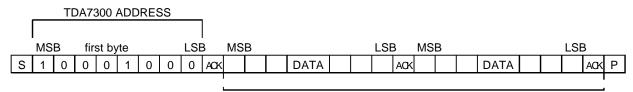
Avoiding to detect the acknowledge of the audio-processor, the μP can use a simplier transmission: simply it waits one clock without checking the slave acknowledging, and sends the new data.

This approach of course is less protected from misworking and decreases the noise immunity.

Interface Protocol

The interface protocol comprises:

- A start condition (S)
- A chip address byte, containing the TDA7300 address (the 8th bit of the byte must be 0). The TDA7300 must always acknowledge at the end of each transmitted byte.
- A sequence of data (N-bytes + acknowledge)
- A stop condition (P)



Data Transferred (N-bytes + Acknowledge)

ACK = Acknowledge

S = Start

P = Stop

MAX CLOCK SPEED 100kbits/s

SOFTWARE SPECIFICATION

Chip address (TDA7300 address)

1 0 0 0 1 0 0 0 MSB LSB

DATA BYTES

MSE	3					L	SB	Function
0	0	B2	В1	B0	A2	A1	Α0	Volume Control
1	1	0	B1	B0	A2	Α1	A0	Speaker ATT LR
1	1	1	B1	B0	A2	Α1	A0	Speaker ATT RR
1	0	0	B1	B0	A2	Α1	A0	Speaker ATT LF
1	0	1	B1	B0	A2	Α1	A0	Speaker ATT RF
0	1	0	Χ	Χ	S2	S1	S0	Audio switch
0	1	1	0	C3	C2	C1	C0	Bass control
0	1	1	1	C3	C2	C1	C0	Treble control

X = don't care Ax = 2dB steps Bx = 10dB stepsCx = 2.5dB steps

Status after power-on reset

Volume	-68dB
Speaker	-38dB
Audio Switch	Mono
Bass	+2.5dB
Treble	+2.5dB

SOFTWARE SPECIFICATION (continued) DATA BYTES (detailed description)

VOLUME

MSB							LSB	
0	0	B2	B1	B0	A2	A1	A0	Volume 2dB Steps
					0	0	0	0
					0	0	1	-2
					0	1	0	-4
					0	1	1	-6
					1	0	0	-8
					1	0	1	Not allowed
					1	1	0	Not allowed
					1	1	1	Not allowed
0	0	B2	B1	B0				Volume 10dB steps
		0	0	0				+10
		0	0	1				0
		0	1	0				-10
		0	1	1				-20
		1	0	0				-30
		1	0	1				-40
		1	1	0				-50
		1	1	1				-60

For example if you want setting the volume at -32dB the 8 bit string is: 0010001

SPEAKER ATTENUATORS

MSB							LSB	
1 1 1	0 0 1 1	0 1 0 1	B1 B1 B1 B1	B0 B0 B0 B0	A2 A2 A2 A2	A1 A1 A1 A1	A0 A0 A0 A0	Speaker LF Speaker RF Speaker LR Speaker RR
					0 0 0 0 1 1 1	0 0 1 1 0 0 1 1	0 1 0 1 0 1 0	0 -2 -4 -6 -8 Not allowed Not allowed Not allowed Not allowed
			0 0 1 1	0 1 0 1				0 -10 -20 -30

For example attenuation of 24dB on speaker RF is given by: 1 0 1 1 0 0 1 0

SOFTWARE SPECIFICATION (continued)

AUDIO SWITCH - Select the input Channel to Activate

MSB							LSB	
0	1	0	Χ	Χ	S2	S1	S0	Audio Switch
			Χ	Χ	0	0	0	Stereo 1
			Χ	Χ	0	0	1	Stereo 2
			Χ	Χ	0	1	0	Stereo 3
			Χ	Χ	0	1	1	Stereo 4
			Х	Χ	1	0	0	Mono
			Х	Χ	1	0	1	Not Allowed
			Х	Χ	1	1	0	Not Allowed
			Χ	Χ	1	1	1	Not Allowed

X = don't care

For example to set the stereo 2 channel the 8 bit string may be: 0 1 0 0 0 0 0 1

BASS AND TREBLE - Control Range of \pm 15dB (boost and cut) Steps of 2.5dB

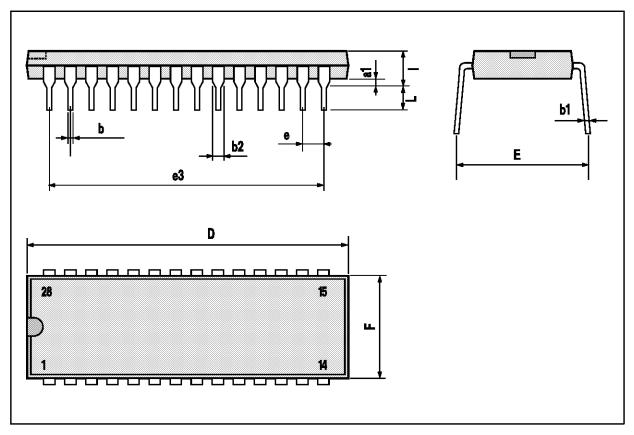
0	1 1	1 1	0 1	C3	C2 C2	C1 C1	C0 C0	Bass Treble
				0	0	0	0	- 15
				0	0	0	1	- 15
				0	0	1	0	- 12.5
				0	0	1	1	- 10
				0	1	0	0	- 7.5
				0	1	0	1	- 5
				0	1	1	0	- 2.5
				0	1	1	1	- 0
				1	1	1	1	0
				1	1	1	0	2.5
				1	1	0	1	5
				1	1	0	0	7.5
				1	0	1	1	10
				1	0	1	0	12.5
				1	0	0	1	15
				1	0	0	0	15

C3 = Sign

For example Bass at -12.5dB is obtained by the following 8 bit string: 01 100010

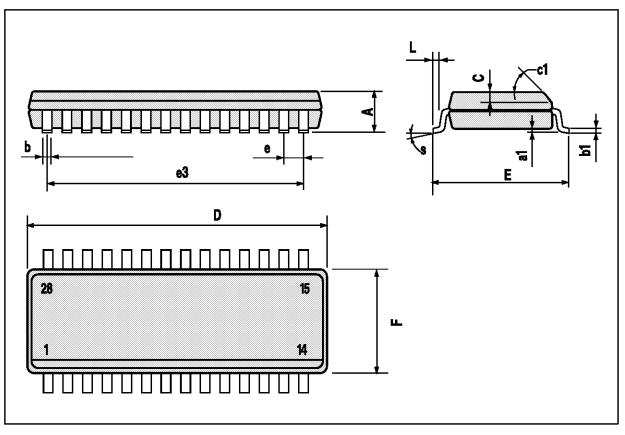
DIP28 PACKAGE MECHANICAL DATA

DIM.		mm		inch			
-	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
a1		0.63			0.025		
b		0.45			0.018		
b1	0.23		0.31	0.009		0.012	
b2		1.27			0.050		
D			37.34			1.470	
E	15.2		16.68	0.598		0.657	
е		2.54			0.100		
e3		33.02			1.300		
F			14.1			0.555	
I		4.445			0.175		
L		3.3			0.130		



SO28 PACKAGE MECHANICAL DATA

DIM.		mm		inch			
DIWI.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
А			2.65			0.104	
a1	0.1		0.3	0.004		0.012	
b	0.35		0.49	0.014		0.019	
b1	0.23		0.32	0.009		0.013	
С		0.5			0.020		
c1			45°	(typ.)			
D	17.7		18.1	0.697		0.713	
E	10		10.65	0.394		0.419	
е		1.27			0.050		
e3		16.51			0.65		
F	7.4		7.6	0.291		0.299	
L	0.4		1.27	0.016		0.050	
S			8° (r	nax.)			



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