DATA SHEET

TDA1072A

AM receiver circuit

Product specification

File under Integrated Circuits, IC01

May 1984

Philips
Semiconductors

PHILIPS

AM receiver circuit   TDA1072A

GENERAL DESCRIPTION

The TDA1072A integrated AM receiver circuit performs the active and part of the filtering functions of an AM radio receiver. It is intended for use in mains-fed home receivers and car radios. The circuit can be used for oscillator frequencies up to 50 MHz and can handle r.f. signals up to 500 mV. R.F. radiation and sensitivity to interference are minimized by an almost symmetrical design. The voltage-controlled oscillator provides signals with extremely low distortion and high spectral purity over the whole frequency range even when tuning with variable capacitance diodes. If required, band switching diodes can easily be applied. Selectivity is obtained using a block filter before the i.f. amplifier.

Features

- Inputs protected against damage by static discharge
- Gain-controlled r.f. stage
- Double balanced mixer
- Separately buffered, voltage-controlled and temperature-compensated oscillator, designed for simple coils
- Gain-controlled i.f. stage with wide a.g.c. range
- Full-wave, balanced envelope detector
- Internal generation of a.g.c. voltage with possibility of second-order filtering
- Buffered field strength indicator driver with short-circuit protection
- A.F. preamplifier with possibilities for simple a.f. filtering
- Electronic standby switch.

QUICK REFERENCE DATA

Supply voltage range	V_P		7,5 to 18	V
Supply current range	I_P		15 to 30	mA
R.F. input voltage for S + N/N = 6 dB at m = 30%	V_i	typ.	1,5	µV
R.F. input voltage for 3% total harmonic distortion (THD) at m = 80%	V_i	typ.	500	mV
A.F. output voltage with V_i = 2 mV; f_i = 1 MHz; m = 30% and f_m = 400 Hz	$V_{o(af)}$	typ.	310	mV
A.G.C. range: change of V_i for 1 dB change of $V_{o(af)}$		typ.	86	dB
Field strength indicator voltage at V_i = 500 mV; $R_{L(9)}$ = 2,7 kΩ	V_{IND}	typ.	2,8	V

PACKAGE OUTLINE

16-lead DIL; plastic (SOT38); SOT38-1; 1996 August 09.

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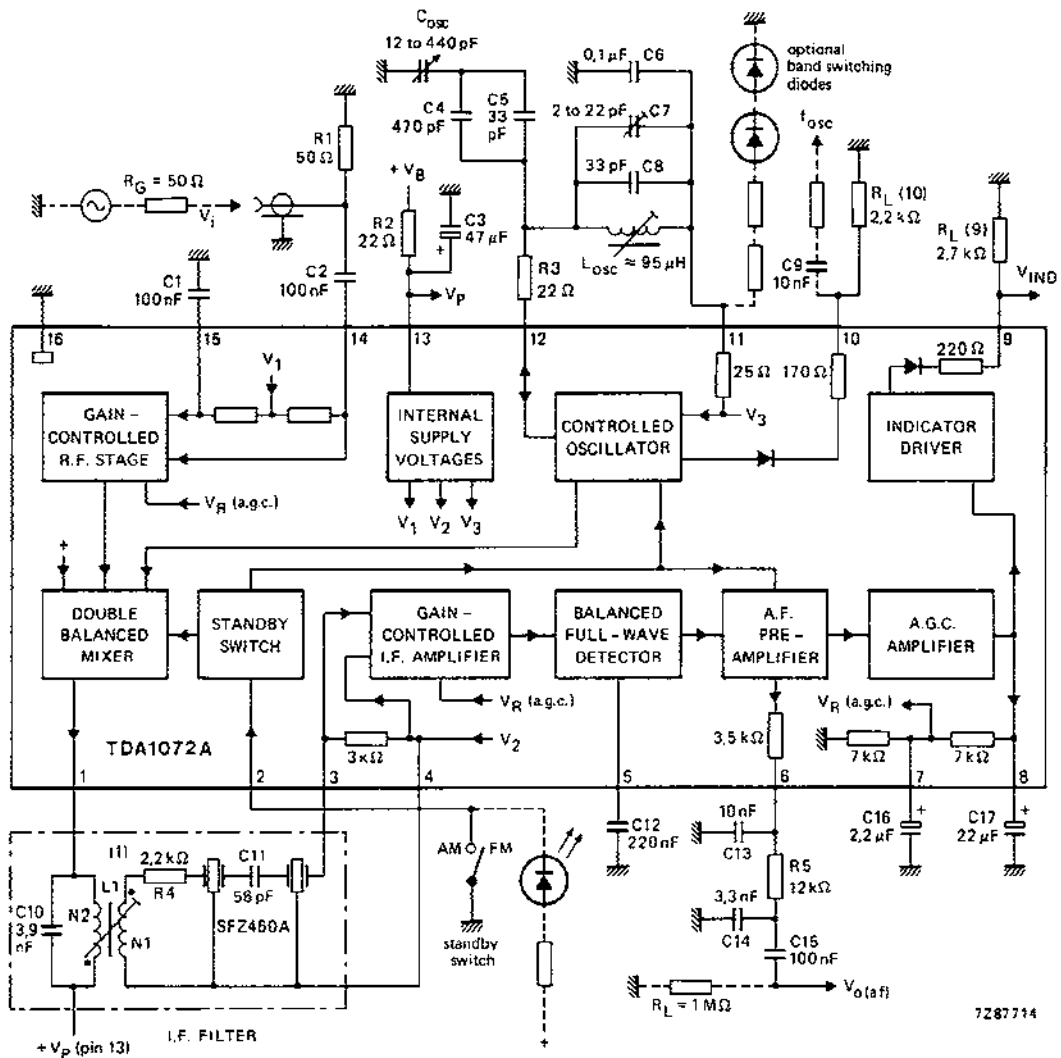
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(1) Coil data: TOKO sample no. 7XNS-A7523DY; L1 : N1/N2 = 12/32; Q_o = 65; Q_B = 57.
  Filter data: Z_F = 700 Ω at R_{3-4} = 3 kΩ; Z_I = 4,8 kΩ.

Fig.1 Block diagram and test circuit (connections shown in broken lines are not part of the test circuit).

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FUNCTIONAL DESCRIPTION

Gain-controlled r.f. stage and mixer

The differential amplifier in the r.f. stage employs an a.g.c. negative feedback network to provide a wide dynamic range. Very good cross-modulation behaviour is achieved by a.g.c. delays at the various signal stages. Large signals are handled with low distortion and the S/N ratio of small signals is improved. Low noise working is achieved in the differential amplifier by using transistors with low base resistance.

A double balanced mixer provides the i.f. output signal to pin 1.

Oscillator

The differential amplifier oscillator is temperature compensated and is suitable for simple coil connection. The oscillator is voltage-controlled and has little distortion or spurious radiation. It is specially suitable for electronic tuning using variable capacitance diodes. Band switching diodes can easily be applied using the stabilized voltage $V_{11\text{-}16}$. An extra buffered oscillator output (pin 10) is available for driving a synthesizer. If this is not needed, resistor $R_{L(10)}$ can be omitted.

Gain-controlled i.f. amplifier

This amplifier comprises two cascaded, variable-gain differential amplifier stages coupled by a band-pass filter. Both stages are gain-controlled by the a.g.c. negative feedback network.

Detector

The full-wave, balanced envelope detector has very low distortion over a wide dynamic range. Residual i.f. carrier is blocked from the signal path by an internal low-pass filter.

A.F. preamplifier

This stage preamplifies the audio frequency output signal. The amplifier output has an emitter follower with a series resistor which, together with an external capacitor, yields the required low-pass for a.f. filtering.

A.G.C. amplifier

The a.g.c. amplifier provides a control voltage which is proportional to the carrier amplitude. Second-order filtering of the a.g.c. voltage achieves signals with very little distortion, even at low audio frequencies. This method of filtering also gives fast a.g.c. settling time which is advantageous for electronic search tuning. The a.g.c. settling time can be further reduced by using capacitors of smaller value in the external filter (C16 and C17). The a.g.c. voltage is fed to the r.f. and i.f. stages via suitable a.g.c. delays. The capacitor at pin 7 can be omitted for low-cost applications.

Field strength indicator output

A buffered voltage source provides a high-level field strength output signal which has good linearity for logarithmic input signals over the whole dynamic range. If the field strength information is not needed, $R_{L(9)}$ can be omitted.

Standby switch

This switch is primarily intended for AM/FM band switching. During standby mode the oscillator, mixer and a.f. preamplifier are switched off.

Short-circuit protection

All pins have short-circuit protection to ground.

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RATINGS

Limiting values in accordance with the Absolute Maximum Rating System (IEC 134)

Supply voltage	$V_P = V_{13-16}$	max.	20 V
Total power dissipation	P_{tot}	max.	875 mW
Input voltage	$|V_{14-15}|$	max.	12 V
	$-V_{14-16}, -V_{15-16}$	max.	0,6 V
	V_{14-16}, V_{15-16}	max.	V_P V
Input current	$|I_{14}|, |I_{15}|$	max.	200 mA
Operating ambient temperature range	T_{amb}	-40 to	$+80$ °C
Storage temperature range	T_{stg}	-55 to	$+150$ °C
Junction temperature	T_j	max.	$+125$ °C

THERMAL RESISTANCE

From junction to ambient	$R_{th\,j-a}$	=	80	K/W

DEVICE CHARACTERISTICS

$V_P = V_{13-16} = 8,5$ V; $T_{amb} = 25$ °C; $f_i = 1$ MHz; $f_m = 400$ Hz; $m = 30\%$; $f_{if} = 460$ kHz; measured in test circuit of Fig.1; unless otherwise specified

PARAMETER	SYMBOL	MIN.	TYP.	MAX.	UNIT
Supplies					
Supply voltage	$V_P = V_{13-16}$	7,5	8,5	18	V
Supply current	$I_P = I_{13}$	15	23	30	mA
R.F. stage and mixer					
Input voltage (d.c. value)	V_{14-16}, V_{15-16}	–	$V_P/2$	–	V
R.F. input impedance at $V_i < 300$ µV	R_{14-16}, R_{15-16}	–	5,5	–	kΩ
	C_{14-16}, C_{15-16}	–	25	–	pF
R.F. input impedance at $V_i > 10$ mV	R_{14-16}, R_{15-16}	–	8	–	kΩ
	C_{14-16}, C_{15-16}	–	22	–	pF
I.F. output impedance	R_{1-16}	500	–	–	kΩ
	C_{1-16}	–	6	–	pF
Conversion transconductance before start of a.g.c.	I_1/V_i	–	6,5	–	mA/V
Maximum i.f. output voltage, inductive coupling to pin 1	$V_{1-13(p-p)}$	–	5	–	V
D.C. value of output current (pin 1) at $V_i = 0$ V	I_1	–	1,2	–	mA
A.G.C. range of input stage		–	30	–	dB
R.F. signal handling capability: input voltage for THD = 3% at m = 80%	$V_{i(rms)}$	–	500	–	mV

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PARAMETER	SYMBOL	MIN.	TYP.	MAX.	UNIT
Oscillator					
Frequency range	f_{osc}	0,6	–	60	MHz
Oscillator amplitude (pins 11 to 12)	$V_{11\text{-}12}$	–	130	150	mV
External load impedance	$R_{12\text{-}11(ext)}$	0,5	–	200	kΩ
External load impedance for no oscillation	$R_{12\text{-}11(ext)}$	–	–	60	Ω
Ripple rejection at $V_{P(rms)}$ = 100 mV;					
$\quad f_P$ = 100 Hz					
$\quad$ (RR = 20 log [$V_{13\text{-}16}/V_{11\text{-}16}$])	RR	–	55	–	dB
Source voltage for switching diodes ($6 \times V_{BE}$)	$V_{11\text{-}16}$	–	4,2	–	V
D.C. output current (for switching diodes)	$-I_{11}$	0	–	20	mA
Change of output voltage at					
$\quad \Delta I_{11}$ = 20 mA (switch to maximum load)	$\Delta V_{11\text{-}16}$	–	0,5	–	V
Buffered oscillator output					
D.C. output voltage	$V_{10\text{-}16}$	–	0,7	–	V
Output signal amplitude	$V_{10\text{-}16(p\text{-}p)}$	–	320	–	mV
Output impedance	R_{10}	–	170	–	Ω
Output current	$-I_{10(peak)}$	–	–	3	mA
I.F., a.g.c. and a.f. stages					
D.C. input voltage	$V_{3\text{-}16}$, $V_{4\text{-}16}$	–	2,0	–	V
I.F. input impedance	$R_{3\text{-}4}$	2,4	3	3,9	kΩ
	$C_{3\text{-}4}$	–	7	–	pF
I.F. input voltage for					
$\quad$ THD = 3% at m = 80%	$V_{3\text{-}4}$	–	90	–	mV
Voltage gain before start of a.g.c.	$V_{3\text{-}4}/V_{6\text{-}16}$	–	68	–	dB
A.G.C. range of i.f. stages: change of					
$\quad V_{3\text{-}4}$ for 1 dB change of $V_{o(af)}$;					
$\quad V_{3\text{-}4(ref)}$ = 75 mV	$\Delta V_{3\text{-}4}$	–	55	–	dB
A.F. output voltage at $V_{3\text{-}4(if)}$ = 50 µV	$V_{o(af)}$	–	130	–	mV
A.F. output voltage at $V_{3\text{-}4(if)}$ = 1 mV	$V_{o(af)}$	–	310	–	mV
A.F. output impedance (pin 6)	$|Z_o|$	–	3,5	–	kΩ
Indicator driver					
Output voltage at V_i = 0 mV;					
$\quad R_{L(9)}$ = 2,7 kΩ	$V_{9\text{-}16}$	–	20	150	mV
Output voltage at V_i = 500 mV;					
$\quad R_{L(9)}$ = 2,7 kΩ	$V_{9\text{-}16}$	2,5	2,8	3,1	V
Load resistance	$R_{L(9)}$	1,5	–	–	kΩ

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PARAMETER	SYMBOL	MIN.	TYP.	MAX.	UNIT
Standby switch					
Switching threshold at V_P = 7,5 to 18 V;					
$\quad T_{amb}$ = −40 to + 80 °C					
$\quad$ on-voltage	$V_{2\text{-}16}$	0	–	2,0	V
$\quad$ off-voltage	$V_{2\text{-}16}$	3,5	–	20	V
$\quad$ on-current at $V_{2\text{-}16}$ = 0 V	$-I_2$	–	–	200	µA
$\quad$ off-current at $V_{2\text{-}16}$ = 20 V	$|I_2|$	–	–	10	µA

OPERATING CHARACTERISTICS

V_P = 8,5 V; f_i = 1 MHz; m = 30%; f_m = 400 Hz; T_{amb} = 25 °C; measured in Fig.1; unless otherwise specified

PARAMETER	SYMBOL	MIN.	TYP.	MAX.	UNIT
R.F. sensitivity					
R.F. input required for S + N/N = 6 dB	V_i	–	1,5	–	µV
R.F. input required for S + N/N = 26 dB	V_i	–	15	–	µV
R.F. input required for S + N/N = 46 dB	V_i	–	150	–	µV
R.F. input at start of a.g.c.	V_i	–	30	–	µV
R.F. large signal handling					
R.F. input at THD = 3%; m = 80%	V_i	–	500	–	mV
R.F. input at THD = 3%; m = 30%	V_i	–	700	–	mV
R.F. input at THD = 10%; m = 30%	V_i	–	900	–	mV
A.G.C. range					
Change of V_i for 1 dB change					
$\quad$ of $V_{o(af)}$; $V_{i(ref)}$ = 500 mV	ΔV_i	–	86	–	dB
Change of V_i for 6 dB change					
$\quad$ of $V_{o(af)}$; $V_{i(ref)}$ = 500 mV	ΔV_i	–	91	–	dB
Output signal					
A.F. output voltage at					
$\quad V_i$ = 4 µV; m = 80%	$V_{o(af)}$	–	130	–	mV
A.F. output voltage at V_i = 1 mV	$V_{o(af)}$	240	310	390	mV
THD at V_i = 1 mV; m = 80%	d_{tot}	–	0,5	–	%
THD at V_i = 500 mV; m = 30%	d_{tot}	–	1	–	%
Signal-to-noise ratio at V_i = 100 mV	(S + N)/N	–	58	–	dB
Ripple rejection at V_i = 2 mV;					
$\quad V_{P(rms)}$ = 100 mV; f_P = 100 Hz					
$\quad$ (RR = 20 log [$V_P/V_{o(af)}$])	RR	–	38	–	dB

PARAMETER	SYMBOL	MIN.	TYP.	MAX.	UNIT
Unwanted signals					
Suppression of i.f. whistles at					
V_i = 15 μV; m = 0% related to a.f. signal of m = 30%					
at $f_i \approx 2 \times f_{if}$	α_{2if}	–	37	–	dB
at $f_i \approx 3 \times f_{if}$	α_{3if}	–	44	–	dB
I.F. suppression at r.f. input					
for symmetrical input	α_{if}	–	40	–	dB
for asymmetrical input	α_{if}	–	40	–	dB
Residual oscillator signal at mixer output					
at f_{osc}	$I_{1(osc)}$	–	1	–	μA
at $2 \times f_{osc}$	$I_{1(2osc)}$	–	1,1	–	μA

APPLICATION INFORMATION

(1) Capacitor values depend on crystal type.
(2) Coil data: 9 windings of 0,1 mm dia laminated Cu wire on TOKO coil set 7K 199CN; Q_o = 80.

Fig.2 Oscillator circuit using quartz crystal; centre frequency = 27 MHz.
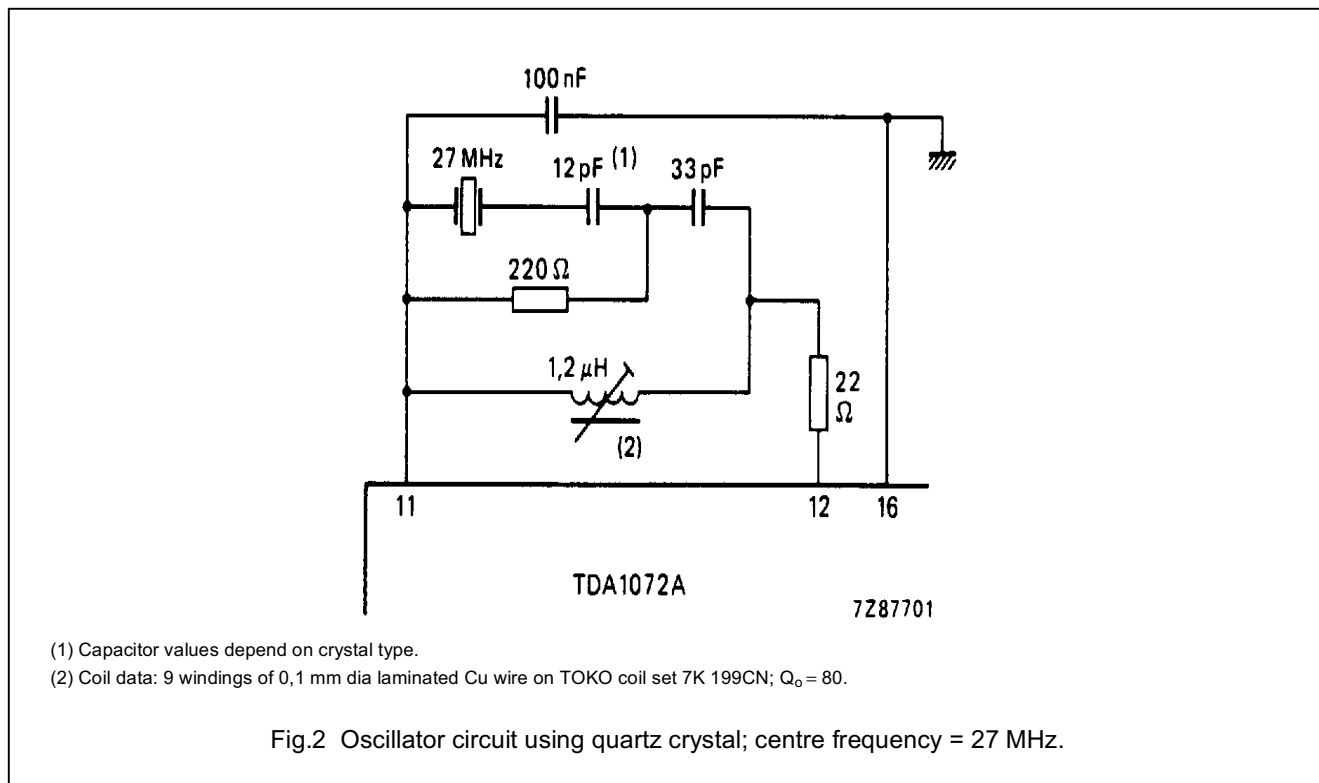
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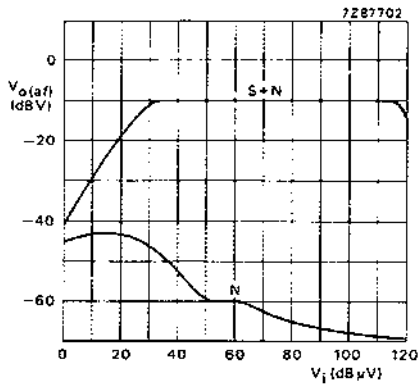
Fig.3 A.F. output as a function of r.f. input in the circuit of Fig.1; f_i = 1 MHz; f_m = 400 Hz; m = 30%.
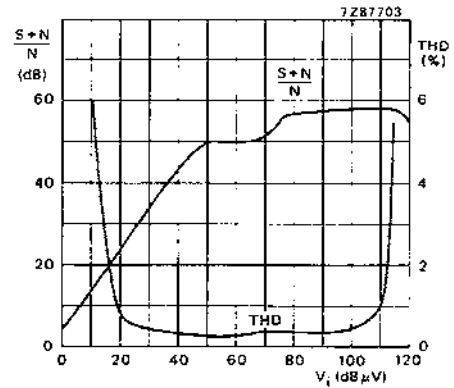
Fig.4 Total harmonic distortion and (S + N)/N as functions of r.f. input in the circuit of Fig.1; m = 30% for (S + N)/N curve and m = 80% for THD curve.
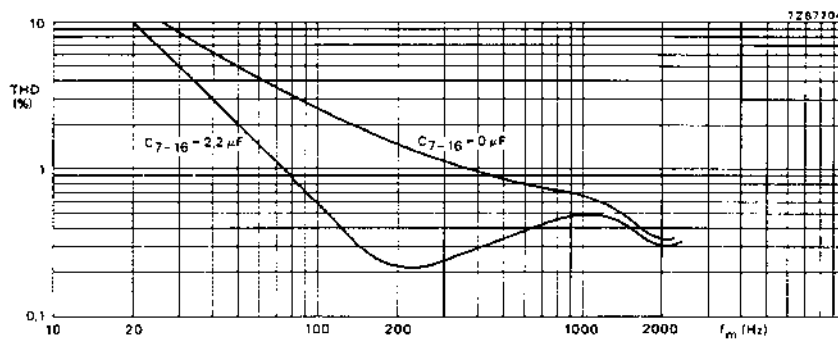
Fig.5 Total harmonic distortion as a function of modulation frequency at V_i = 5 mV; m = 80%; measured in the circuit of Fig.1 with $C_{7-16(ext)}$ = 0 µF and 2,2 µF.
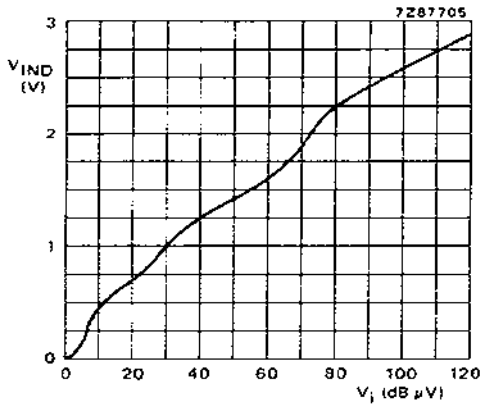
Fig.6 Indicator driver voltage as a function of r.f. input in the circuit of Fig.1.
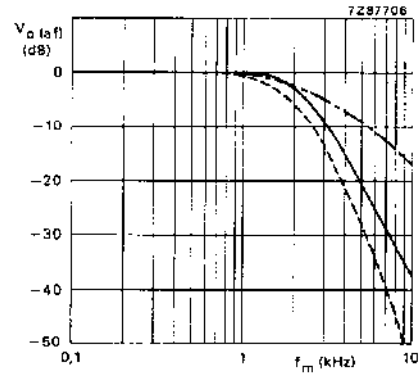
Fig.7 Typical frequency response curves from Fig.1 showing the effect of filtering as follows:

———— with i.f. filter;
– – – – – with a.f. filter;
– – – – – with i.f. and a.f. filters.
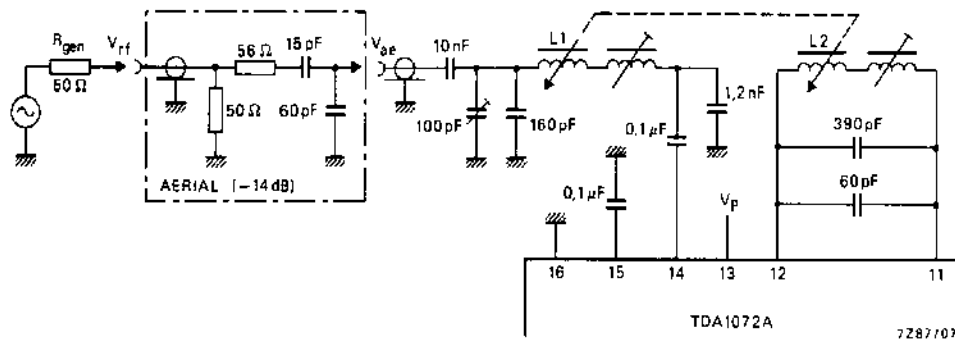
Fig.8 Car radio application with inductive tuning.
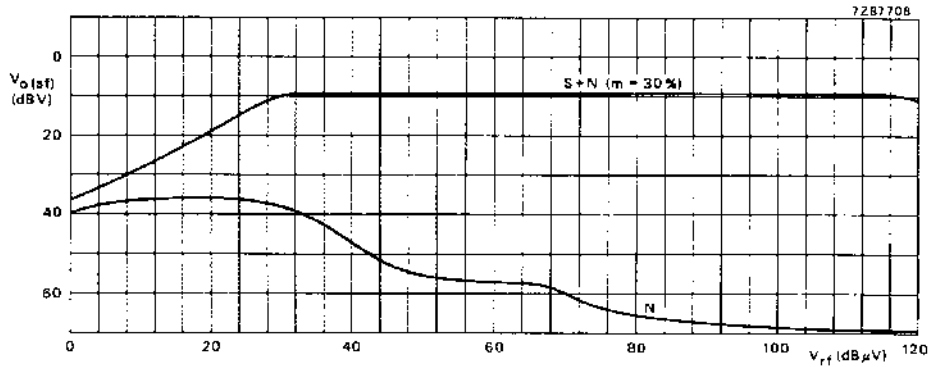
Fig.9 A.F. output as a function of r.f. input using the circuit of Fig.8 with that of Fig.1.
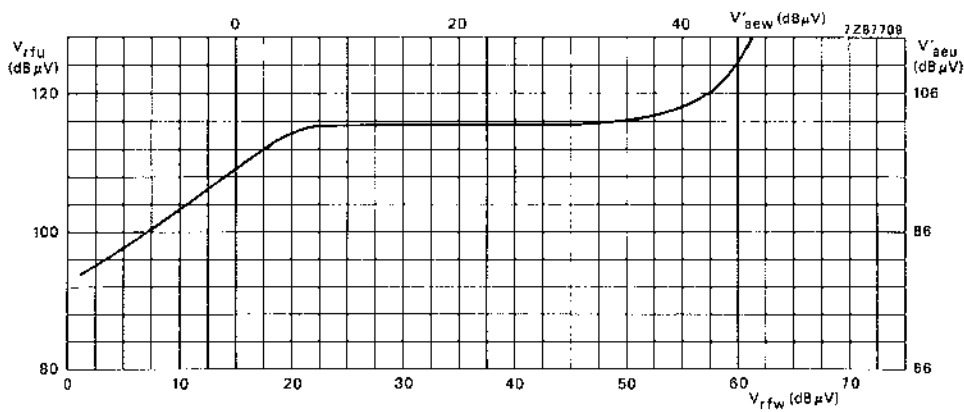
Fig.10 Suppression of cross-modulation as a function of input signal, measured in the circuit of Fig.8 with the input circuit as shown in Fig.11. Curve is for Wanted $V_{o(af)}$/Unwanted $V_{o(af)}$ = 20 dB; V_{rfw}, V_{rfu} are signals at the aerial input, V'_{aew}, V'_{aeu} are signals at the unloaded output of the aerial.
Wanted signal (V'_{aew}, V_{rfw}): f_i = 1 MHz; f_m = 400 Hz; m = 30%.
Unwanted signal (V'_{aeu}, V_{rfu}): f_i = 900 kHz; f_m = 400 Hz; m = 30%.
Effective selectivity of input tuned circuit = 21 dB.
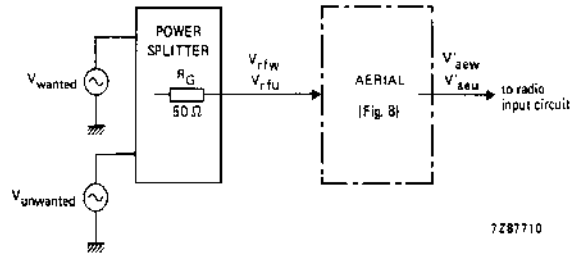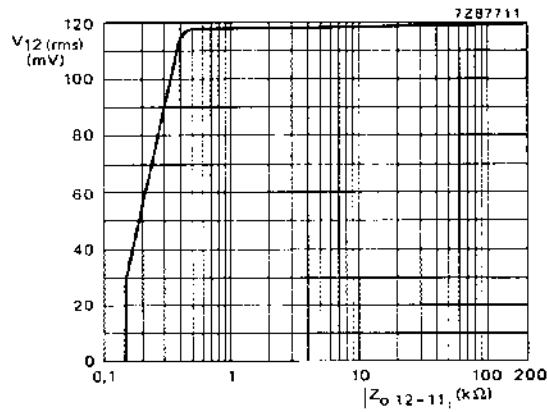
Fig.11 Input circuit to show cross-modulation suppression (see Fig.10).

Fig.12 Oscillator amplitude as a function of pin 11, 12 impedance in the circuit of Fig.8.
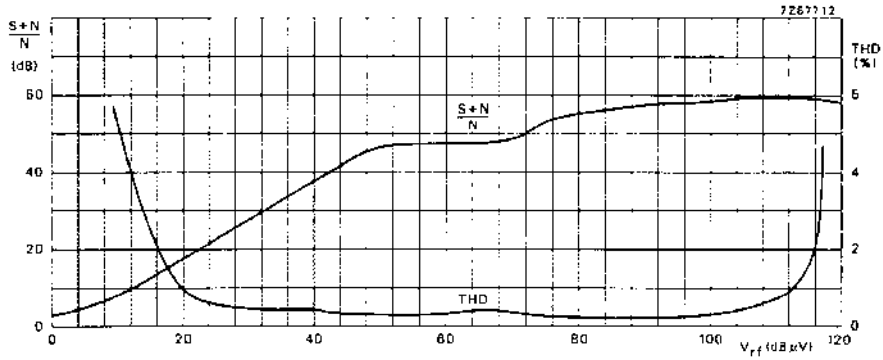
Fig.13 Total harmonic distortion and (S + N)/N as functions of r.f. input using the circuit of Fig.8 with that of Fig.1.

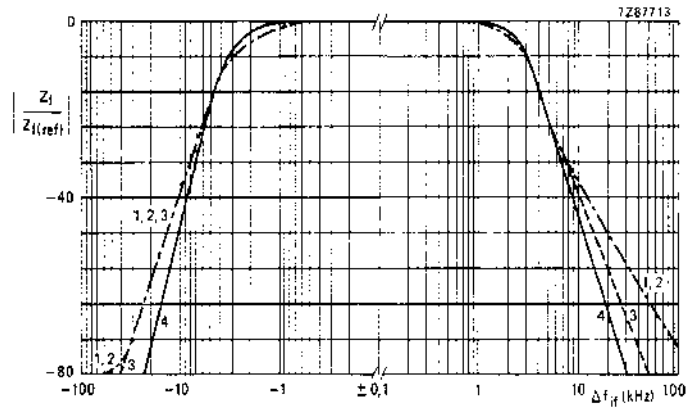
Fig.14 Forward transfer impedance as a function of intermediate frequency for filters 1 to 4 shown
in Fig.15; centre frequency = 455 kHz.

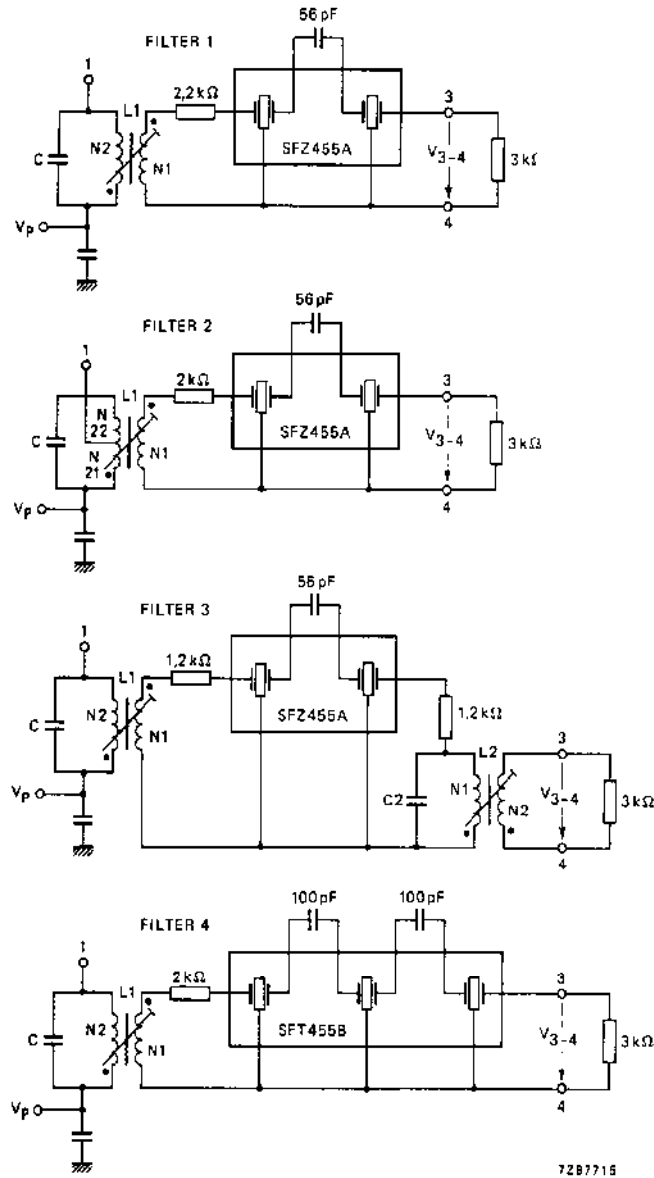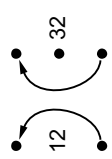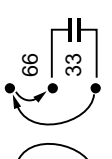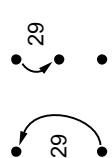Fig.15 I.F. filter variants applied to the circuit of Fig.1. For filter data, refer to Table 1.

Table 1 Data for I.F. filters shown in Fig.15. Criterium for adjustment is Z_F = maximum (optimum selectivity curve at centre frequency f_0 = 455 kHz). See also Fig.14.

FILTER NO.	1	2	3		4	UNIT
Coil data	L1	L1	L2	L1	L1	
Value of C	3900	430	4700	3900	3900	pF
N1: N2	12 : 32	13 : (33 + 66)	29 : 29	15 : 31	13 : 31	
Diameter of Cu laminated wire	0,09	0,08	0,08	0,09	0,09	mm
Q_o	65 (typ.)	50	60	75	75	
Schematic* of windings	(winding schematic: 12, 32)	(winding schematic: 13, 66, 33)	(winding schematic: 29, 29) (N1) (N2)	(winding schematic: 15, 31)	(winding schematic: 13, 31)	
Toko order no.	7XNS-A7523DY	L7PES-A0060BTG	7XNS-A7521AIH	7XNS-A7518DY	7XNS-A7519DY	
Resonators						
Murata type	SFZ455A	SFZ455A		SFZ455A	SFT455B	
D (typical value)	4	4		4	6	dB
R_G, R_L	3	3		3	3	kΩ
Bandwidth (−3 dB)	4,2	4,2		4,2	4,5	kHz
S_{9kHz}	24	24		24	38	dB
Filter data						
Z_I	4,8	3,8	52 (L1)	4,2	4,8	kΩ
Q_B	57	40			55	
Z_F	0,70	0,67	18 (L2)	0,68	0,68	kΩ
Bandwidth (−3 dB)	3,6	3,8		3,6	4,0	kHz
S_{9kHz}	35	31		36	42	dB
S_{18kHz}	52	49		54	64	dB
S_{27kHz}	63	58		66	74	dB

* The beginning of an arrow indicates the beginning of a winding; N1 is always the inner winding, N2 the outer winding.
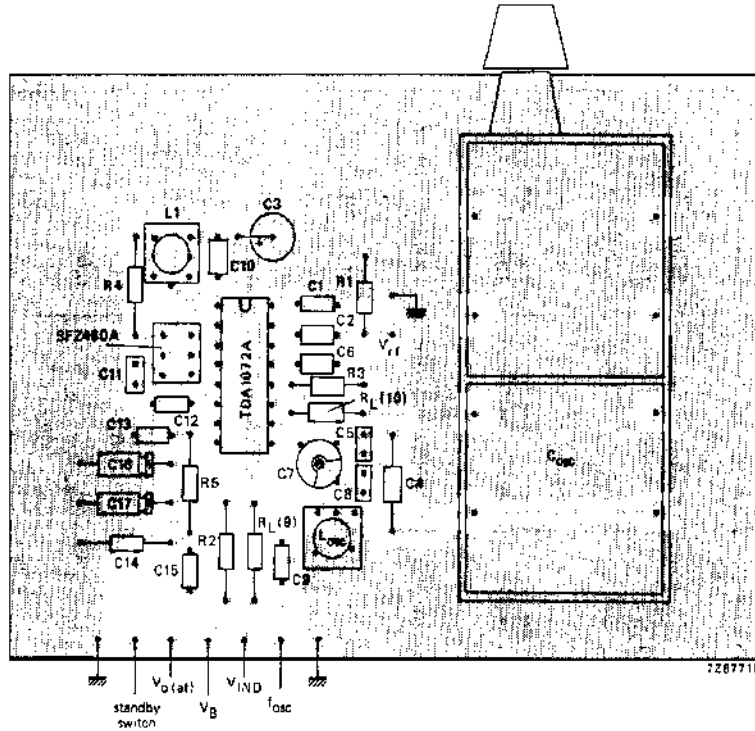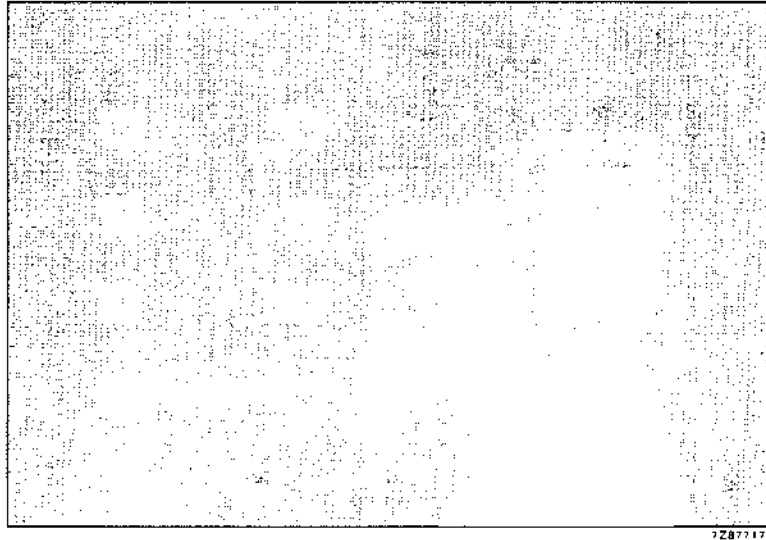
AM receiver circuit

Fig.16 Printed-circuit board component side, showing component layout. For circuit diagram see Fig.1.

AM receiver circuit

Fig.17 Printed-circuit board showing track side.

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(1) Values of capacitors depend on the selected group of capacitive diodes BB112.
(2) For i.f. filter and coil data refer to Fig.1.

Fig.18 Car radio application with capacitive diode tuning and electronic MW/LW switching. The circuit includes pre-stage a.g.c. optimized for good large-signal handling.
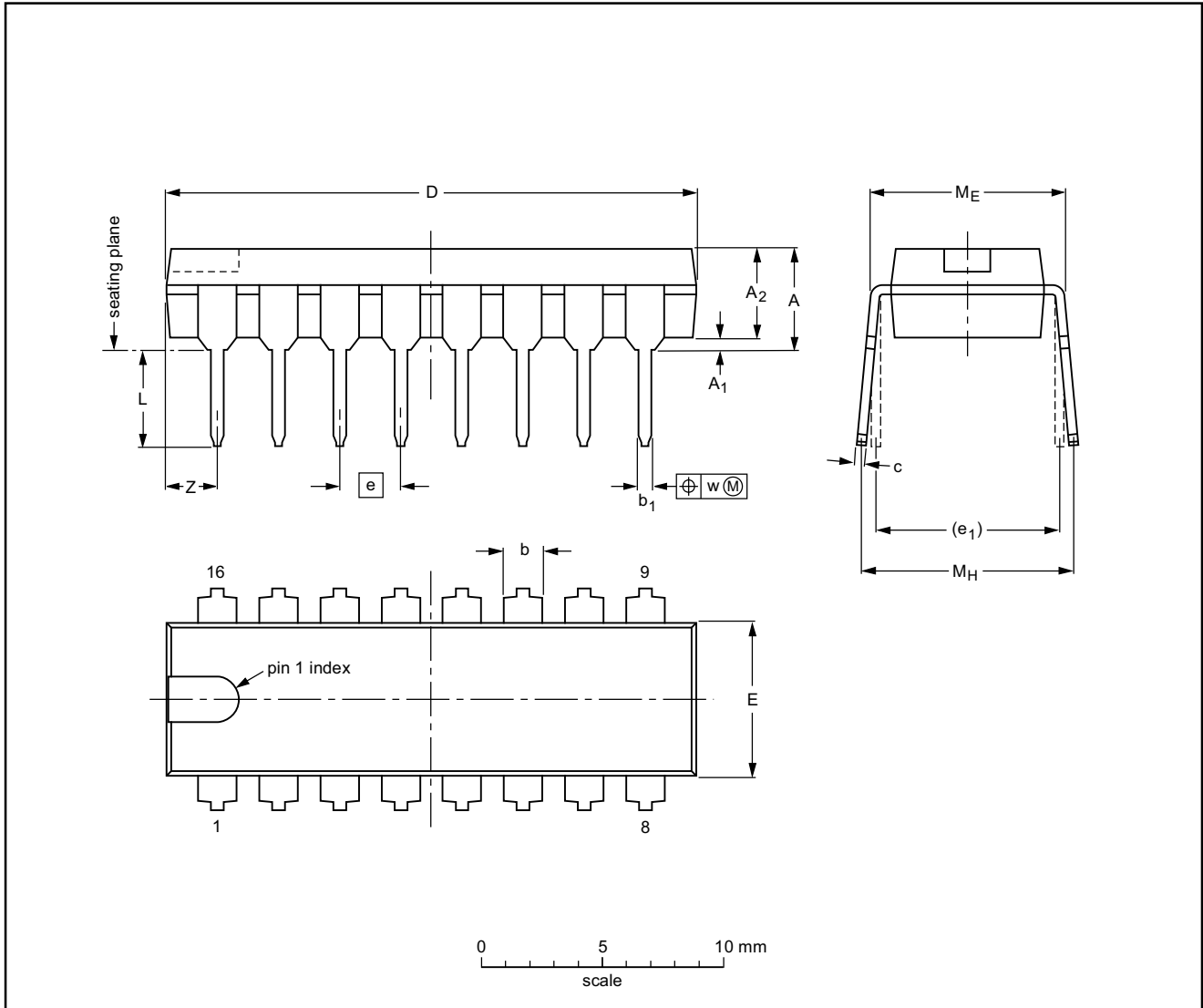
AM receiver circuit

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PACKAGE OUTLINE

DIP16: plastic dual in-line package; 16 leads (300 mil); long body

SOT38-1

0   5   10 mm

scale

DIMENSIONS (inch dimensions are derived from the original mm dimensions)

UNIT	A max.	A₁ min.	A₂ max.	b	b₁	c	D (1)	E (1)	e	e₁	L	M_E	M_H	w	Z (1) max.
mm	4.7	0.51	3.7	1.40 1.14	0.53 0.38	0.32 0.23	21.8 21.4	6.48 6.20	2.54	7.62	3.9 3.4	8.25 7.80	9.5 8.3	0.254	2.2
inches	0.19	0.020	0.15	0.055 0.045	0.021 0.015	0.013 0.009	0.86 0.84	0.26 0.24	0.10	0.30	0.15 0.13	0.32 0.31	0.37 0.33	0.01	0.087

Note

1. Plastic or metal protrusions of 0.25 mm maximum per side are not included.

OUTLINE VERSION	REFERENCES				EUROPEAN PROJECTION	ISSUE DATE
	IEC	JEDEC	EIAJ			
SOT38-1	050G09	MO-001AE				~~92-10-02~~ 95-01-19

SOLDERING

Introduction

There is no soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and surface mounted components are mixed on one printed-circuit board. However, wave soldering is not always suitable for surface mounted ICs, or for printed-circuits with high population densities. In these situations reflow soldering is often used.

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our *"IC Package Databook"* (order code 9398 652 90011).

Soldering by dipping or by wave

The maximum permissible temperature of the solder is 260 °C; solder at this temperature must not be in contact with the joint for more than 5 seconds. The total contact time of successive solder waves must not exceed 5 seconds.

The device may be mounted up to the seating plane, but the temperature of the plastic body must not exceed the specified maximum storage temperature ($T_{stg\ max}$). If the printed-circuit board has been pre-heated, forced cooling may be necessary immediately after soldering to keep the temperature within the permissible limit.

Repairing soldered joints

Apply a low voltage soldering iron (less than 24 V) to the lead(s) of the package, below the seating plane or not more than 2 mm above it. If the temperature of the soldering iron bit is less than 300 °C it may remain in contact for up to 10 seconds. If the bit temperature is between 300 and 400 °C, contact may be up to 5 seconds.

DEFINITIONS

Data sheet status	
Objective specification	This data sheet contains target or goal specifications for product development.
Preliminary specification	This data sheet contains preliminary data; supplementary data may be published later.
Product specification	This data sheet contains final product specifications.
Limiting values	
Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability.	
Application information	
Where application information is given, it is advisory and does not form part of the specification.	

LIFE SUPPORT APPLICATIONS

These products are not designed for use in life support appliances, devices, or systems where malfunction of these products can reasonably be expected to result in personal injury. Philips customers using or selling these products for use in such applications do so at their own risk and agree to fully indemnify Philips for any damages resulting from such improper use or sale.