

**CA081, CA082, CA084****BiMOS Operational Amplifiers**

With MOS/FET Input, Composite Bipolar/MOS Output

Single Amplifier: CA081 Dual Amplifier: CA082, Quad Amplifier: CA084

**Features:**

- Very low input bias and offset currents
- Input impedance typically  $1.5 \times 10^{12} \Omega$
- Low input offset voltage
- Wide common-mode input voltage range
- Low power consumption
- Fast slew rate
- Unity-gain bandwidth = 5 MHz [typ.]
- Wide output voltage swing
- Low distortion
- Continuous short circuit protection
- Direct replacement for industry type TL080 series in most applications

The RCA-CA081, CA082, and CA084 BiMOS operational amplifiers combine the advantages of MOS and bipolar transistors on the same monolithic chip. The gate-protected MOS/FET (PMOS) input transistors provide high input impedance and a wide common-mode input voltage range. The bipolar and MOS output transistors allow a wide output voltage swing and provide a high output current capability.

The CA081, CA082, and CA084 are internally phase-compensated. All types except the CA082 have provisions for external offset nulling.

These types have an operating-temperature range of 0 to +70°C.

**Applications:**

- Inverters
- High-Q notch filters
- IC preamplifiers
- Unity Gain Absolute Value Amplifiers
- Sample and hold amplifiers
- Active filters

The CA081 and CA082 types are supplied in the 8-lead dual-in-line plastic package (E suffix). The CA084 types are supplied in the 14-lead dual-in-line plastic package (E suffix). They are also available in chip form (H suffix).

**MAXIMUM RATINGS, Absolute Maximum Values:**

DC SUPPLY VOLTAGE V $\pm$ .....	$\pm 18$ V
DIFFERENTIAL INPUT VOLTAGE.....	$\pm 16$ V
INPUT VOLTAGE RANGE .....	$\pm 15$ V
INPUT CURRENT .....	1 mA
OUTPUT SHORT-CIRCUIT DURATION .....	UNLIMITED*
POWER DISSIPATION, P <sub>D</sub> :	
At T <sub>A</sub> = 25°C .....	625 mW
Derating Factors:	
Mini-DIP .....	Derate linearly at 6.67 mW/°C above 56°C
14-Lead DIP.....	Derate linearly at 6.67 mW/°C above 56°C
AMBIENT TEMPERATURE RANGE: .....	0 to +70°C
STORAGE TEMPERATURE RANGE, ALL TYPES.....	-65 to +150°C
LEAD TEMPERATURE (DURING SOLDERING):	
At distance $1/16 \pm 1/32$ (1.59 $\pm$ 0.79 mm) from case for 10 seconds max.	+265°C

\*The output may be shorted to ground or either supply if the maximum temperature and dissipation ratings are observed.

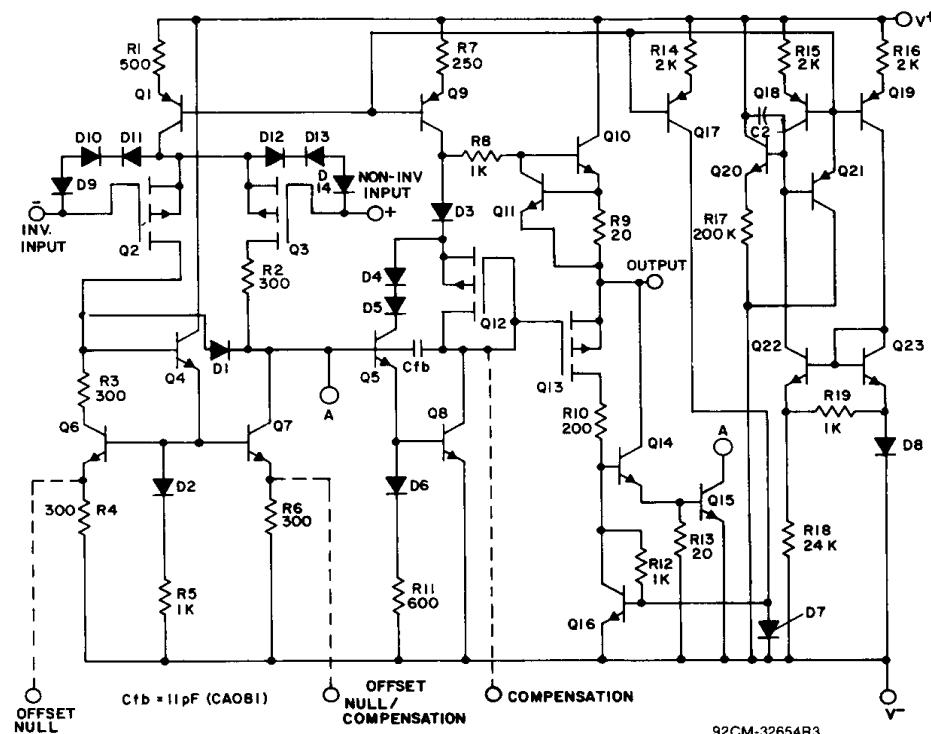
**Operational Amplifiers****CA081, CA082, CA084**

Fig. 1 - Schematic diagram of the CA081, CA082, and CA084.

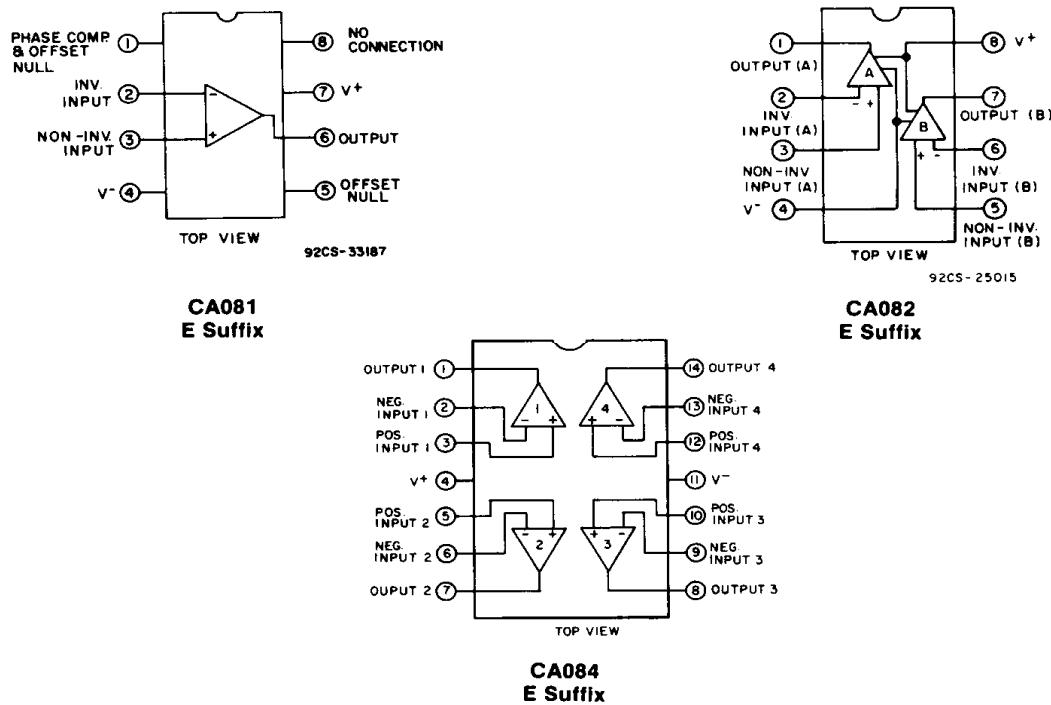


Fig. 2 - Terminal assignments.

**CA081, CA082, CA084****TYPICAL OPERATING CHARACTERISTICS at  
 $V_{\pm} = 15 \text{ V}$ ,  $T_A = 25^\circ\text{C}$** 

CHARACTERISTIC	TEST CONDITIONS	VALUE	UNITS
Slew Rate at Unity Gain, SR	$V_I = 10 \text{ V}$ , $R_L = 2 \text{ k}\Omega$ , $C_L = 100 \text{ pF}$ , $A_{VD} = 1$	13	$\text{V}/\mu\text{s}$
Rise Time, $t_r$	$V_I = 10 \text{ V}$ , $R_L = 2 \text{ k}\Omega$ , $C_L = 100 \text{ pF}$ , $A_{VD} = 1$	0.1	$\mu\text{s}$
Overshoot Factor	$C_L = 100 \text{ pF}$ , $A_{VD} = 1$	10	%
Equivalent Input Noise Voltage, $e_n$	$R_S = 100 \Omega$ , $f = 1 \text{ kHz}$	40	$\text{nV}/\sqrt{\text{Hz}}$

**ELECTRICAL CHARACTERISTICS at  $T_A = 25^\circ\text{C}$ ,  $T_A = 0$  to  $+70^\circ\text{C}$   $V_{\pm} = \pm 15 \text{ V}$** 

CHARACTERISTIC	TEST CONDITIONS	LIMITS			UNITS	
		CA081AE CA082AE CA084AE				
		Min.	Typ.	Max.		
Input Offset Voltage, $V_{IO}$	$R_S = 50\Omega$	X	—	3	6	
		X	—	—	7.5	
Temperature Coefficient of Input Offset Voltage, $\alpha V_{IO}$	$R_S = 50 \Omega$	X	—	10	—	
Input Offset Current, $I_{IO}$		X	—	5	20	
		X	—	—	0.6	
Input Current		X	—	15	40	
		X	—	—	1	
Common-Mode Input Voltage Range, $V_{ICR}$		X	$\pm 12$	—	—	
Maximum Output Voltage Swing, $V_{OP-P}$	$R_L = 10 \text{ k}\Omega$	X	24	27	—	
	$R_L > 10 \text{ k}\Omega$	X	24	—	—	
	$R_L > 2 \text{ k}\Omega$	X	20	24	—	
Large-Signal Differential Voltage Gain, $A_{VD}$	$R_L > 2 \text{ k}\Omega$ , $V_O = \pm 10 \text{ V}$	X	50	200	—	
		X	—	—	—	
Unity-Gain Bandwidth		X	—	5	—	
Input Resistance, $R_I$		X	—	1.5	—	
Common-Mode Rejection Ratio, CMRR	$R_S < 10 \text{ k}\Omega$	X	80	86	—	
Power Supply Rejection Ratio, PSRR ( $\Delta V_{+/-} / \pm \Delta V_{IO}$ )	$R_S < 10 \text{ k}\Omega$	X	80	86	—	
Supply Current, $I^+$ (per ampl., CA082, CA084)	No load, No Signal	X	—	1.4	2.8	
Channel Separation, $V_{O1}/V_{O2}$ (between ampl., CA082)	$A_{VD} = 100$	X	—	120	—	
					dB	

**CA081, CA082, CA084**ELECTRICAL CHARACTERISTICS at  $T_A = 25^\circ\text{C}$ ,  $T_A = 0$  to  $70^\circ\text{C}$   $V \pm = \pm 15\text{ V}$ 

CHARACTERISTIC	TEST CONDITIONS	LIMITS			UNITS	
		CA081E CA082E CA084E				
		Min.	Typ.	Max.		
Input Offset Voltage, $V_{IO}$	$R_S = 50\Omega$	X	—	5	15	
		X	—	—	20	
Temperature Coefficient of Input Offset Voltage, $\alpha V_{IO}$	$R_S = 50\Omega$	X	—	10	—	
Input Offset Current, $I_{IO}$		X	—	5	30	
Input Current		X	—	15	50	
		X	—	—	2	
Common-Mode Input Voltage Range, $V_{ICR}$		X	$\pm 10$	—	—	
Maximum Output Voltage Swing, $V_{OP-P}$	$R_L = 10\text{ k}\Omega$	X	24	27	—	
	$R_L > 10\text{ k}\Omega$	X	24	—	—	
	$R_L > 2\text{ k}\Omega$	X	20	24	—	
Large-Signal Differential Voltage Gain, $AVD$	$R_L > 2\text{ k}\Omega$ , $V_O = \pm 10\text{ V}$	X	25	200	—	
		X	—	—	—	
Unity-Gain Bandwidth		X	—	5	—	
Input Resistance, $R_I$		X	—	1.5	—	
Common-Mode Rejection Ratio, CMRR	$R_S < 10\text{ k}\Omega$	X	70	76	—	
Power Supply Rejection Ratio, PSRR ( $\Delta V_+ / \pm \Delta V_{IO}$ )	$R_S < 10\text{ k}\Omega$	X	70	76	—	
Supply Current, $I^+$ (per ampl., CA082, CA084)	No load, No Signal	X	—	1.4	2.8	
Channel Separation, $V_{O1}/V_{O2}$ (between ampl., CA082)	$AVD = 100$	X	—	120	—	

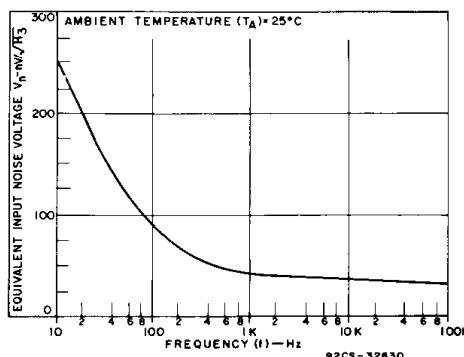


Fig. 3 - Noise voltage as a function of frequency.

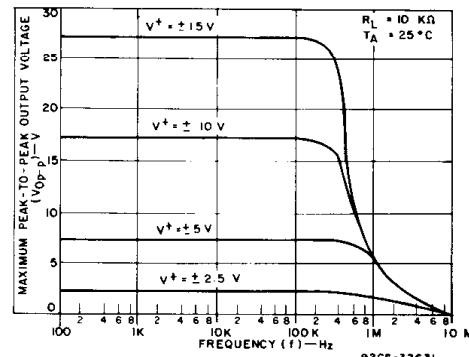


Fig. 4 - Output voltage as a function of frequency.

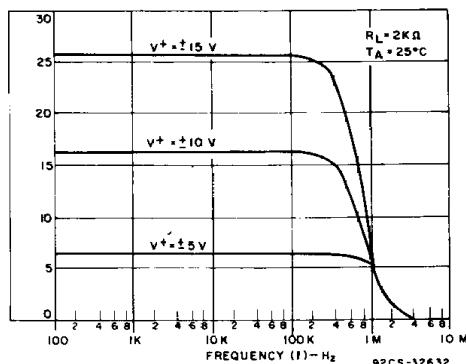
**CA081, CA082, CA084**

Fig. 5 - Output voltage as a function of frequency.

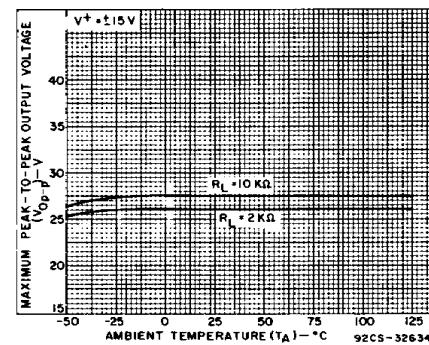


Fig. 6 - Output voltage as a function of ambient temperature.

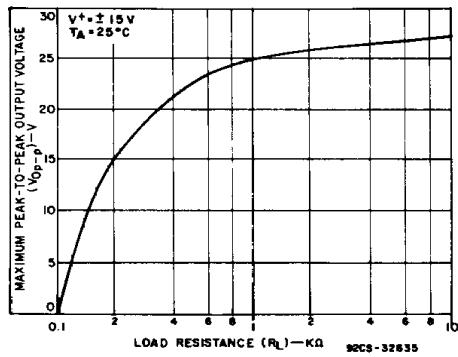


Fig. 7 - Output voltage as a function of load resistance.

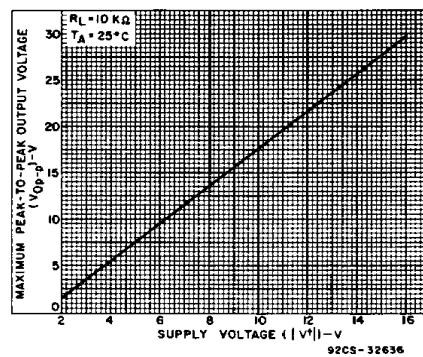


Fig. 8 - Output voltage as a function of supply voltage.

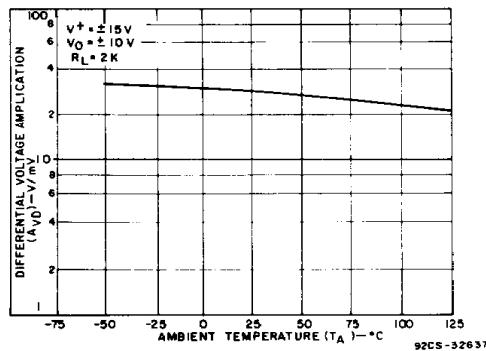


Fig. 9 - Differential voltage amplification as a function of ambient temperature.

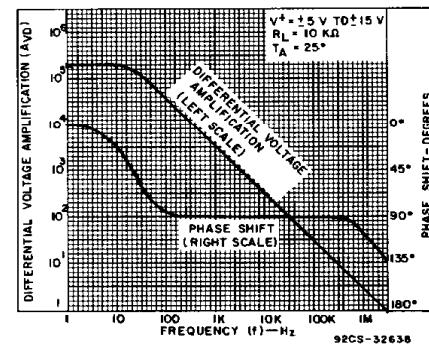
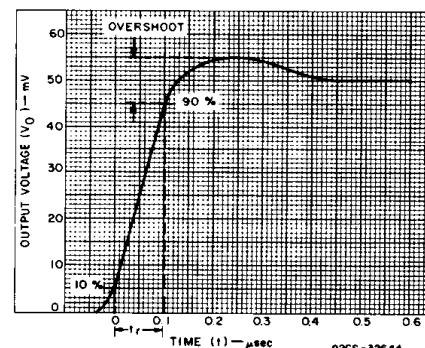
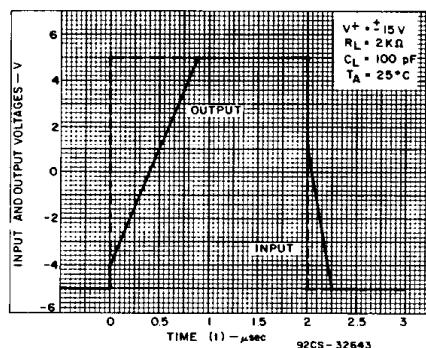
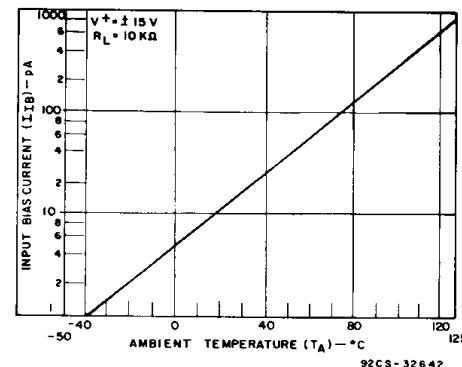
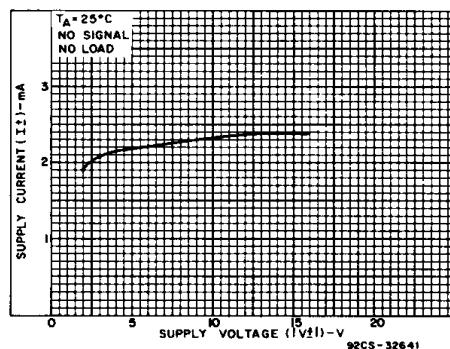
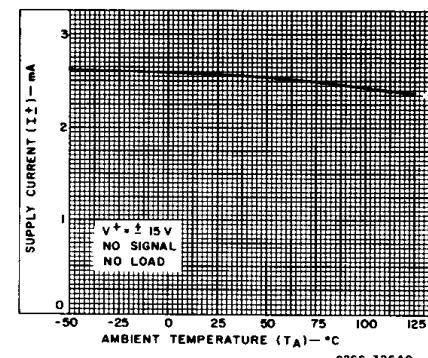
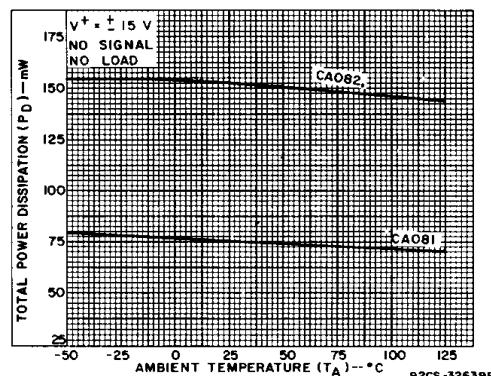


Fig. 10 - Differential voltage amplification as a function of frequency.

**CA081, CA082, CA084**

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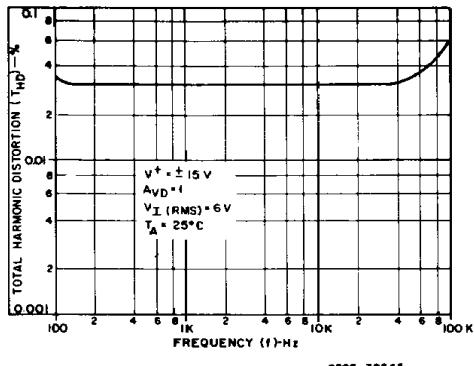
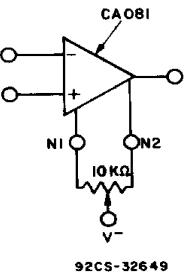
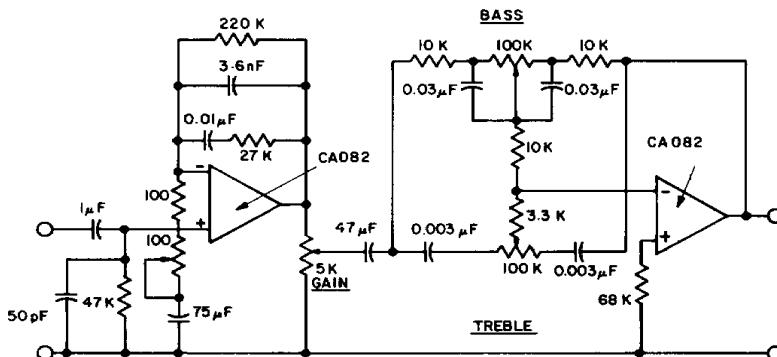


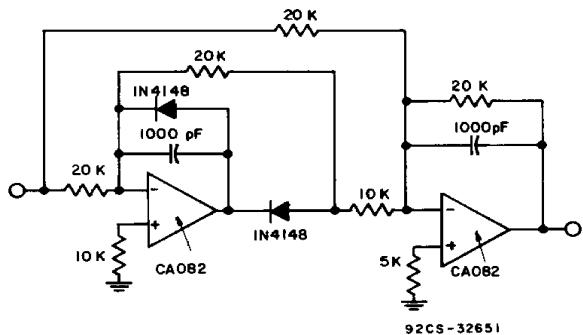
Fig. 17 - Total harmonic distortion as a function of frequency.



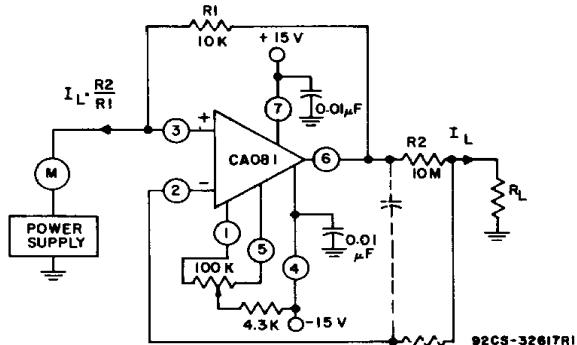
*Fig. 18 - Input offset voltage null circuit.*



*Fig. 19 - IC preamplifier.*



*Fig. 20 - Unity-gain absolute-value amplifier.*



*Fig. 21 - Basic current amplifier for low-current measurement systems.*

## CURRENT AMPLIFIER

The low input-terminal current needed to drive the CA081 makes it ideal for use in current-amplifier applications such as the one shown in Fig. 21. In this circuit, low current is supplied at the input potential as the power supply to load resistor  $R_L$ . This load current is increased by the multiplication factor  $R_2/R_1$ , when the load current is monitored by the power supply meter M. Thus, if the load current is 100 nA, with values shown, the load current presented to the supply will be 100  $\mu$ A; a much easier current to measure in many systems.

Note that the input and output voltages are transferred at the same potential and only the output current is multiplied by the scale factor.

The dotted components show a method of decoupling the circuit from the effects of high output-load capacitance and the potential oscillation in this situation. Essentially, the necessary high-frequency feedback is provided by the capacitor with the dotted series resistor providing load decoupling.