



ACT™ 1 Field Programmable Gate Arrays

Features

- Up to 2000 Gate Array Gates (6000 PLD/LCA™ equivalent gates)
- Replaces up to 53 TTL Packages
- Replaces up to 17 20-Pin PAL™ Packages
- Design Library with over 250 Functions
- Gate Array Architecture Allows Completely Automatic Place and Route
- Up to 547 Programmable Logic Modules
- Up to 273 Flip-Flops
- Flip-Flop Toggle Rates to 100 MHz
- Two In-Circuit Diagnostic Probe Pins Support Speed Analysis to 25 MHz (8 MHz for A1010A and A1020A)
- Built-In High Speed Clock Distribution Network
- I/O Drive to 4 mA
- Nonvolatile, User Programmable
- Logic Fully Tested Prior to Shipment

Description

The ACT™ 1 family of field programmable gate arrays (FPGAs) offers a variety of package, speed, and application combinations. Devices are implemented in silicon gate, 1-micron or 1.2-micron two-level metal CMOS, and they employ Actel's PLICE™ antifuse technology. The unique architecture offers gate array flexibility, high performance, and instant turnaround through user programming. Device utilization is typically 95 percent of available logic modules.

ACT 1 devices also provide system designers with unique on-chip diagnostic probe capabilities, allowing convenient testing and debugging. Additional features include an on-chip clock driver with a hardwired distribution network. The network provides efficient clock distribution with minimum skew.

The user-definable I/Os are capable of driving at both TTL and CMOS drive levels. Available packages include plastic and ceramic J-leaded chip carriers, ceramic and plastic quad flatpacks, and ceramic pin grid array.

A security fuse may be programmed to disable all further programming and to protect the design from being copied or reverse engineered.

Product Family Profile

Device	A1010A A1010B	A1020A A1020B
Capacity		
Gate Array Equivalent Gates	1200	2000
PLD/LCA Equivalent Gates	3000	6000
TTL Equivalent Packages	34	53
20-Pin PAL Equivalent Packages	12	17
Logic Modules		
	295	547
Flip-Flops (maximum)		
	147	273
Routing Resources		
Horizontal Tracks/Channel	22	22
Vertical Tracks/Column	13	13
PLICE Antifuse Elements	112,000	186,000
User I/Os (maximum)		
	57	69
Packages: A1010A, A1010B, A1020A, A1020B		
	44 PLCC	44 PLCC
	68 PLCC	68 PLCC
	84 PLCC	84 PLCC
	100 PQFP	100 PQFP
A1010A, A1020A Only		
		44 JQCC
		68 JQCC
		84 JQCC
		84 CQFP
	84 CPGA	84 CPGA
Performance		
Flip-Flop Toggle Rate (maximum)	100 MHz	100 MHz
System Speed (maximum)	40 MHz	40 MHz
CMOS Process		
A1010A, A1020A	1.2 μm	1.2 μm
A1010B, A1020B	1.0 μm	1.0 μm

Note:

1. See Product Plan on page 1-4 for package availability.

The Action Logic System

The ACT 1 device family is supported by Actel's Action Logic™ System (ALS), allowing logic design implementation with minimum effort. The ALS interfaces with the resident CAE system to provide a complete gate array design environment: schematic capture, simulation, fully automatic place and route, timing verification, and device programming. The Action Logic System is available for 386™ PC and for Apollo™ and Sun™ workstations and for running Viewlogic®, Mentor Graphics®, Valid™, and OrCAD™.

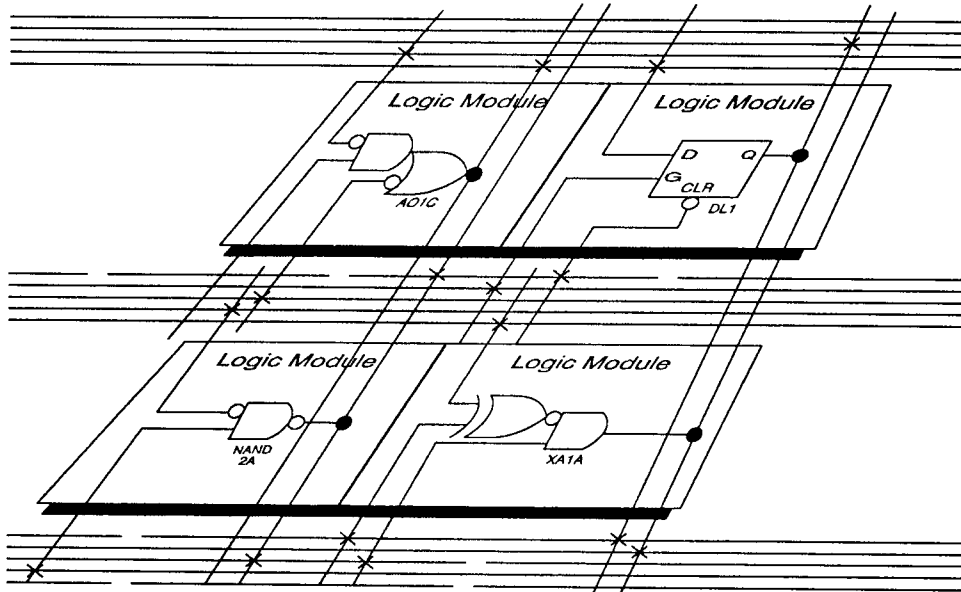


Figure 1. Partial View of an ACT 1 Device

ACT 1 Device Structure

A partial view of an ACT 1 device (Figure 1) depicts four logic modules and distributed horizontal and vertical interconnect tracks. PLICE antifuses, located at intersections of the horizontal and vertical tracks, connect logic module inputs and outputs. During programming, these antifuses are addressed and programmed to make the connections required by the circuit application.

The Actel Logic Module

The Actel logic module is an 8-input, one-output logic circuit chosen for the wide range of functions it implements and for its efficient use of interconnect routing resources (Figure 2).

The logic module can implement the four basic logic functions (NAND, AND, OR, and NOR) in gates of two, three, or four inputs. Each function may have many versions, with different combinations of active-low inputs. The logic module can also implement a variety of D-latches, exclusivity functions, AND-ORs, and OR-ANDs. No dedicated hardwired latches or flip-flops are required in the array, since latches and flip-flops may be constructed from logic modules wherever needed in the application.

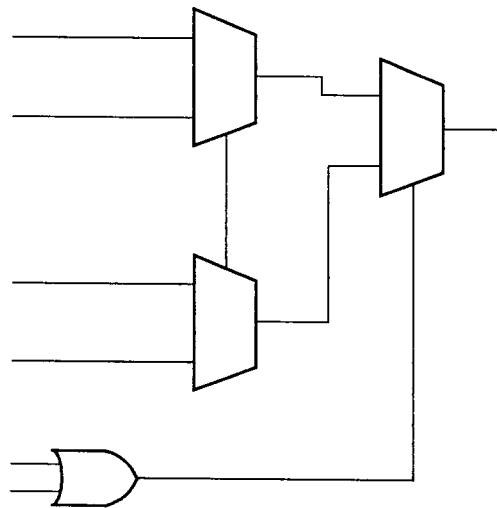


Figure 2. ACT 1 Logic Module

I/O Buffers

Each I/O pin is available as an input, output, three-state, or bidirectional buffer. Input and output levels are compatible with standard TTL and CMOS specifications. Outputs sink or source 4 mA at TTL levels. See Electrical Specifications for additional I/O buffer specifications.

Device Organization

ACT 1 devices consist of a matrix of logic modules arranged in rows separated by wiring channels. This array is surrounded by a ring of peripheral circuits including I/O buffers, testability circuits, and diagnostic probe circuits providing real-time diagnostic capability. Between rows of logic modules are routing channels containing sets of segmented metal tracks with PLICE antifuses. Each channel has 22 signal tracks. Vertical routing is permitted via 13 vertical tracks per logic module column. The resulting network allows arbitrary and flexible interconnections between logic modules and I/O modules.

Probe Pin

ACT 1 devices have two independent diagnostic probe pins. These pins allow the user to observe any two internal signals by entering the appropriate net name in the diagnostic software. Signals may be viewed on a logic analyzer using Actel's Actionprobe™ diagnostic tools. The probe pins can also be used as user-defined I/Os when debugging is finished.

ACT 1 Array Performance

Temperature and Voltage Effects

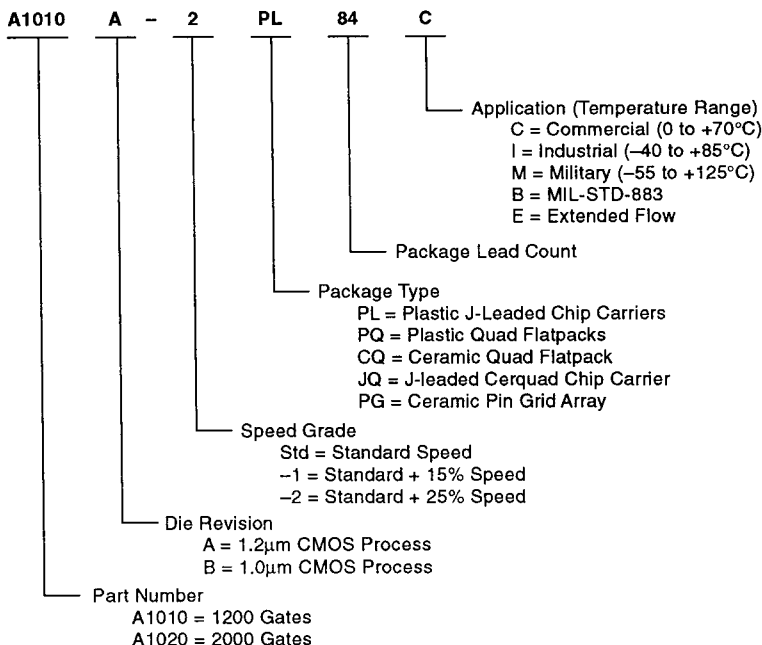
Worst-case delays for ACT 1 arrays are calculated in the same manner as for masked array products. A typical delay parameter is multiplied by a derating factor to account for temperature, voltage, and processing effects. However, in an ACT 1 array, temperature and voltage effects are less dramatic than with masked devices. The electrical characteristics of module interconnections on ACT 1 devices remain constant over voltage and temperature fluctuations.

As a result, the total derating factor from typical to worst-case for a standard speed ACT 1 array is only 1.19 to 1, compared to 2 to 1 for a masked gate array.

Logic Module Size

Logic module size also affects performance. A mask programmed gate array cell with four transistors usually implements only one logic level. In the more complex logic module (similar to the complexity of a gate array macro) of an ACT 1 array, implementation of multiple logic levels within a single module is possible. This eliminates interlevel wiring and associated RC delays. The effect is termed "net compression."

Ordering Information





Product Plan

	Speed Grade*			Application				
	Std	-1	-2	C	I	M	B	E
A1010A Device								
44-pin Plastic Leaded Chip Carrier (PL)	✓	✓	✓	✓	✓	—	—	—
68-pin Plastic Leaded Chip Carrier (PL)	✓	✓	✓	✓	✓	—	—	—
100-pin Plastic Quad Flatpack (PQ)	✓	✓	✓	✓	✓	—	—	—
84-pin Ceramic Pin Grid Array (PG)	✓	✓	—	✓	—	✓	✓	—
A1010B Device								
44-pin Plastic Leaded Chip Carrier (PL)	✓	✓	✓	✓	✓	—	—	—
68-pin Plastic Leaded Chip Carrier (PL)	✓	✓	✓	✓	✓	—	—	—
100-pin Plastic Quad Flatpack (PQ)	✓	✓	✓	✓	✓	—	—	—
A1020A Device								
44-pin Plastic Leaded Chip Carrier (PL)	✓	✓	✓	✓	✓	—	—	—
68-pin Plastic Leaded Chip Carrier (PL)	✓	✓	✓	✓	✓	—	—	—
84-pin Plastic Leaded Chip Carrier (PL)	✓	✓	✓	✓	✓	—	—	—
100-pin Plastic Quad Flatpack (PQ)	✓	✓	✓	✓	✓	—	—	—
84-pin Ceramic Pin Grid Array (PG)	✓	✓	—	✓	—	✓	✓	—
84-pin Ceramic Quad Flatpack (PQ)	✓	✓	—	✓	—	✓	✓	✓
44-pin J-leaded Cerquad Chip Carrier (JQ)	✓	✓	—	✓	—	✓	✓	—
68-pin J-leaded Cerquad Chip Carrier (JQ)	✓	✓	—	✓	—	✓	✓	—
84-pin J-leaded Cerquad Chip Carrier (JQ)	✓	✓	—	✓	—	✓	✓	—
A1020B Device								
44-pin Plastic Leaded Chip Carrier (PL)	✓	✓	✓	✓	✓	—	—	—
68-pin Plastic Leaded Chip Carrier (PL)	✓	✓	✓	✓	✓	—	—	—
84-pin Plastic Leaded Chip Carrier (PL)	✓	✓	✓	✓	✓	—	—	—
100-pin Plastic Quad Flatpack (PQ)	✓	✓	✓	✓	✓	—	—	—

Applications: C = Commercial Availability: ✓ = Available * Speed Grade: -1 = 15% faster than Standard
 I = Industrial P = Planned -2 = 25% faster than Standard
 M = Military — = Not Planned
 B = MIL-STD-883
 E = Extended Flow

Device Resources

Device Series	Logic Modules	Gates	User I/Os			
			44-pin	68-pin	84-pin	100-pin
A1010	295	1200	34	57	57	57
A1020	547	2000	34	57	69	69

Pin Description

CLK **Clock (Input)**

TTL Clock input for global clock distribution network. The Clock input is buffered prior to clocking the logic modules. This pin can also be used as an I/O.

DCLK **Diagnostic Clock (Input)**

TTL Clock input for diagnostic probe and device programming. DCLK is active when the MODE pin is HIGH. This pin functions as an I/O when the MODE pin is LOW.

GND **Ground (Input)**

Input LOW supply voltage.

I/O **Input/Output (Input, Output)**

I/O pin functions as an input, output, three-state, or bidirectional buffer. Input and output levels are compatible with standard TTL and CMOS specifications. Unused I/O pins are automatically driven LOW by the ALS software.

MODE **Mode (Input)**

The MODE pin controls the use of multifunction pins (DCLK, PRA, PRB, SDI). When the MODE pin is HIGH, the special functions are active. When the MODE pin is LOW, the pins function as I/O.

NC **No Connection**

This pin is not connected to circuitry within the device.

$\overline{\text{PRA}}$, PRA **Probe A (Output)**

The Probe A pin is used to output data from any user-defined design node within the device. This independent diagnostic pin is used in conjunction with the Probe B pin to allow real-time

diagnostic output of any signal path within the device. The Probe A pin can be used as a user-defined I/O when debugging has been completed. The pin's probe capabilities can be permanently disabled to protect the programmed designs confidentiality. PRA is active when the MODE pin is HIGH. This pin functions as an I/O when the MODE pin is LOW. A1010A and A1020A devices have inverting probes ($\overline{\text{PRA}}$). A1010B and A1020B devices have non-inverting probes (PRA) similar to ACT 2 devices.

$\overline{\text{PRB}}$, PRB **Probe B (Output)**

The Probe B pin is used to output data from any user-defined design node within the device. This independent diagnostic pin is used in conjunction with the Probe A pin to allow real-time diagnostic output of any signal path within the device. The Probe B pin can be used as a user-defined I/O when debugging has been completed. The pin's probe capabilities can be permanently disabled to protect the programmed design's confidentiality. PRB is active when the MODE pin is HIGH. This pin functions as an I/O when the MODE pin is LOW. A1010A and A1020A devices have inverting probes ($\overline{\text{PRB}}$). A1010B and A1020B devices have non-inverting probes (PRB) similar to ACT 2 devices.

SDI **Serial Data Input (Input)**

Serial data input for diagnostic probe and device programming. SDI is active when the MODE pin is HIGH. This pin functions as an I/O when the MODE pin is LOW.

VCC **Supply Voltage (Input)**

Input HIGH supply voltage.

VPP **Programming Voltage (Input)**

Input supply voltage used for device programming. This pin must be connected to V_{CC} during normal operation.

Absolute Maximum Ratings

Free air temperature range

Symbol	Parameter	Limits	Units
V_{CC}	DC Supply Voltage ¹	-0.5 to +7.0	Volts
V_I	Input Voltage	-0.5 to $V_{CC} + 0.5$	Volts
V_O	Output Voltage	-0.5 to $V_{CC} + 0.5$	Volts
I_{IK}	Input Clamp Current	± 20	mA
I_{OK}	Output Clamp Current	± 20	mA
I_{OK}	Continuous Output Current	± 25	mA
T_{STG}	Storage Temperature	-65 to +150	$^{\circ}C$

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. Exposure to absolute maximum rated conditions for extended periods may affect device reliability. Device should not be operated outside the Recommended Operating Conditions.

Note:

1. $V_{PP} = V_{CC}$, except during device programming.

Recommended Operating Conditions

Parameter	Commercial	Industrial	Military	Units
Temperature Range ¹	0 to +70	-40 to +85	-55 to +125	$^{\circ}C$
Power Supply Tolerance	± 5	± 10	± 10	$\%V_{CC}$

Note:

1. Ambient temperature (T_A) used for commercial and industrial; case temperature (T_C) used for military.

Electrical Specifications

Parameter	Commercial		Industrial		Military		Units
	Min.	Max.	Min.	Max.	Min.	Max.	
V_{OH}^1	$(I_{OH} = -4 \text{ mA})$		3.84				V
	$(I_{OH} = -3.2 \text{ mA})$				3.7		V
V_{OL}^1	$(I_{OL} = 4 \text{ mA})$		0.33		0.40		V
V_{IL}	-0.3	0.8	-0.3	0.8	-0.3	0.8	V
V_{IH}	2.0	$V_{CC} + 0.3$	2.0	$V_{CC} + 0.3$	2.0	$V_{CC} + 0.3$	V
Input Transition Time t_R, t_F^2			500		500		ns
C_{IO} I/O Capacitance ^{2,3}			10		10		pF
Standby Current, I_{CC}^4			10		20		mA
Leakage Current ⁵	-10	10	-10	10	-10	10	μA
I_{OS} Output Short Circuit Current ⁶	$(V_O = V_{CC})$		140		140		mA
	$(V_O = GND)$		-100		-100		mA

Notes:

1. Only one output tested at a time. $V_{CC} = \text{min.}$
2. Not tested, for information only.
3. Includes worst-case 84-pin PLCC package capacitance. $V_{OUT} = 0 \text{ V}$, $f = 1 \text{ MHz}$.
4. Typical standby current = 3 mA. All outputs unloaded. All inputs = V_{CC} or GND.
5. $V_O, V_{IN} = V_{CC}$ or GND.
6. Only one output tested at a time. Min. at $V_{CC} = 4.5 \text{ V}$; Max. at $V_{CC} = 5.5 \text{ V}$.

Package Thermal Characteristics

The device junction to case thermal characteristics is θ_{jc} , and the junction to ambient air characteristics is θ_{ja} . The thermal characteristics for θ_{ja} are shown with two different air flow rates.

Maximum junction temperature is 150°C.

A sample calculation of the maximum power dissipation for an 84-pin plastic leaded chip carrier at commercial temperature is as follows:

$$\frac{\text{Max junction temp. (}^\circ\text{C)} - \text{Max commercial temp. (}^\circ\text{C)}}{\theta_{ja} (\text{}^\circ\text{C/W)}} = \frac{150^\circ\text{C} - 70^\circ\text{C}}{44^\circ\text{C/W}} = 1.82 \text{ W}$$

Package Type	Pin Count	θ_{jc}	θ_{ja} Still Air	θ_{ja} 300 ft/min	Units
Plastic J-leaded Chip Carrier	44	15	52	40	$^\circ\text{C/W}$
	68	13	45	35	$^\circ\text{C/W}$
	84	12	44	33	$^\circ\text{C/W}$
Plastic Quad Flatpack	100	13	55	47	$^\circ\text{C/W}$
Ceramic Pin Grid Array	84	8	33	20	$^\circ\text{C/W}$
Ceramic Quad Flatpack	84	5	40	30	$^\circ\text{C/W}$
J-leaded Cerquad Chip Carrier	44	8	38	30	$^\circ\text{C/W}$
	68	8	35	25	$^\circ\text{C/W}$
	84	8	34	24	$^\circ\text{C/W}$

Power Dissipation

The following formula is used to calculate total device dissipation.

$$\text{Total Device Power (mW)} = (0.20 \times N \times F1) + (0.085 \times M \times F2) + (0.80 \times P \times F3)$$

Where:

F1 = Average logic module switching rate in MHz

F2 = CLKBUF macro switching rate in MHz

F3 = Average I/O module switching rate in MHz

M = Number of logic modules connected to the CLKBUF macro

N = Total number of logic modules used in the design (including M)

P = Number of outputs loaded with 50 pF

Average switching rate of logic modules and of I/O modules is some fraction of the device operating frequency (usually CLKBUF). Logic modules and I/O modules switch states (from low-to-high or from high-to-low) only if the input data changes when the module is enabled. A conservative estimate for average logic module and I/O module switching rates (variables F1 and F3, respectively) is 10% of device clock driver frequency.

If the CLKBUF macro is not used in the design, eliminate the second term (including F2 and M variables) from the formula.

Sample A1020 Device Power Calculation

To illustrate the power calculation, consider a large design operating at high frequency. This sample design utilizes 85% of available logic modules on the A1020-series device (.85 x 547 = 465 logic modules used). The design contains 104 flip-flops (208 logic modules). Operating frequency of the design is 16 MHz. In this design, the CLKBUF macro drives the clock network. Logic modules and I/O modules are switching states at approximately 10% of the clock frequency rate (.10 x 16 MHz = 1.6 MHz). Sixteen outputs are loaded with 50 pF.

To summarize the design described above: N = 464; M = 208; F2 = 16; F1 = 4; F3 = 4; P = 16. Total device power can be calculated by substituting these values for variables in the device dissipation formula.

Total device power for this example =

$$(0.20 \times 465 \times 1.6) + (0.085 \times 208 \times 16) + (0.80 \times 16 \times 1.6) = 452 \text{ mW}$$

Functional Timing Tests

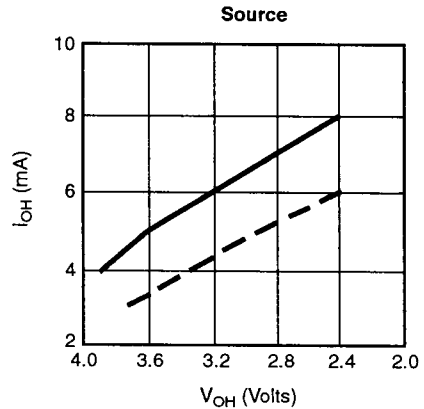
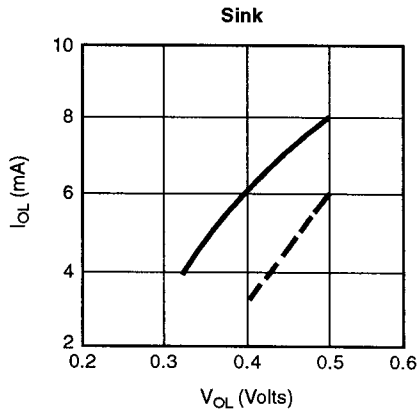
AC timing for logic module internal delays is determined after place and route. The ALS Timer utility displays actual timing parameters for circuit delays. ACT 1 devices are AC tested to a "binning" circuit specification.

The circuit consists of one input buffer + n logic modules + one output buffer (n = 16 for A1010; n = 28 for A1020). The logic

modules are distributed along two sides of the device, as inverting or non-inverting buffers. The modules are connected through programmed antifuses with typical capacitive loading.

Propagation delay [$t_{PD} = (t_{PLH} + t_{PHL})/2$] is tested to the following AC test specifications.

Output Buffer Performance Derating



- - - - - Military, worst-case values at 125°C, 4.5 V.
 ————— Commercial, worst-case values at 70°C, 4.75 V.

Note:

The above curves are based on characterizations of sample devices and are not completely tested on all devices.

Timing Derating

Operating temperature, operating voltage, and device processing conditions, along with device die size and speed grade, account for variations in array timing characteristics. These variations are summarized into a derating factor for ACT 1 array typical timing

specifications. The derating factors shown in the table below are based on the recommended operating conditions for ACT 1 commercial, industrial, and military applications. The derating curves show worst-to-best case operating voltage range and best-to-worst case operating temperature range.

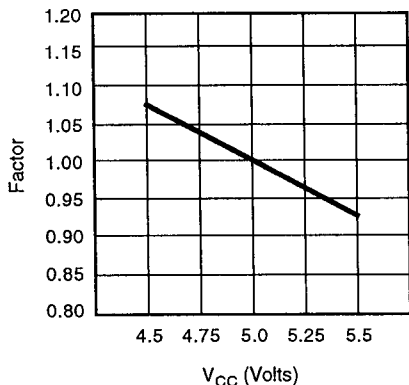
Timing Derating Factor (x typical)

Device	Commercial		Industrial		Military	
	Best-Case	Worst-Case	Best-Case	Worst-Case	Best-Case	Worst-Case
A1010A, A1020A						
Standard Speed	0.45	1.54	0.40	1.65	0.37	1.79
-1 Speed Grade	0.45	1.28	0.40	1.37	0.37	1.49
-2 Speed Grade	0.45	1.13	0.40	1.20	0.37	1.32
A1010B, A1020B						
Standard Speed	0.45	1.54	0.40	1.65		
-1 Speed Grade	0.45	1.28	0.40	1.37		
-2 Speed Grade	0.45	1.13	0.40	1.20		

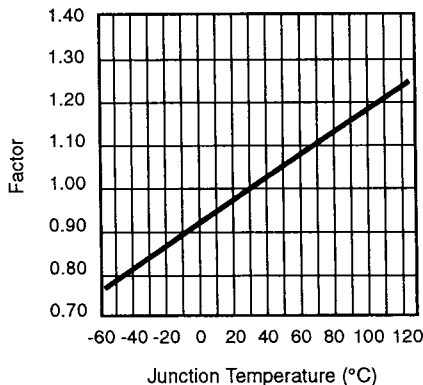
Note: "Best-case" reflects maximum operating voltage, minimum operating temperature, and best-case processing. "Worst-case" reflects minimum

operating voltage, maximum operating temperature, and worst-case processing. Best-case derating is based on sample data only and is not guaranteed.

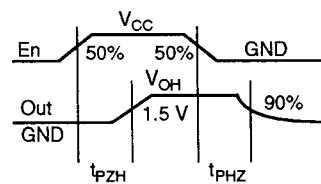
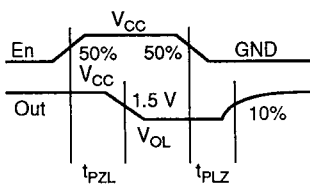
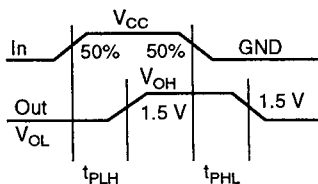
Voltage Derating Curve



Temperature Derating Curve

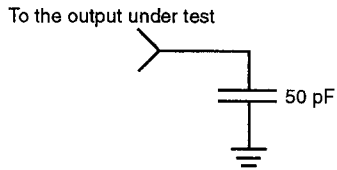


Output Buffer Delays

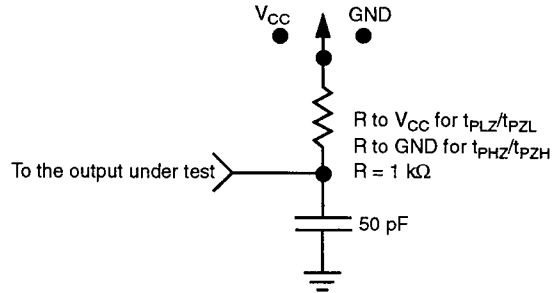


AC Test Loads

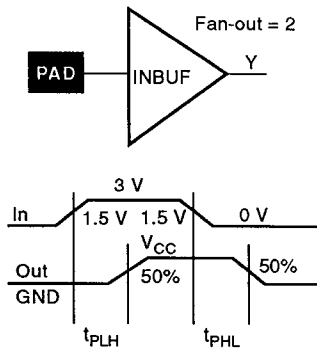
Load 1
(Used to measure propagation delay)



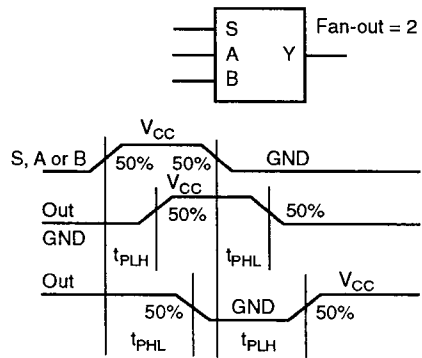
Load 2
(Used to measure rising/falling edges)



Input Buffer Delays

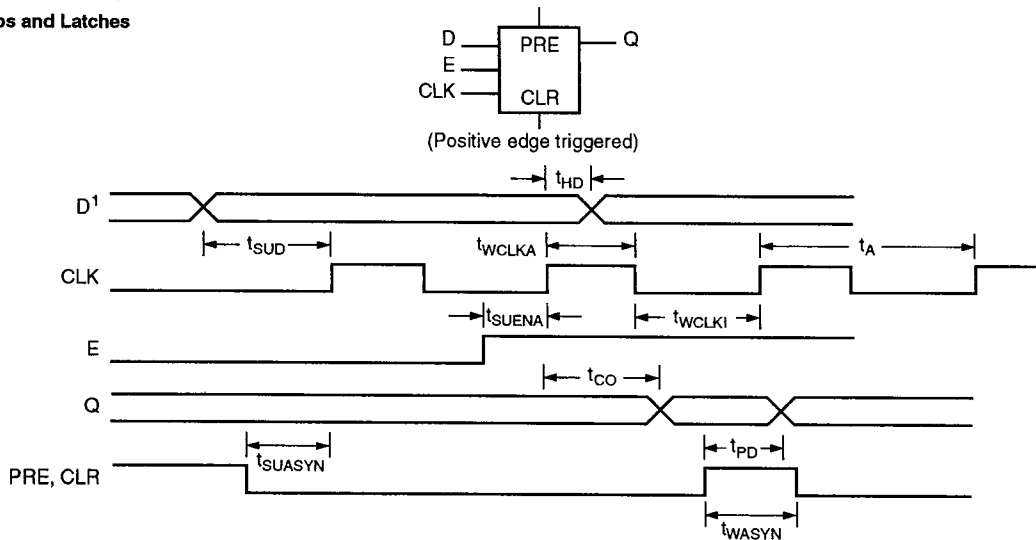


Module Delays



Sequential Timing Characteristics

Flip-Flops and Latches



Notes:

1. D represents all data functions involving A, B, and S for multiplexed flip-flops.

Timing Characteristics

Timing is design-dependent; actual delay values are determined after place and route of the design using the ALS Timer utility. The following delay values use statistical estimates for wiring delays based on 85% to 90% module utilization. Device utilization above 95% will result in performance degradation.

With ALS place and route programs, the user can assign criticality level to a net, based on timing requirements. Delays for

both typical and critical (speed-sensitive) nets are given below. Most nets will fall into the “typical” category.

Less than 1% of all routing in a design requires the use of “long tracks.” Long tracks, long vertical or horizontal routing paths, are used by the autorouter only as needed. Delays due to the use of long tracks range from 15 ns to 35 ns. Long tracks may be used to route the least critical nets in a given design.

Logic Module Timing

$V_{CC} = 5.0$ V; $T_J = 25^\circ\text{C}$; Process = Typical; $t_{PD} = 3.0$ ns @ FO = 0

Single Logic Module Macros
(e.g., most gates, latches, multiplexors)¹

Parameter	Output Net	FO = 1	FO = 2	FO = 3	FO = 4	FO = 8	Units
t_{PD}	Critical	5.4	5.8	6.2	8.5	Note 2	ns
t_{PD}	Typical	6.3	6.7	7.7	8.6	10.8	ns

Dual Logic Module Macros
(e.g., adders, wide input gates)¹

Parameter	Output Net	FO = 1	FO = 2	FO = 3	FO = 4	FO = 8	Units
t_{PD}	Critical	9.2	9.6	10.0	12.3	Note 2	ns
t_{PD}	Typical	10.2	10.6	11.6	12.5	14.6	ns

Sequential Element Timing Characteristics

Parameter		Fan-Out					Units
		FO = 1	FO = 2	FO = 3	FO = 4	FO = 8	
t_{SU}	Set Up Time, Data Latches	3.5	3.9	4.2	4.5	4.8	ns
t_{SU}	Set Up Time, Flip-Flops	3.9	3.9	3.9	3.9	3.9	ns
t_H	Hold Time	0	0	0	0	0	ns
t_W	Pulse Width, Minimum ³	7.7	8.5	9.2	10.0	14.0	ns
t_{CQ}	Delay, Critical Net	5.4	5.8	6.2	8.5	Note 2	ns
t_{CQ}	Delay, Typical Net	6.3	6.7	7.7	8.6	10.8	ns

Notes:

1. Most flip-flops exhibit single module delays.
2. Critical nets have a maximum fan-out of six.
3. Minimum pulse width, t_W , applies to CLK, PRE, and CLR inputs.

I/O Buffer Timing

$V_{CC} = 5.0\text{ V}$; $T_J = 25^\circ\text{C}$; Process = Typical

INBUF Macros

Parameter	From – To	FO = 1	FO = 2	FO = 3	FO = 4	FO = 8	Units
t_{PHL}	Pad to Y	6.9	7.6	8.9	10.7	14.3	ns
t_{PLH}	Pad to Y	5.9	6.5	7.7	8.4	12.4	ns

CLKBUF (High Fan-Out Clock Buffer) Macros

Parameter	FO = 40	FO = 160	FO = 320	Units
t_{PHL}	9.0	12.0	15.0	ns
t_{PLH}	9.0	12.0	15.0	ns

Notes:

1. A clock balancing feature is provided to minimize clock skew.
2. There is no limit to the number of loads that may be connected to the CLKBUF macro.

OUTBUF, TRIBUF, and BIBUF Macros¹

$C_L = 50\text{ pF}$

Parameter	From – To	CMOS	TTL	Units
t_{PHL}	D to Pad	3.9	4.9	ns
t_{PLH}	D to Pad	7.2	5.7	ns
t_{PHZ}	E to Pad	5.2	3.4	ns
t_{PZH}	E to Pad	6.5	4.9	ns
t_{PLZ}	E to Pad	6.9	5.2	ns
t_{PZL}	E to Pad	4.9	5.9	ns

Change in Propagation Delay with Load Capacitance²

Parameter	From – To	CMOS	TTL	Units
t_{PHL}	D to Pad	0.03	0.046	ns/pF
t_{PLH}	D to Pad	0.07	0.039	ns/pF
t_{PHZ}	E to Pad	0.08	0.046	ns/pF
t_{PZH}	E to Pad	0.07	0.039	ns/pF
t_{PLZ}	E to Pad	0.07	0.039	ns/pF
t_{PZL}	E to Pad	0.03	0.039	ns/pF

Notes:

1. The BIBUF macro input section exhibits the same delays as the INBUF macro.
2. Load capacitance delay delta can be extrapolated down to 15 pF minimum.

Example:

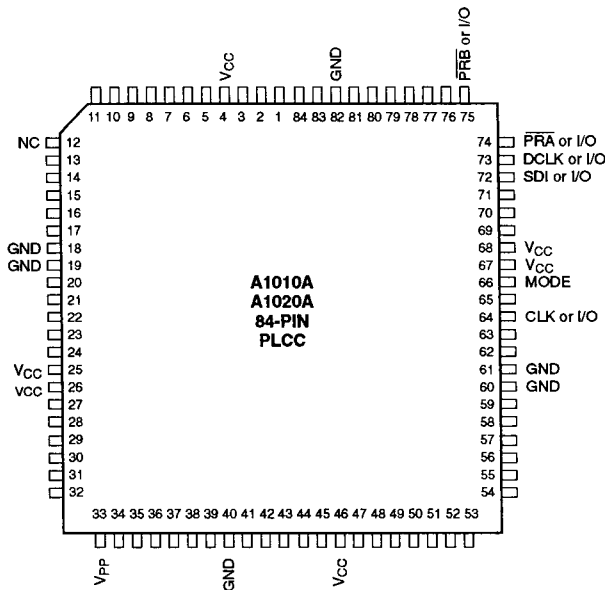
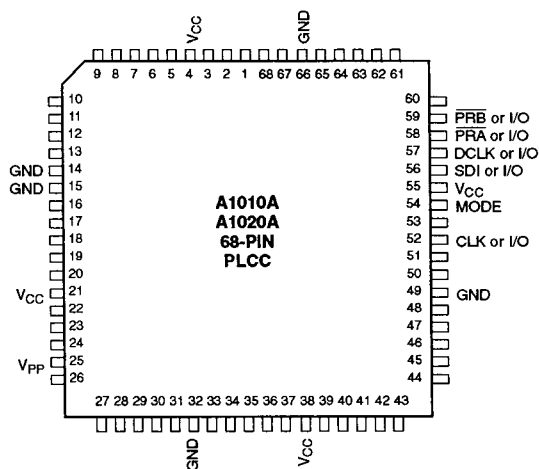
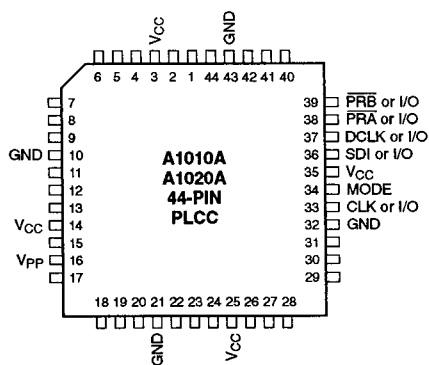
Delay for OUTBUF driving a 100-pF TTL load:

$$t_{PHL} = 4.9 + (.046 \times (100 - 50)) = 4.9 + 2.3 = 7.2\text{ ns}$$

$$t_{PLH} = 5.7 + (.039 \times (100 - 50)) = 5.7 + 2.0 = 7.7\text{ ns}$$

Package Pin Assignments

(Top View)

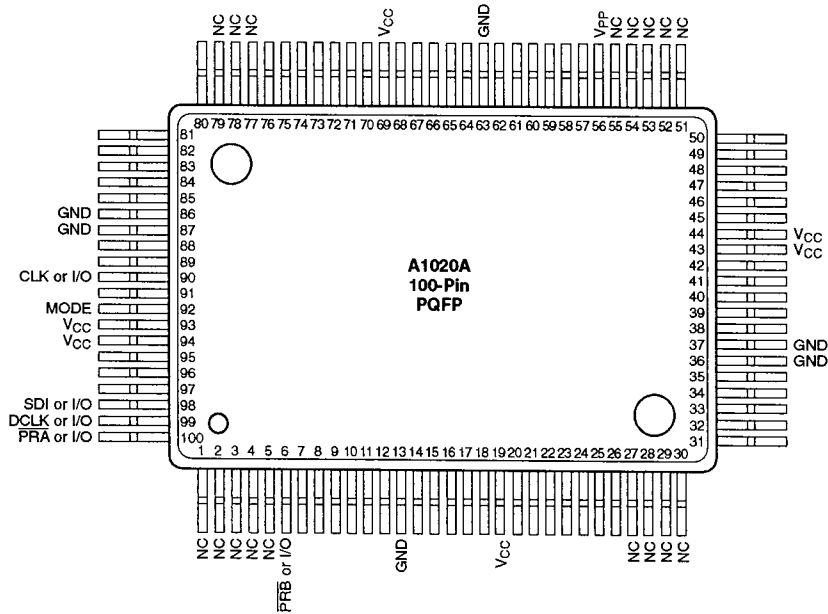
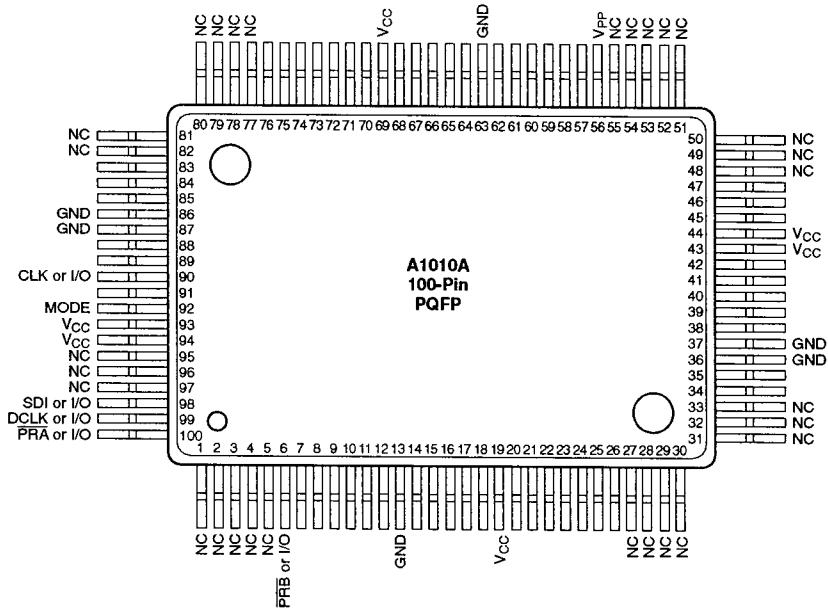


Notes:

1. V_{PP} must be terminated to V_{CC} , except during device programming.
2. MODE must be terminated to circuit ground, except during device programming or debugging.
3. Unused I/O pins are designated as outputs by ALS and are driven low.
4. All unassigned pins are available for use as I/Os.

Package Pin Assignments (continued)

(Top View)

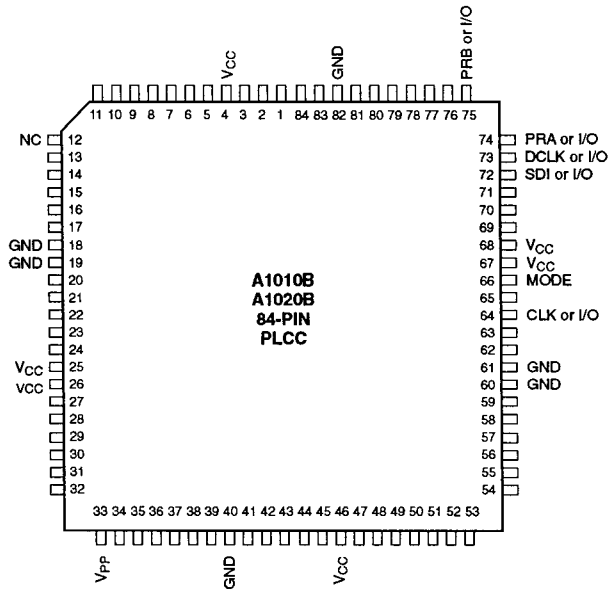
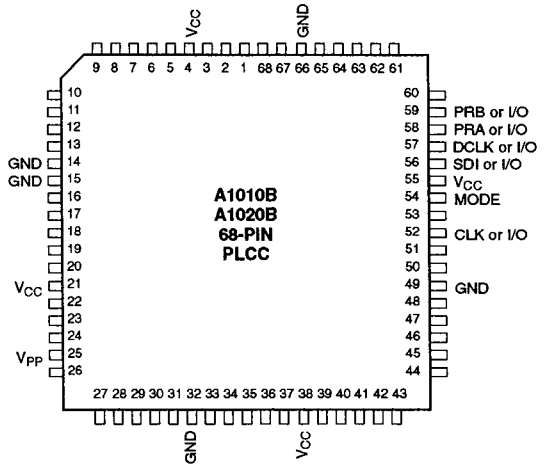
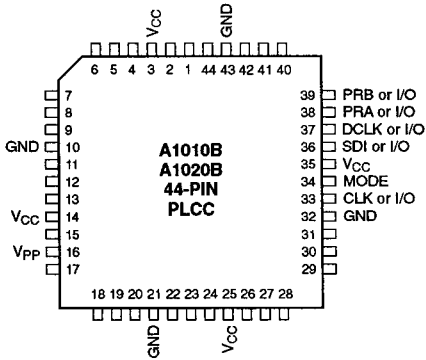


Notes:

1. V_{pp} must be terminated to V_{CC} , except during device programming.
2. $MODE$ must be terminated to circuit ground, except during device programming or debugging.
3. Unused I/O pins are designated as outputs by ALS and are driven low.
4. All unassigned pins are available for use as I/Os.

Package Pin Assignments

(Top View)

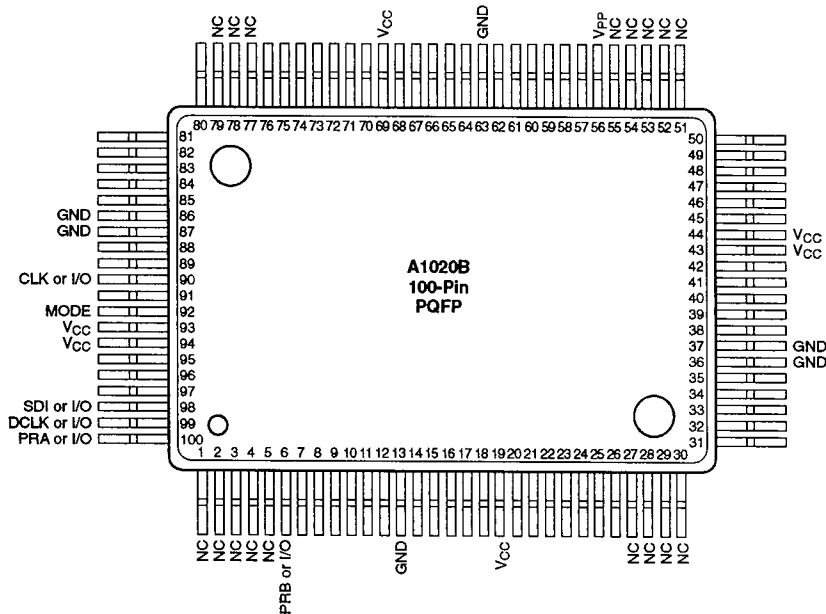
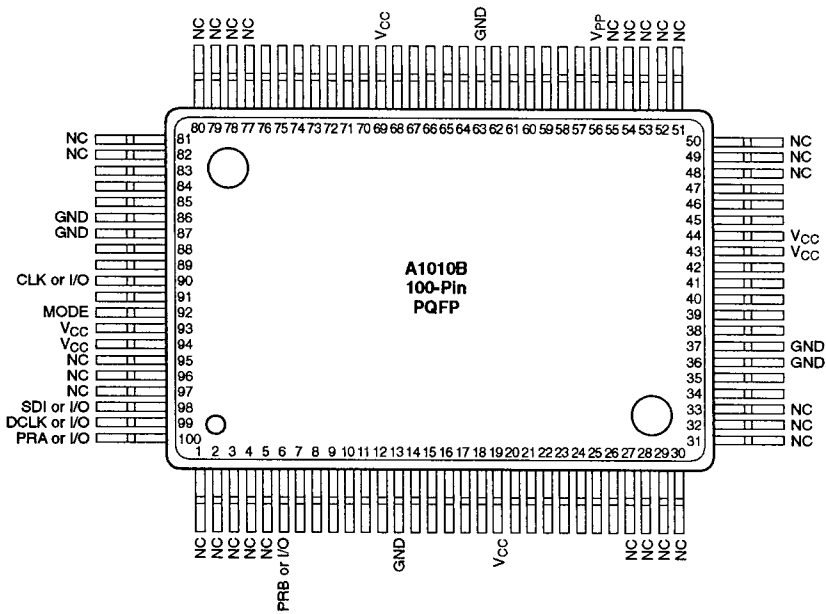


Notes:

1. V_{pp} must be terminated to V_{CC} , except during device programming.
2. MODE must be terminated to circuit ground, except during device programming or debugging.
3. Unused I/O pins are designated as outputs by ALS and are driven low.
4. All unassigned pins are available for use as I/Os.

Package Pin Assignments (continued)

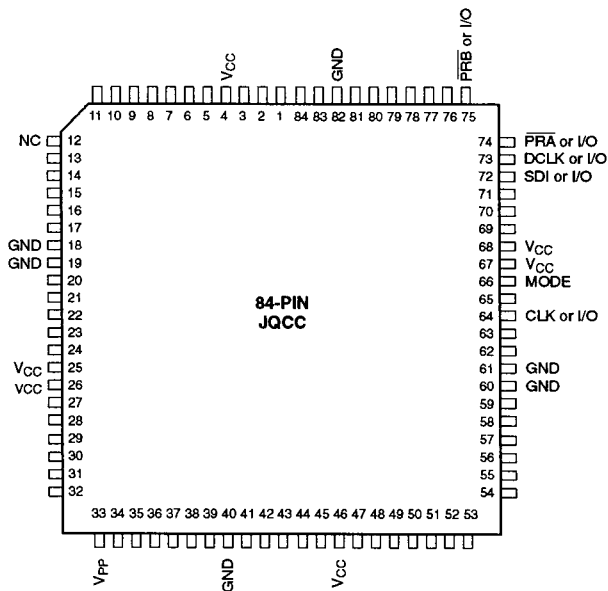
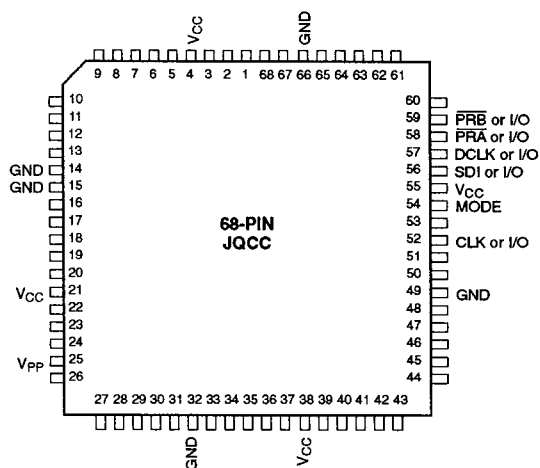
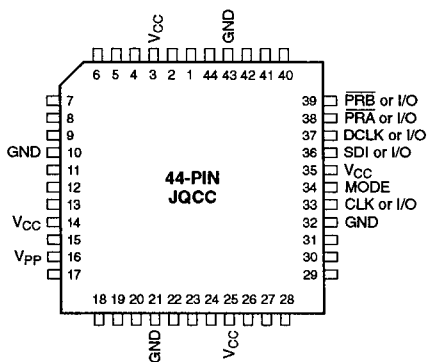
(Top View)



Notes:

1. V_{pp} must be terminated to V_{CC} , except during device programming.
2. MODE must be terminated to circuit ground, except during device programming or debugging.
3. Unused I/O pins are designated as outputs by ALS and are driven low.
4. All unassigned pins are available for use as I/Os.

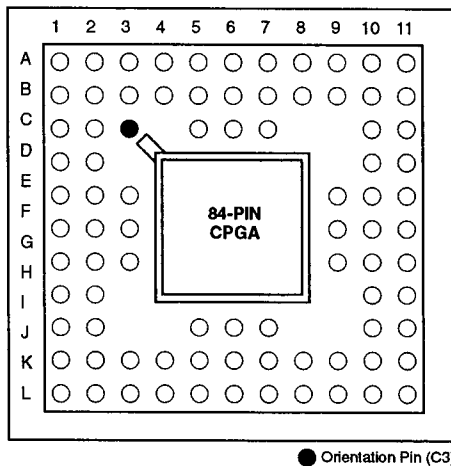
Package Pin Assignments (continued)



Notes:

1. V_{pp} must be terminated to V_{CC}, except during device programming.
2. MODE must be terminated to circuit ground, except during device programming and debugging.
3. Unused I/O pins are designated as outputs by ALS and are driven low.
4. All unassigned pins are available for use as I/Os.

Package Pin Assignments (continued)

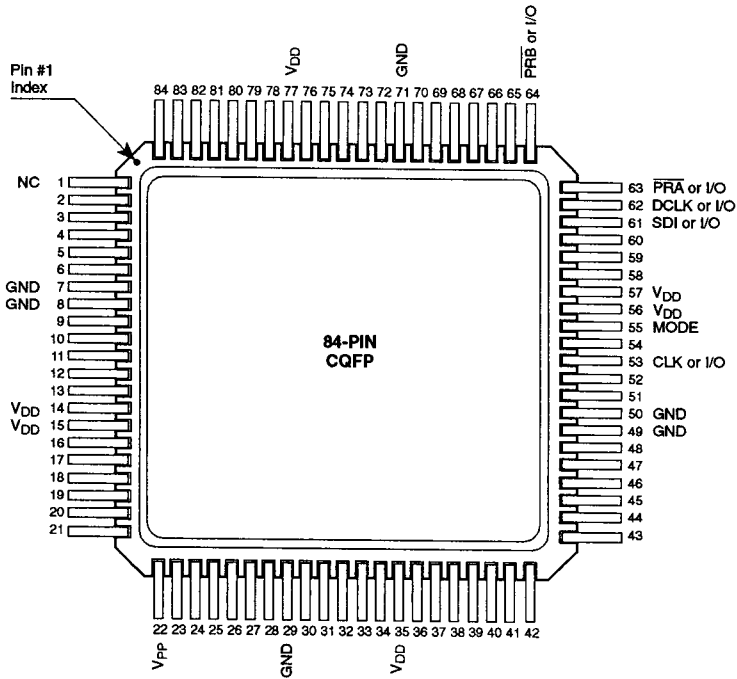


Signal	A1010-Series Devices	A1020A-Series Devices
$\overline{\text{PRA}}$	A11	A11
$\overline{\text{PRB}}$	B10	B10
MODE	E11	E11
SDI	B11	B11
DCKL	C10	C10
V_{PP}	K2	K2
CLK or I/O	F9	F9
GND	B7, E2, E3, K5, F10, G10	B7, E2, E3, K5, F10, G10
V_{CC}	B5, F1, G2, K7, E9, E10	B5, F1, G2, K7, E9, E10
N/C (No Connection)	B1, B2, C1, C2, K1, J2, J10, K10, K11, C11, D10, D11	B2

Notes:

- V_{pp} must be terminated to V_{CC} , except during device programming.
- MODE must be terminated to circuit ground, except during device programming or debugging.
- Unused I/O pins are designated as outputs by ALS and are driven low.
- All unassigned pins are available for use as I/Os.

Package Pin Assignments (continued)



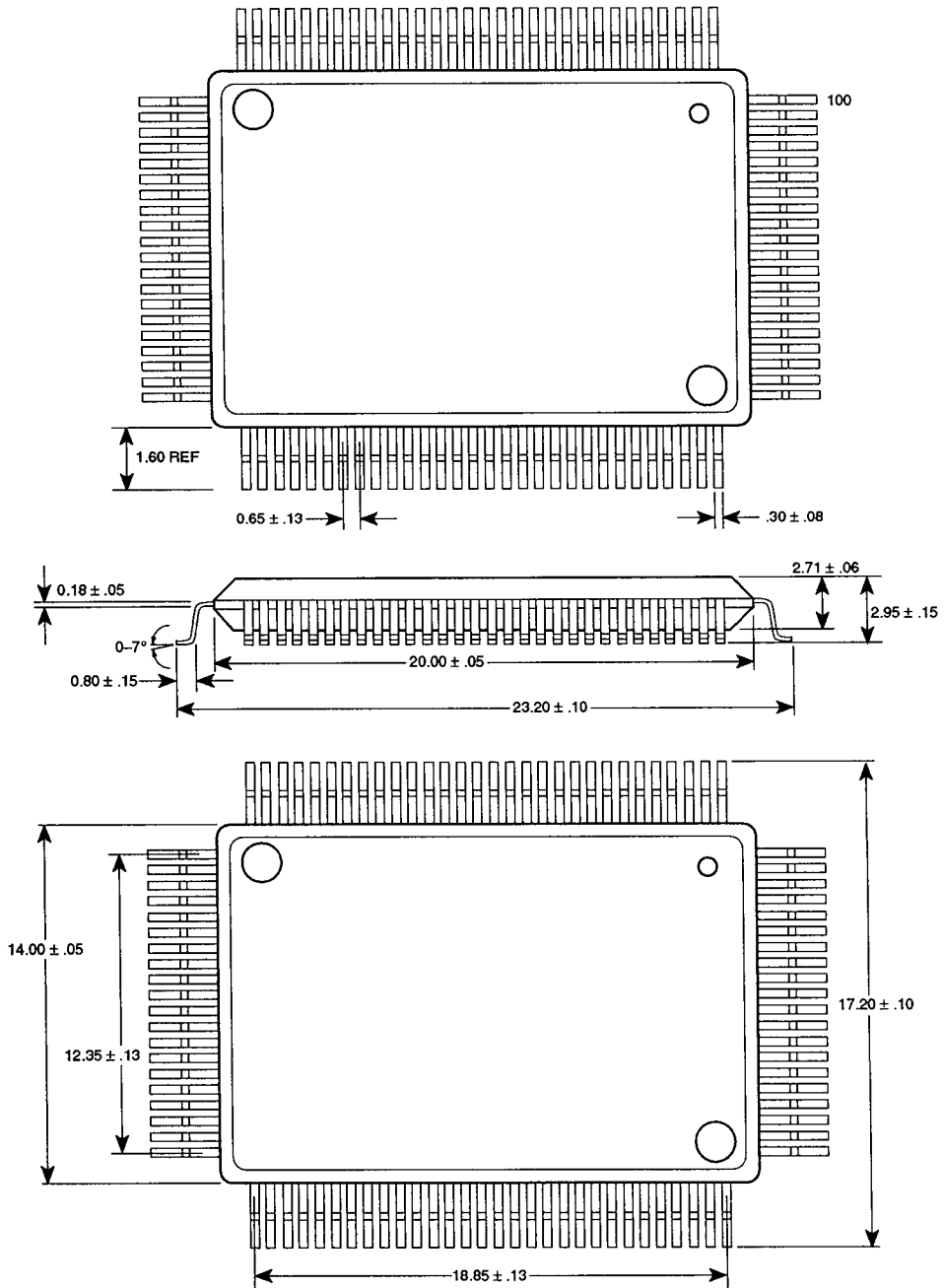
Notes:

1. V_{pp} must be terminated to V_{CC} , except during device programming.
2. MODE must be terminated to circuit ground, except during device programming or debugging.
3. Unused I/O pins are designated as outputs by ALS and are driven low.
4. All unassigned pins are available for use as I/Os.

Package Mechanical Details (continued)

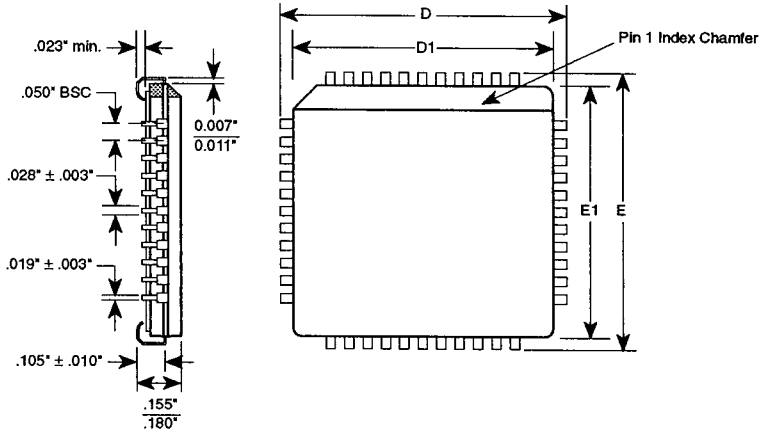
Plastic Quad Flatpack

Dimensions in millimeters



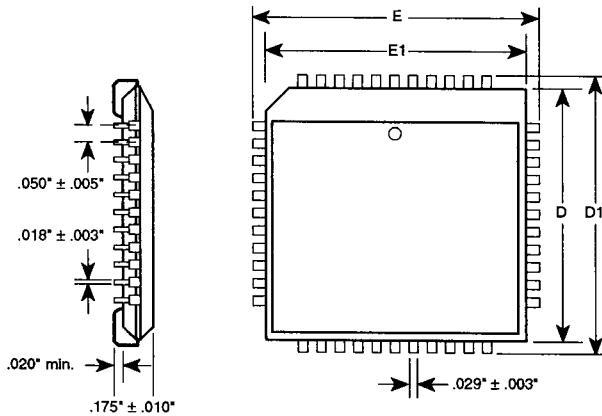
Package Mechanical Details

J-Leaded Cerquad Chip Carrier



Lead Count	D,E	D1, E1
44	.690" ± .005"	.650" ± .008"
68	.990" ± .005"	.950" ± .008"
84	1.190" ± .005"	1.150" ± .008"

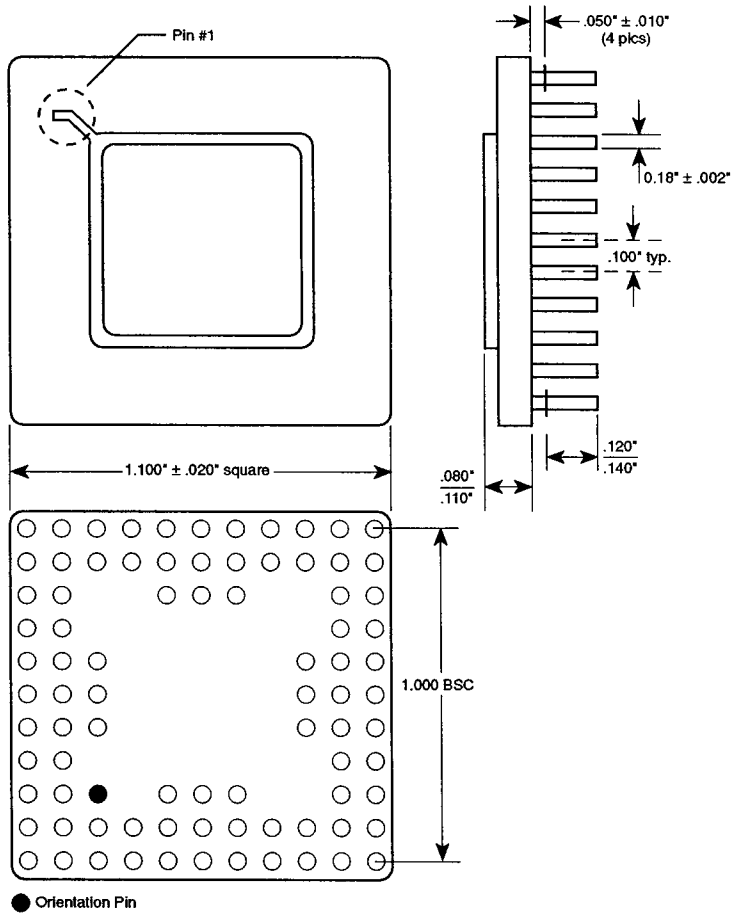
Plastic J-Leaded Chip Carrier



Lead Count	D,E	D1, E1
44	.690" ± .005"	.655" ± .005"
68	.990" ± .005"	.955" ± .005"
84	1.190" ± .005"	1.155" ± .005"

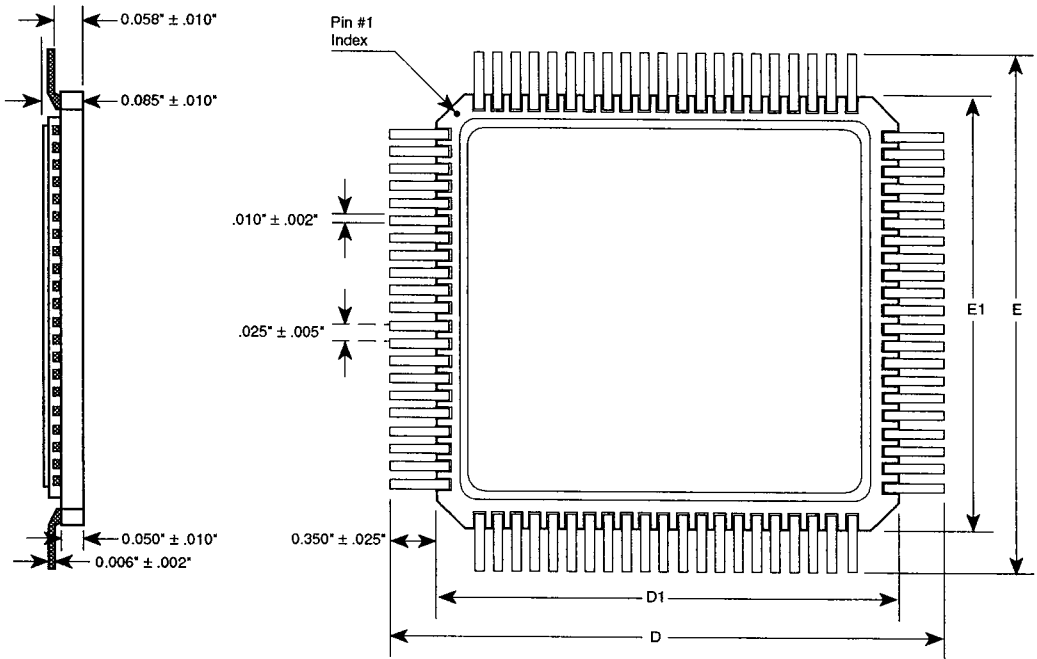
Package Mechanical Details (continued)

Ceramic Pin Grid Array



Package Mechanical Details (continued)

Ceramic Quad Flatpack



Lead Count	D,E	D1, E1
84	$1.350'' \pm .030''$	$0.650'' \pm .010''$